

FEATURES

- **3.8nV/ $\sqrt{\text{Hz}}$ Input Noise Voltage**
- **3.7mA Supply Current**
- **200MHz Gain Bandwidth**
- **Low Total Harmonic Distortion: -85dBc at 1MHz**
- **70V/ μs Slew Rate**
- **400 μV Maximum Input Offset Voltage**
- **300nA Maximum Input Bias Current**
- **Unity-Gain Stable**
- **Capacitive Load Stable Up to 100pF**
- **23mA Minimum Output Current**
- **Specified at $\pm 5\text{V}$ and Single 5V**

APPLICATIONS

- Video and RF Amplification
- ADSL, HDSL II, VDSL Receivers
- Active Filters
- Wideband Amplifiers
- Buffers
- Data Acquisition Systems

DESCRIPTION

The LT[®]1722/LT1723/LT1724 are single/dual/quad, low noise, low power, high speed operational amplifiers. These products feature lower input offset voltage, lower input bias current and higher DC gain than devices with comparable bandwidth. The 200MHz gain bandwidth ensures high open-loop gain at video frequencies.

The low input noise voltage is achieved with reduced supply current. The total noise is optimized for a source resistance between 0.8k and 12k. Due to the input bias current cancellation technique used, the resistance seen by each input does not need to be balanced.

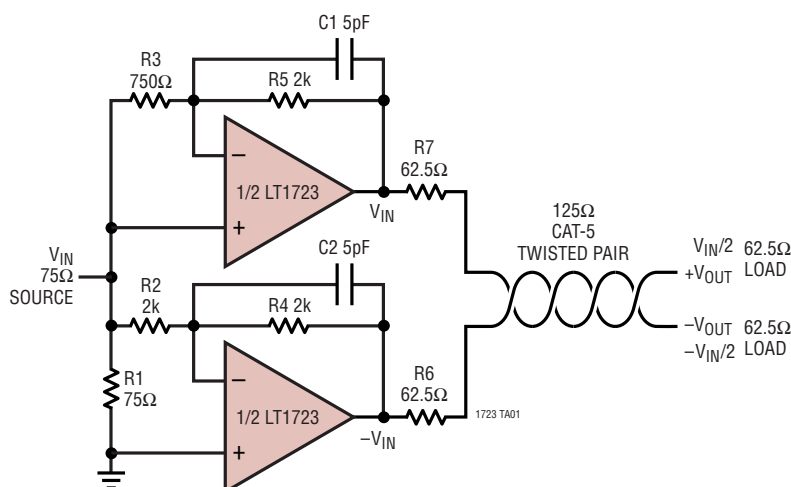
The output drives a 150 Ω load to $\pm 3\text{V}$ with $\pm 5\text{V}$ supplies. On a single 5V supply the output swings from 1.5V to 3.5V with a 500 Ω load connected to 2.5V. The amplifier is unity-gain stable ($C_{\text{LOAD}} \leq 100\text{pF}$).

The LT1722/LT1723/LT1724 are manufactured on Linear Technology's advanced low voltage complementary bipolar process. The LT1722 is available in the SO-8 and 5-pin SOT-23 packages. The LT1723 is available in the SO-8 and MS8 packages. The LT1724 is available in the 14-lead SO package.

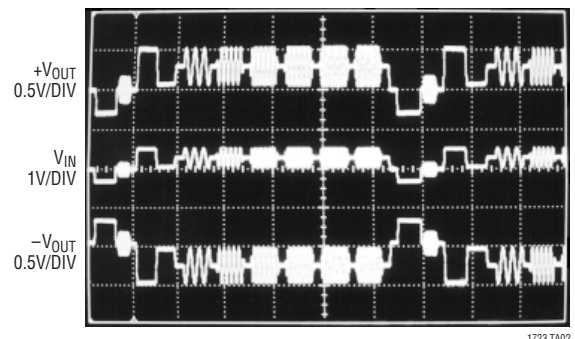
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TYPICAL APPLICATION

Differential Video Line Driver



Line Driver Multiburst Video Signal



1723 TA02

172234fa

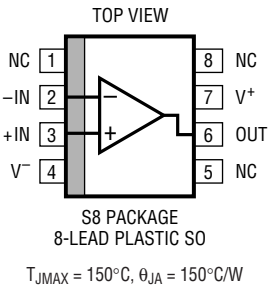
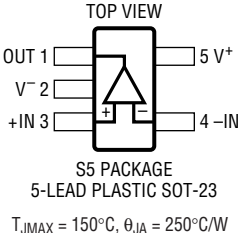
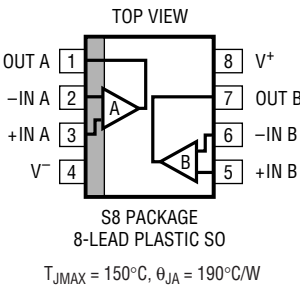
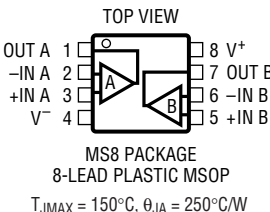
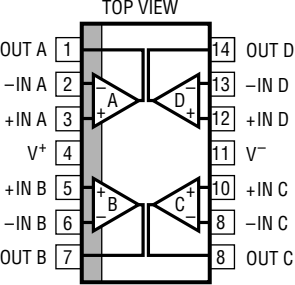
LT1722/LT1723/LT1724

ABSOLUTE MAXIMUM RATINGS (Note 1)

Total Supply Voltage (V^+ to V^-) 12.6V
 Input Voltage $\pm V_S$
 Differential Input Voltage (Note 2) $\pm 0.7V$
 Input Current (Note 2) $\pm 10mA$
 Output Short-Circuit Duration (Note 3) Indefinite

Operating Temperature Range (Note 4) ... $-40^\circ C$ to $85^\circ C$
 Specified Temperature Range (Note 5) ... $-40^\circ C$ to $85^\circ C$
 Maximum Junction Temperature $150^\circ C$
 Storage Temperature Range $-65^\circ C$ to $150^\circ C$
 Lead Temperature (Soldering, 10 sec) $300^\circ C$

PACKAGE/ORDER INFORMATION

 S8 PACKAGE 8-LEAD PLASTIC SO $T_{JMAX} = 150^\circ C, \theta_{JA} = 150^\circ C/W$	ORDER PART NUMBER LT1722CS8 LT1722IS8 S8 PART MARKING 1722 1722I	 S5 PACKAGE 5-LEAD PLASTIC SOT-23 $T_{JMAX} = 150^\circ C, \theta_{JA} = 250^\circ C/W$	ORDER PART NUMBER LT1722CS5 LT1722IS5 S5 PART MARKING* LTZB
 S8 PACKAGE 8-LEAD PLASTIC SO $T_{JMAX} = 150^\circ C, \theta_{JA} = 190^\circ C/W$	ORDER PART NUMBER LT1723CS8 LT1723IS8 S8 PART MARKING 1723 1723I	 MS8 PACKAGE 8-LEAD PLASTIC MSOP $T_{JMAX} = 150^\circ C, \theta_{JA} = 250^\circ C/W$	ORDER PART NUMBER LT1723CMS8 LT1723IMS8 MS8 PART MARKING LTYC LTZA
 S PACKAGE 14-LEAD PLASTIC SO $T_{JMAX} = 150^\circ C, \theta_{JA} = 100^\circ C/W$	ORDER PART NUMBER LT1724CS LT1724IS		

Consult LTC Marketing for parts specified with wider operating temperature ranges.

*The temperature grades are identified by a label on the shipping container.

ELECTRICAL CHARACTERISTICS

$T_A = 25^\circ\text{C}$, $V_S = \pm 5\text{V}$, $V_{CM} = 0\text{V}$, unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
V_{OS}	Input Offset Voltage	(Note 6) LT1722 SOT-23 and LT1723 MS8		100 150	400 650	μV μV
I_{OS}	Input Offset Current			40	300	nA
I_B	Input Bias Current			40	300	nA
e_n	Input Noise Voltage	$f = 10\text{kHz}$		3.8		$\text{nV}/\sqrt{\text{Hz}}$
i_n	Input Noise Current	$f = 10\text{kHz}$		1.2		$\text{pA}/\sqrt{\text{Hz}}$
R_{IN}	Input Resistance	$V_{CM} = \pm 3.5\text{V}$ Differential	5	35 50		$\text{M}\Omega$ $\text{k}\Omega$
C_{IN}	Input Capacitance			2		pF
	Input Voltage Range + Input Voltage Range –		3.5	4 –4	–3.5	V V
CMRR	Common Mode Rejection Ratio	$V_{CM} = \pm 3.5\text{V}$	80	100		dB
PSRR	Power Supply Rejection Ratio	$V_S = \pm 2.3\text{V}$ to $\pm 5.5\text{V}$	78	90		dB
A_{VOL}	Large-Signal Voltage Gain	$V_{OUT} = \pm 3\text{V}$, $R_L = 500\Omega$ $V_{OUT} = \pm 3\text{V}$, $R_L = 150\Omega$	10 7	17 14		V/mV V/mV
V_{OUT}	Output Swing	$R_L = 500\Omega$, $V_{IN} = \pm 10\text{mV}$ $R_L = 150\Omega$, $V_{IN} = \pm 10\text{mV}$	± 3.2 ± 3.1	± 3.8 ± 3.4		V V
I_{OUT}	Output Current	$V_{OUT} = \pm 3\text{V}$, 10mV Overdrive	23	50		mA
I_{SC}	Short-Circuit Current	$V_{OUT} = 0\text{V}$, $V_{IN} = \pm 1\text{V}$	35	90		mA
SR	Slew Rate	$A_V = -1$, (Note 7)	45	70		V/ μs
	Full Power Bandwidth	3V peak, (Note 8)		3.7		MHz
GBW	Gain Bandwidth	$f = 200\text{kHz}$	115	200		MHz
t_s	Settling Time	$A_V = -1$, 2V, 0.1% $A_V = -1$, 2V, 0.01%		91 112		ns ns
t_r , t_f	Rise Time, Fall Time	$A_V = 1$, 10% to 90%, $V_{IN} = 0.2V_{P-P}$, $R_L = 150\Omega$		6		ns
	Overshoot	$A_V = 1$, $V_{IN} = 0.2V_{P-P}$, $R_L = 150\Omega$, $R_F = 0\Omega$		15		%
	Propagation Delay	50% V_{IN} to 50% $V_{OUT} = 0.2V_{P-P}$, $R_L = 150\Omega$		3		ns
R_O	Output Resistance	$A_V = 1$, $f = 1\text{MHz}$		0.15		Ω
	Channel Separation	$V_{OUT} = \pm 3\text{V}$, $R_L = 150\Omega$	82	90		dB
I_S	Supply Current	Per Amplifier		3.7	4.5	mA

$T_A = 25^\circ\text{C}$. $V_S = 5\text{V}$, $V_{CM} = 2.5\text{V}$, R_L to 2.5V, unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
V_{OS}	Input Offset Voltage	(Note 6) LT1722 SOT-23 and LT1723 MS8		250 350	550 800	μV μV
I_{OS}	Input Offset Current			20	300	nA
I_B	Input Bias Current			20	300	nA
e_n	Input Noise Voltage	$f = 10\text{kHz}$		4		$\text{nV}/\sqrt{\text{Hz}}$
i_n	Input Noise Current	$f = 10\text{kHz}$		1.1		$\text{pA}/\sqrt{\text{Hz}}$
R_{IN}	Input Resistance	$V_{CM} = 1.5\text{V}$ to 3.5V Differential	5	32 55		$\text{M}\Omega$ $\text{k}\Omega$
C_{IN}	Input Capacitance			2		pF
	Input Voltage Range + Input Voltage Range –		3.5	4 1	1.5	V V
CMRR	Common Mode Rejection Ratio	$V_{CM} = 1.5\text{V}$ to 3.5V	80	100		dB
A_{VOL}	Large-Signal Voltage Gain	$V_{OUT} = 1.5\text{V}$ to 3.5V, $R_L = 500\Omega$	4	10		V/mV
V_{OUT}	Output Swing+ Output Swing–	$R_L = 500\Omega$, $V_{IN} = \pm 10\text{mV}$ $R_L = 500\Omega$, $V_{IN} = \pm 10\text{mV}$	3.6	3.8 0.9	1.4	V V

ELECTRICAL CHARACTERISTICS $T_A = 25^\circ\text{C}$. $V_S = 5\text{V}$, $V_{CM} = 2.5\text{V}$, R_L to 2.5V , unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
I_{OUT}	Output Current	$V_{OUT} = 3.5\text{V}$ or 1.5V , 10mV Overdrive	10	20		mA
I_{SC}	Short-Circuit Current	$V_{OUT} = 2.5\text{V}$, $V_{IN} = \pm 1\text{V}$	22	55		mA
SR	Slew Rate	$A_V = -1$, (Note 7)	40	70		V/ μs
	Full Power Bandwidth	1V peak, (Note 8)		8.7		MHz
GBW	Gain Bandwidth (Note 10)	$f = 200\text{kHz}$	115	180		MHz
t_r , t_f	Rise Time, Fall Time	$A_V = 1$, 10% to 90% , $V_{IN} = 0.2V_{P-P}$, $R_L = 500\Omega$		5		ns
	Overshoot	$A_V = 1$, $V_{IN} = 0.2V_{P-P}$, $R_L = 500\Omega$		16		%
	Propagation Delay	$50\% V_{IN}$ to $50\% V_{OUT}$, 0.1V , $R_L = 500\Omega$		3		ns
R_O	Output Resistance	$A_V = 1$, $f = 1\text{MHz}$		0.19		Ω
	Channel Separation	$V_{OUT} = 1.5\text{V}$ to 3.5V , $R_L = 500\Omega$	82	90		dB
I_S	Supply Current	Per Amplifier		3.8	5	mA

The ● denotes the specifications which apply over the temperature range of $0^\circ\text{C} \leq T_A \leq 70^\circ\text{C}$. $V_S = \pm 5\text{V}$, $V_{CM} = 0\text{V}$, unless otherwise noted. (Note 5)

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
V_{OS}	Input Offset Voltage	(Note 6) LT1722 SOT-23 and LT1723 MS8	●		700	μV
	Input V_{OS} Drift	(Note 9)	●	3	7	$\mu\text{V}/^\circ\text{C}$
I_{OS}	Input Offset Current		●		350	nA
I_B	Input Bias Current		●		350	nA
	Input Voltage Range + Input Voltage Range –		●	3.5	–3.5	V
CMRR	Common Mode Rejection Ratio	$V_{CM} = \pm 3.5\text{V}$	●	75		dB
PSRR	Power Supply Rejection Ratio	$V_S = \pm 2.3\text{V}$ to $\pm 5.5\text{V}$	●	76		dB
A_{VOL}	Large-Signal Voltage Gain	$V_{OUT} = \pm 3\text{V}$, $R_L = 500\Omega$ $V_{OUT} = \pm 3\text{V}$, $R_L = 150\Omega$	●	9		V/mV
			●	6		V/mV
V_{OUT}	Output Swing	$R_L = 500\Omega$, $V_{IN} = \pm 10\text{mV}$ $R_L = 150\Omega$, $V_{IN} = \pm 10\text{mV}$	●	± 3.15		V
			●	± 3.05		V
I_{OUT}	Output Current	$V_{OUT} = \pm 3\text{V}$, 10mV Overdrive	●	22		mA
I_{SC}	Short-Circuit Current	$V_{OUT} = 0\text{V}$, $V_{IN} = \pm 1\text{V}$	●	30		mA
SR	Slew Rate	$A_V = -1$, (Note 7)	●	35		V/ μs
GBW	Gain Bandwidth	$f = 200\text{kHz}$	●	100		MHz
	Channel Separation	$V_{OUT} = \pm 3\text{V}$, $R_L = 150\Omega$	●	81		dB
I_S	Supply Current	Per Amplifier	●		5.45	mA

ELECTRICAL CHARACTERISTICS

The ● denotes the specifications which apply over the temperature range of $0^{\circ}\text{C} \leq T_A \leq 70^{\circ}\text{C}$. $V_S = 5\text{V}$, $V_{CM} = 2.5\text{V}$, R_L to $2.5\text{k}\Omega$, unless otherwise noted. (Note 5)

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
V_{OS}	Input Offset Voltage	(Note 6) LT1722 SOT-23 and LT1723MS8	●		850	μV
	Input V_{OS} Drift	(Note 9)	●	3	7	$\mu\text{V}/^{\circ}\text{C}$
I_{OS}	Input Offset Current		●		350	nA
I_B	Input Bias Current		●		350	nA
	Input Voltage Range +		●	3.5		V
	Input Voltage Range –		●		1.5	V
CMRR	Common Mode Rejection Ratio	$V_{CM} = 1.5\text{V}$ to 3.5V	●	75		dB
A_{VOL}	Large-Signal Voltage Gain	$V_{OUT} = 1.5\text{V}$ to 3.5V , $R_L = 500\Omega$	●	3		V/mV
V_{OUT}	Output Swing+	$R_L = 500\Omega$, $V_{IN} = \pm 10\text{mV}$	●	3.55		V
	Output Swing–	$R_L = 500\Omega$, $V_{IN} = \pm 10\text{mV}$	●		1.45	V
I_{OUT}	Output Current	$V_{OUT} = 3.5\text{V}$ or 1.5V , 10mV Overdrive	●	9		mA
I_{SC}	Short-Circuit Current	$V_{OUT} = 2.5\text{V}$, $V_{IN} = \pm 1\text{V}$	●	11		mA
SR	Slew Rate	$A_V = -1$, (Note 7)	●	30		V/ μs
GBW	Gain Bandwidth (Note 10)	$f = 200\text{kHz}$	●	100		MHz
	Channel Separation	$V_{OUT} = 1.5\text{V}$ to 3.5V , $R_L = 500\Omega$	●	81		dB
I_S	Supply Current		●		5.95	mA

The ● denotes the specifications which apply over the temperature range of $-40^{\circ}\text{C} \leq T_A \leq 85^{\circ}\text{C}$. $V_S = \pm 5\text{V}$, $V_{CM} = 0\text{V}$, unless otherwise noted. (Note 5)

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
V_{OS}	Input Offset Voltage	(Note 6) LT1722 SOT-23 and LT1723 MS8	●		900	μV
	Input V_{OS} Drift	(Note 9)	●	3	10	$\mu\text{V}/^{\circ}\text{C}$
I_{OS}	Input Offset Current		●		400	nA
I_B	Input Bias Current		●		400	nA
	Input Voltage Range +		●	3.5		V
	Input Voltage Range –				–3.5	V
CMRR	Common Mode Rejection Ratio	$V_{CM} = \pm 3.5\text{V}$	●	75		dB
PSRR	Power Supply Rejection Ratio	$V_S = \pm 2.0\text{V}$ to $\pm 5.5\text{V}$	●	75		dB
A_{VOL}	Large-Signal Voltage Gain	$V_{OUT} = \pm 3\text{V}$, $R_L = 500\Omega$	●	8		V/mV
		$V_{OUT} = \pm 3\text{V}$, $R_L = 150\Omega$	●	5		V/mV
V_{OUT}	Output Swing	$R_L = 500\Omega$, $V_{IN} = \pm 10\text{mV}$	●	± 3.1		V
		$R_L = 150\Omega$, $V_{IN} = \pm 10\text{mV}$	●	± 3.0		V
I_{OUT}	Output Current	$V_{OUT} = \pm 3\text{V}$, 10mV Overdrive	●	20		mA
I_{SC}	Short-Circuit Current	$V_{OUT} = 0\text{V}$, $V_{IN} = \pm 1\text{V}$	●	25		mA
SR	Slew Rate	$A_V = -1$, (Note 7)	●	25		V/ μs
GBW	Gain Bandwidth	$f = 200\text{kHz}$	●	90		MHz
	Channel Separation	$V_{OUT} = \pm 3\text{V}$, $R_L = 150\Omega$	●	80		dB
I_S	Supply Current		●		5.95	mA

LT1722/LT1723/LT1724

ELECTRICAL CHARACTERISTICS

The ● denotes the specifications which apply over the temperature range of $-40^{\circ}\text{C} \leq T_A \leq 85^{\circ}\text{C}$. $V_S = 5\text{V}$, $V_{CM} = 2.5\text{V}$, R_L to 2.5V , unless otherwise noted. (Note 5)

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
V_{OS}	Input Offset Voltage	(Note 6) LT1722 SOT-23 and LT1723 MS8	●		1000	μV
	Input V_{OS} Drift	(Note 9)	●	3	10	$\mu\text{V}/^{\circ}\text{C}$
I_{OS}	Input Offset Current		●		400	nA
I_B	Input Bias Current		●		400	nA
	Input Voltage Range + Input Voltage Range –		● ●	3.5	1.5	V V
CMRR	Common Mode Rejection Ratio	$V_{CM} = 1.5\text{V}$ to 3.5V	●	75		dB
A_{VOL}	Large-Signal Voltage Gain	$V_{OUT} = 1.5\text{V}$ to 3.5V , $R_L = 500\Omega$	●	2		V/mV
V_{OUT}	Output Swing+ Output Swing–	$R_L = 500\Omega$, $V_{IN} = \pm 10\text{mV}$ $R_L = 500\Omega$, $V_{IN} = \pm 10\text{mV}$	● ●	3.5	1.5	V V
I_{OUT}	Output Current	$V_{OUT} = 3.5\text{V}$ or 1.5V , 30mV Overdrive	●	8		mA
I_{SC}	Short-Circuit Current	$V_{OUT} = 2.5\text{V}$, $V_{IN} = \pm 1\text{V}$	●	10		mA
SR	Slew Rate	$A_V = -1$, (Note 7)	●	20		V/ μs
GBW	Gain Bandwidth (Note 10)	$f = 200\text{kHz}$	●	90		MHz
	Channel Separation	$V_{OUT} = 1.5\text{V}$ to 3.5V , $R_L = 500\Omega$	●	80		dB
I_S	Supply Current		●		6.45	mA

Note 1: Absolute Maximum Ratings are those values beyond which the life of a device may be impaired.

Note 2: The inputs are protected by back-to-back diodes. If the differential input voltage exceeds 0.7V, the input current should be limited to less than 10mA.

Note 3: A heat sink may be required to keep the junction temperature below the absolute maximum rating when the output is shorted indefinitely.

Note 4: The LT1722C/LT1722I, LT1723C/LT1723I, LT1724C/LT1724I are guaranteed functional over the operating temperature range of -40°C to 85°C .

Note 5: The LT1722C/LT1723C/LT1724C are guaranteed to meet specified performance from 0°C to 70°C . The LT1722C/LT1723C/LT1724C are

designed, characterized and expected to meet specified performance from -40°C to 85°C but are not tested or QA sampled at these temperatures. The LT1722I/LT1723I/LT1724I are guaranteed to meet specified performance from -40°C to 85°C .

Note 6: Input offset voltage is pulse tested and is exclusive of warm-up drift.

Note 7: Slew rate is measured between $\pm 2\text{V}$ on the output with $\pm 3\text{V}$ input for $\pm 5\text{V}$ supplies and $\pm 1\text{V}$ on the output with $\pm 1.5\text{V}$ input for single 5V supply. (For 5V supply, the voltage levels are 2.5V referred.)

Note 8: Full power bandwidth is calculated from the slew rate:

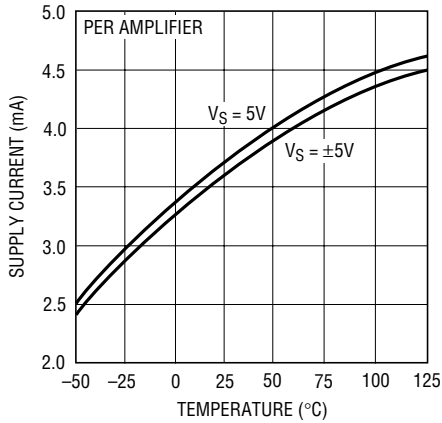
$$\text{FPBW} = \text{SR} / 2\pi V_P$$

Note 9: This parameter is not 100% tested.

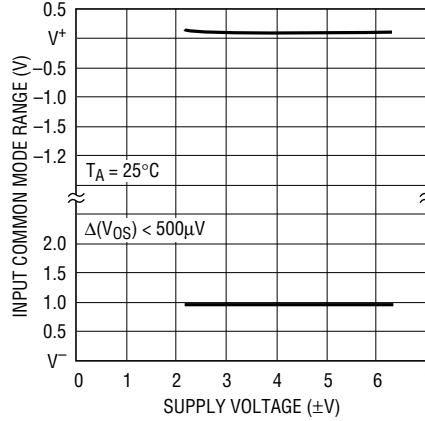
Note 10: This parameter is guaranteed through correlation with slew rate.

TYPICAL PERFORMANCE CHARACTERISTICS

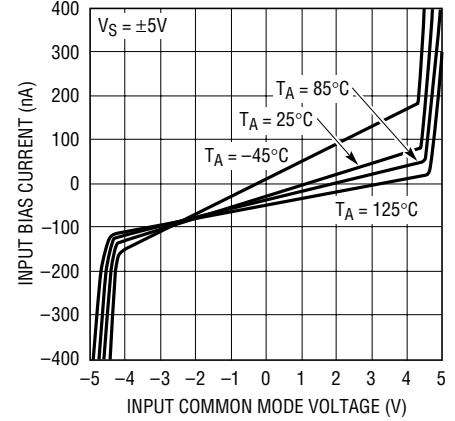
Supply Current vs Temperature



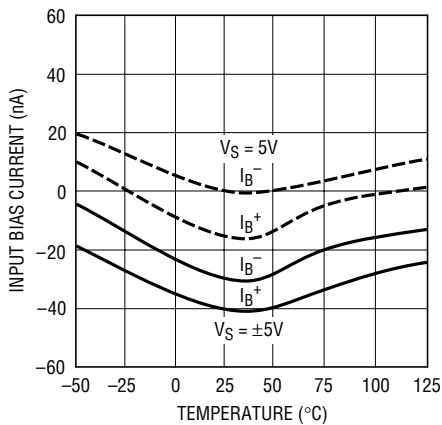
Input Common Mode Range vs Supply Voltage



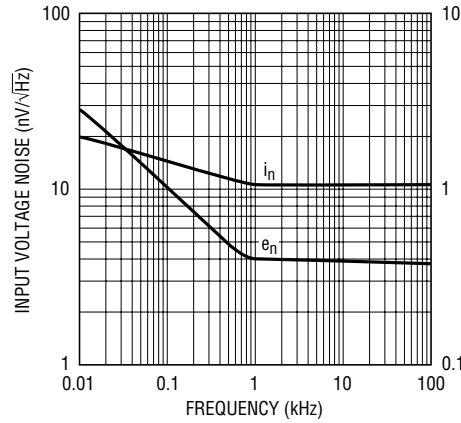
Input Bias Current vs Common Mode Voltage



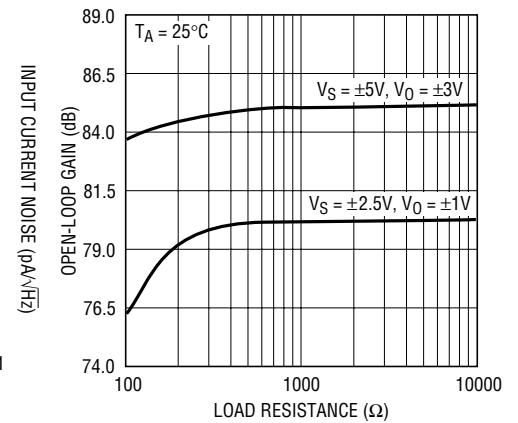
Input Bias Current vs Temperature



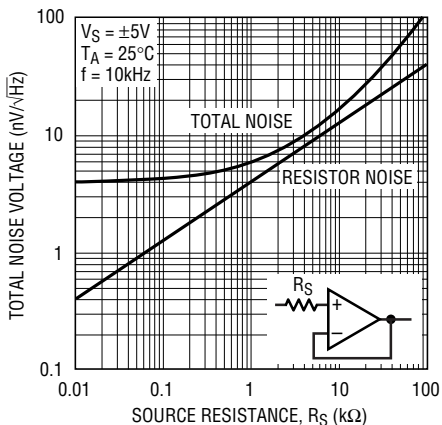
Input Noise Spectral Density



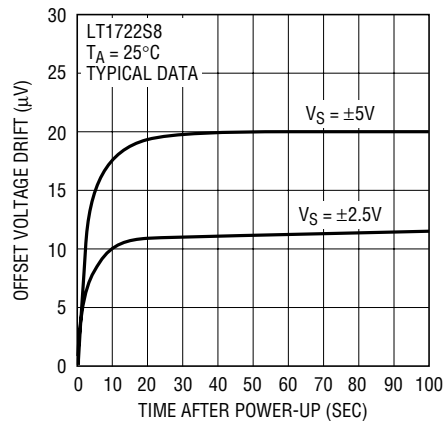
Open-Loop Gain vs Resistive Load



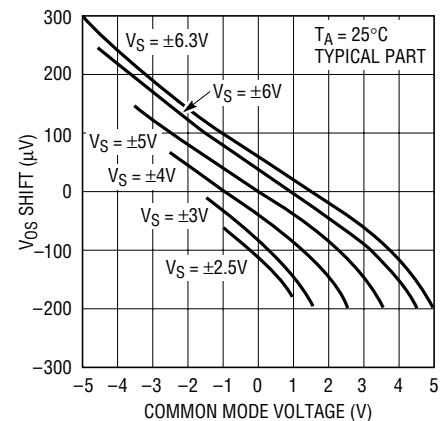
Total Noise vs Unmatched Source Resistance



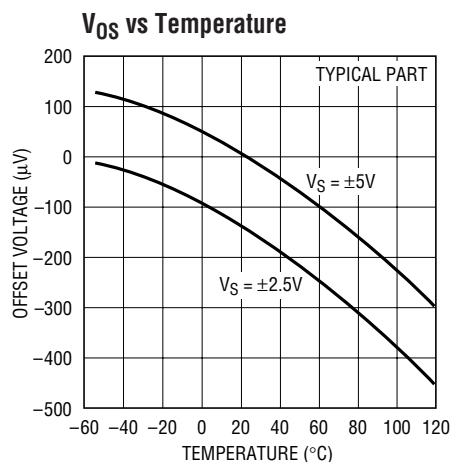
Warm-Up Drift vs Time



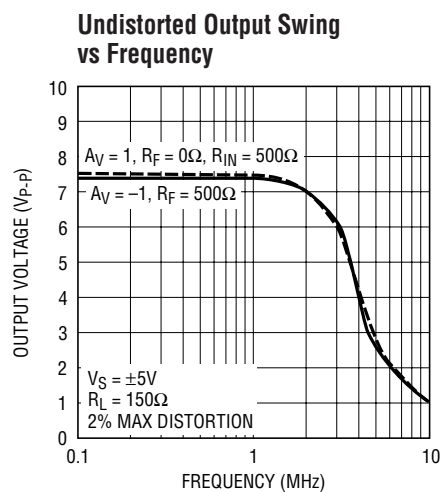
V_OS Shift vs V_CM and V_S



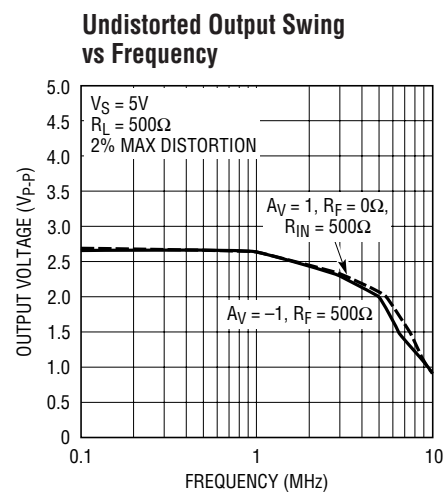
TYPICAL PERFORMANCE CHARACTERISTICS



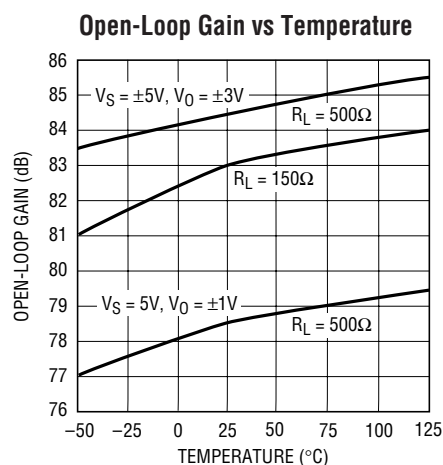
1723 G10



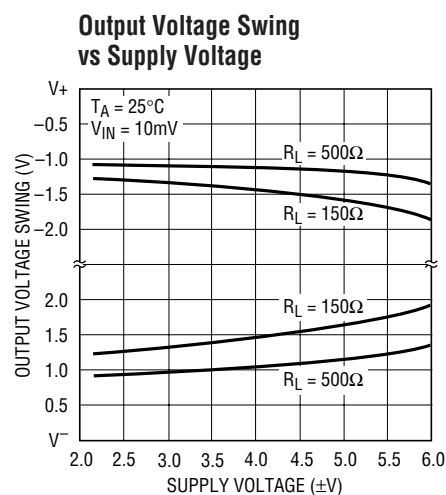
1723 G11



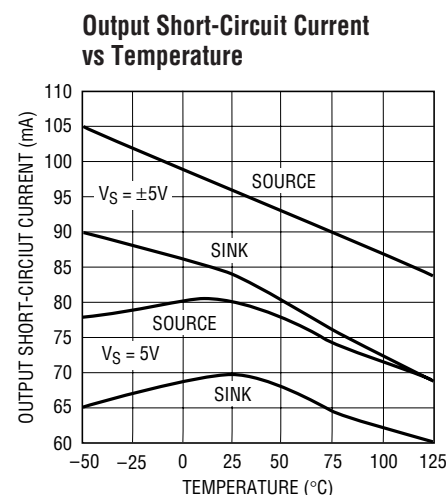
1723 G12



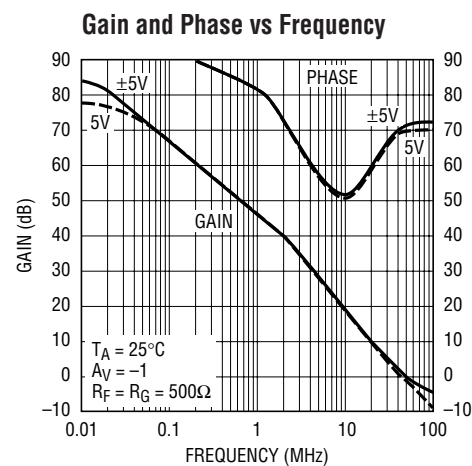
1723 G13



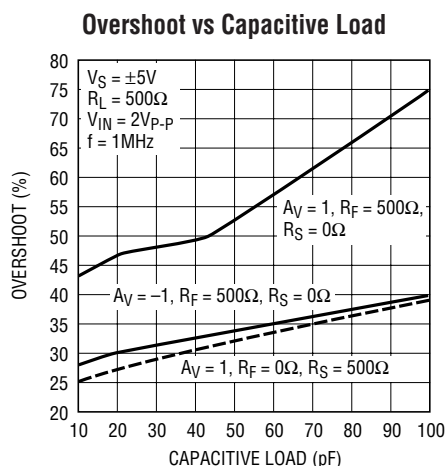
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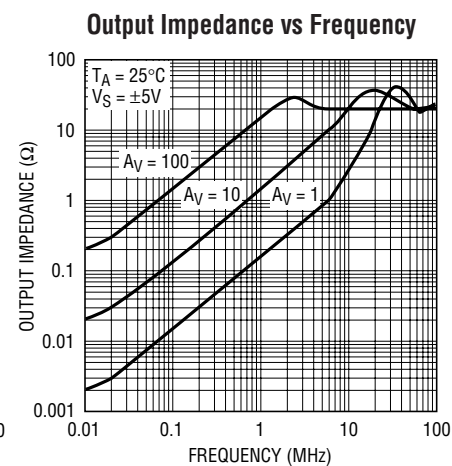
1723 G15



1723 G16



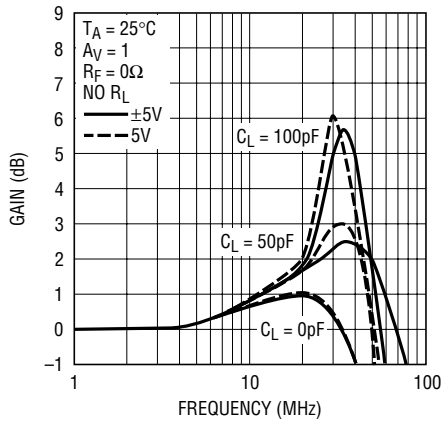
1723 G17



1723 G18

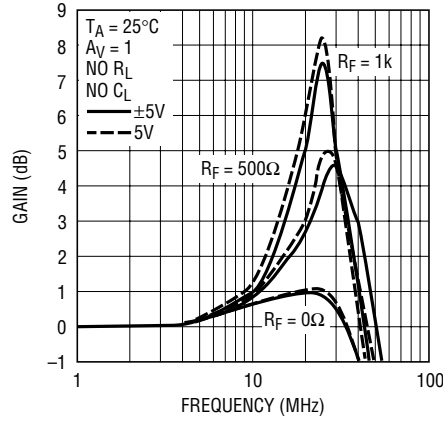
TYPICAL PERFORMANCE CHARACTERISTICS

Gain vs Frequency, $A_V = 1$



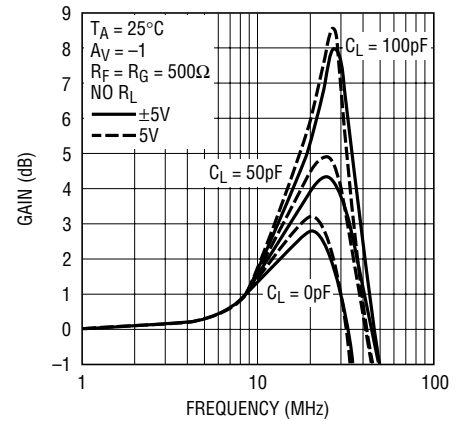
1723 G19

Gain vs Frequency, $A_V = 1$



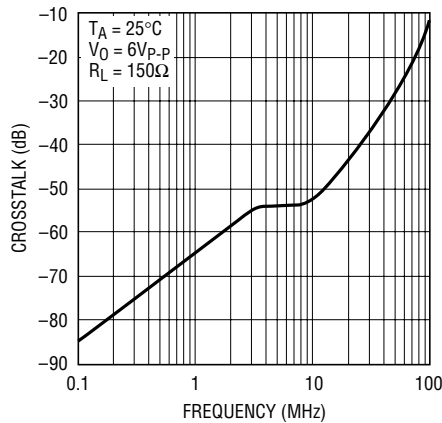
1723 G20

Gain vs Frequency, $A_V = -1$



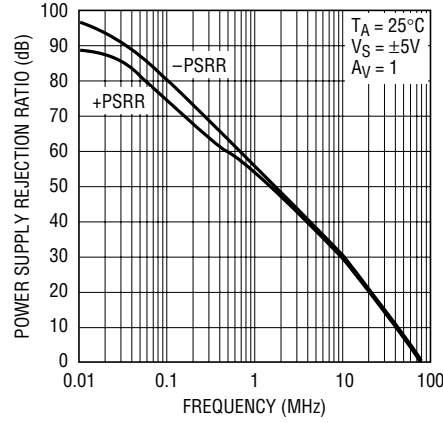
1723 G21

Channel Separation vs Frequency



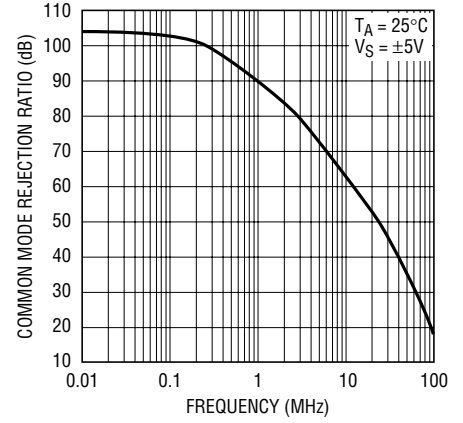
1723 G22

Power Supply Rejection Ratio vs Frequency



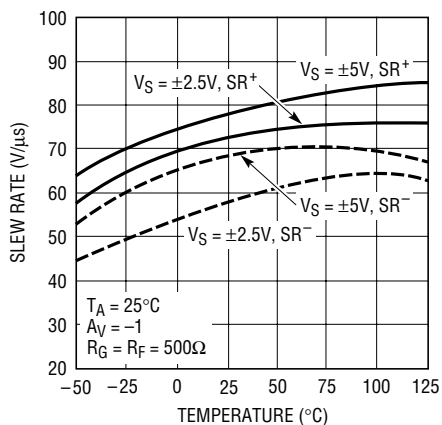
1723 G23

Common Mode Rejection Ratio vs Frequency



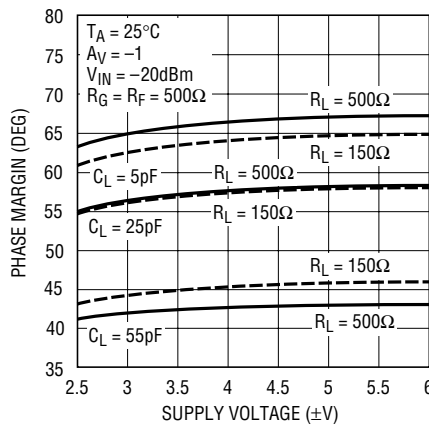
1723 G24

Slew Rate vs Temperature



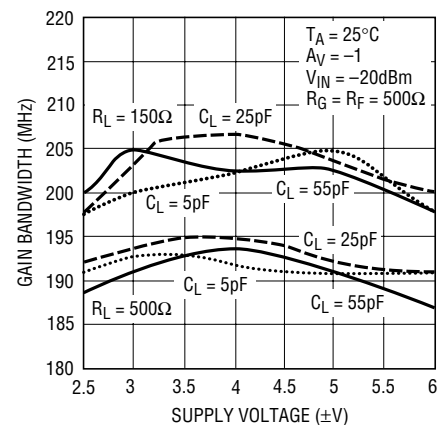
1723 G40

Phase Margin vs Supply Voltage



1723 G41

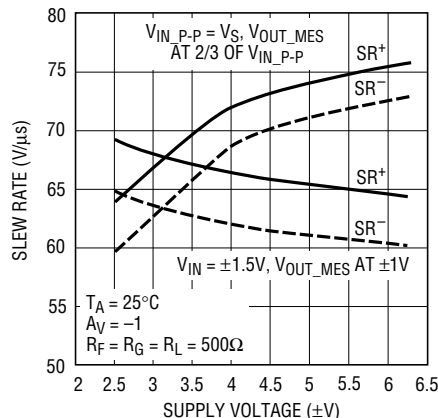
Gain Bandwidth vs Supply Voltage



1723 G42

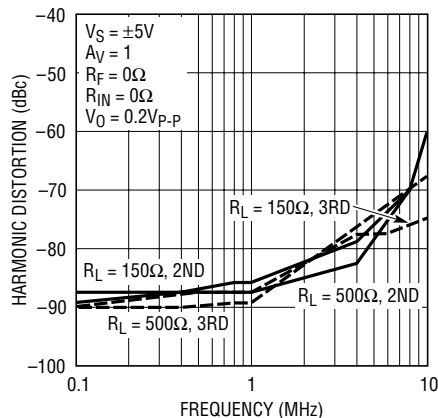
TYPICAL PERFORMANCE CHARACTERISTICS

Slew Rate vs Supply Voltage



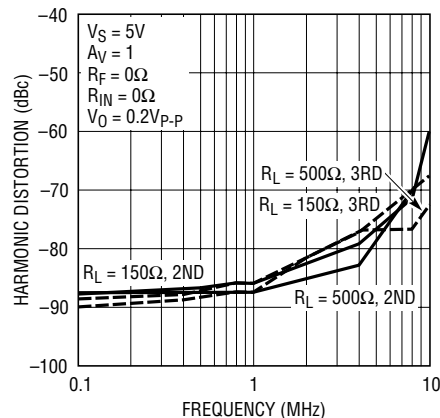
1723 G25

Harmonic Distortion vs Frequency
 $A_V = 1, V_0 = 0.2V_{P-P}$



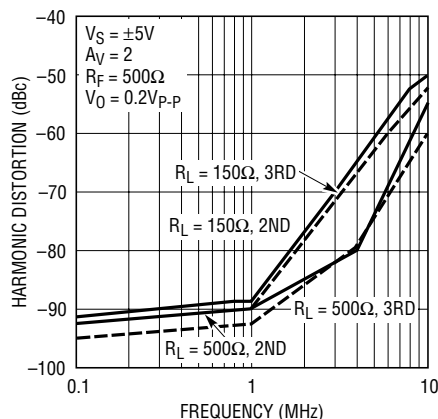
1723 G26

Harmonic Distortion vs Frequency
 $A_V = 1, V_0 = 0.2V_{P-P}$



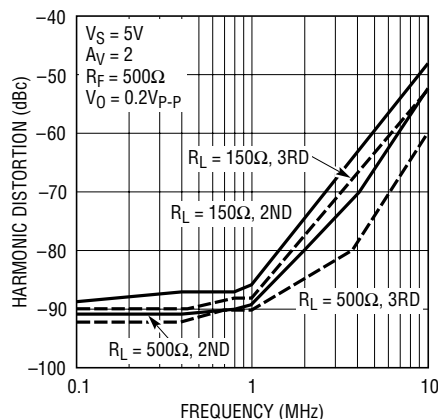
1723 G27

Harmonic Distortion vs Frequency
 $A_V = 2, V_0 = 0.2V_{P-P}$



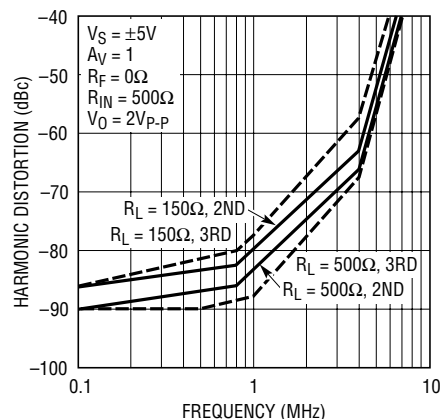
1723 G28

Harmonic Distortion vs Frequency
 $A_V = 2, V_0 = 0.2V_{P-P}$



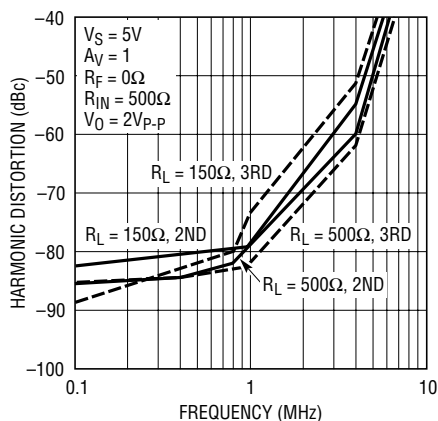
1723 G29

Harmonic Distortion vs Frequency
 $A_V = 1, V_0 = 2V_{P-P}$



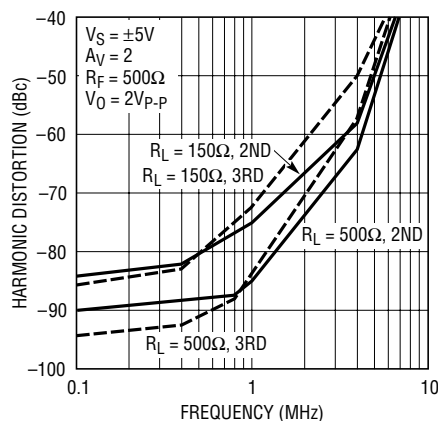
1723 G30

Harmonic Distortion vs Frequency
 $A_V = 1, V_0 = 2V_{P-P}$



1723 G31

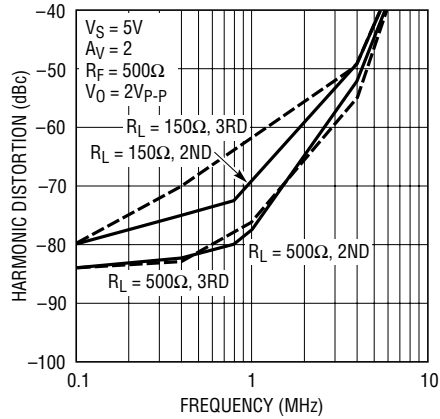
Harmonic Distortion vs Frequency
 $A_V = 2, V_0 = 2V_{P-P}$



1723 G32

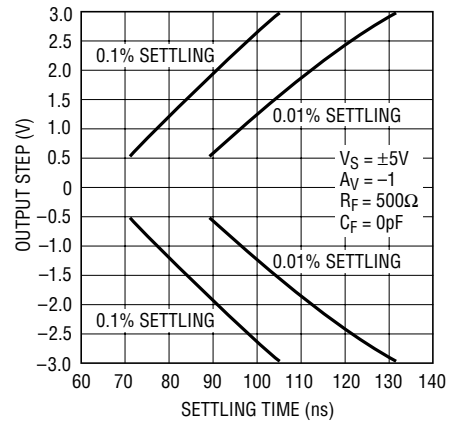
TYPICAL PERFORMANCE CHARACTERISTICS

Harmonic Distortion vs Frequency
 $A_V = 2$, $V_O = 2V_{P-P}$



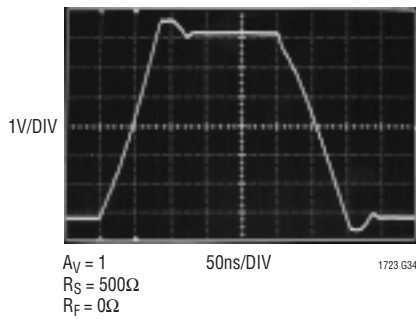
1723 G33

Settling Time vs Output Step



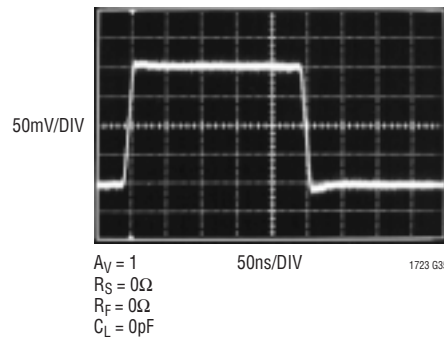
1723 G43

Large-Signal Transient, $A_V = 1$



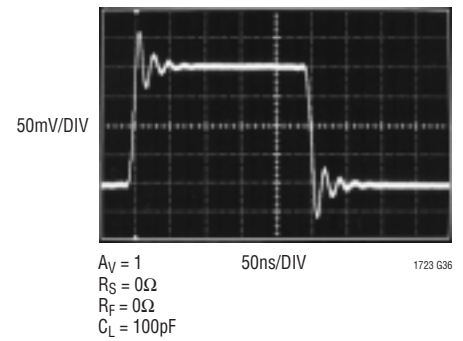
1723 G34

Small-Signal Transient, $A_V = 1$



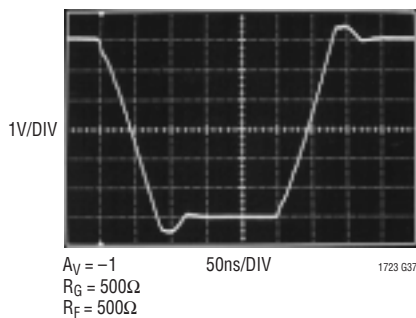
1723 G35

Small-Signal Transient, $A_V = 1$



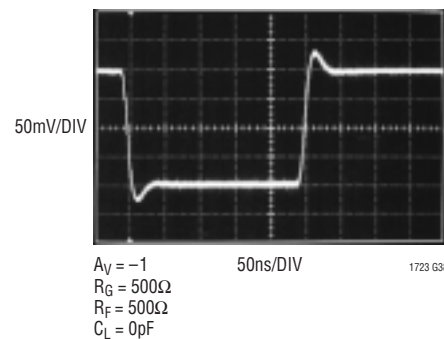
1723 G36

Large-Signal Transient, $A_V = -1$



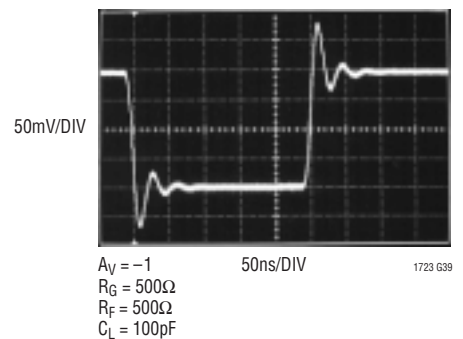
1723 G37

Small-Signal Transient, $A_V = -1$



1723 G38

Small-Signal Transient, $A_V = -1$



1723 G39

APPLICATIONS INFORMATION

The LT1722/LT1723/LT1724 may be inserted directly into many operational amplifier applications improving both DC and AC performance, as well as noise and distortion.

Layout and Passive Components

The LT1722/LT1723/LT1724 amplifiers are more tolerant of less than ideal layouts than other high speed amplifiers. For maximum performance (for example, fast settling time) use a ground plane, short lead lengths and RF quality bypass capacitors (0.01 μ F to 0.1 μ F). For high drive current applications, use low ESR supply bypass capacitors (1 μ F to 10 μ F tantalum). The output/input parasitic coupling should be minimized when high frequency performance is required.

The parallel combination of the feedback resistor and gain setting resistor on the inverting input combine with the input capacitance to form a pole that can cause peaking or even oscillations. In parallel with the feedback resistor, a capacitor of value:

$$C_F > R_G \cdot C_{IN}/R_F$$

should be used to cancel the input pole and optimize dynamic performance. For unity-gain applications where a feedback resistor is used, such as an I-to-V converter, C_F should be five times greater than C_{IN} ; an optimum value for C_F is 10pF.

Input Considerations

Each of the LT1722/LT1723/LT1724 inputs is protected with back-to-back diodes across the bases of the NPN input devices. If greater than 0.7V differential input voltages are anticipated, the input current must be limited to less than 10mA with an external series resistor. Each input also has two ESD clamp diodes—one to each supply. If an input is driven beyond the supply, limit the current with an external resistor to less than 10mA. The input stage protection circuit is shown in Figure 1.

The input currents of the LT1722/LT1723/LT1724 are typically in the tens of nA range due to the bias current cancellation technique used at the input. As the input offset current can be greater than either input current,

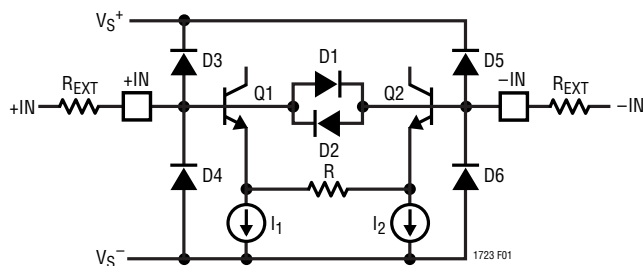


Figure 1. Input Stage Protection

adding resistance to balance source resistance is not recommended. The value of the source resistor should be below 12k as it actually degrades DC accuracy and also increases noise.

Total Input Noise

The total input noise of the LT1722/LT1723/LT1724 is optimized for a source resistance between 0.8k and 12k. Within this range, the total input noise is dominated by the noise of the source resistance itself. When the source resistance is below 0.8k, voltage noise of the amplifier dominates. When the source resistance is above 12k, the input noise current is the dominant contributor.

Capacitive Loading

The LT1722/LT1723/LT1724 drive capacitive loads up to 100pF with unity gain. As the capacitive load increases, both the bandwidth and the phase margin decrease causing peaking in the frequency response and overshoot in the transient response. When there is a need to drive a larger capacitive load, a 25 Ω series resistance assures stability with any value of load capacitor. A feedback capacitor also helps to reduce any peaking.

Power Dissipation

The LT1722/LT1723/LT1724 combine high speed and large output drive in a small package. Maximum junction temperature (T_J) is calculated from the ambient temperature (T_A), power dissipation per amplifier (P_D) and number of amplifiers (n) as follows:

$$T_J = T_A + (n \cdot P_D \cdot \theta_{JA})$$

APPLICATIONS INFORMATION

Power dissipation is composed of two parts. The first is due to the quiescent supply current and the second is due to on-chip dissipation caused by the load current.

Worst-case instantaneous power dissipation for a given resistive load in one amplifier occurs at the maximum supply current and when the output voltage is at half of either supply voltage (or the maximum swing if less than half supply voltage).

Therefore $P_{D(MAX)}$ in one amplifier is:

$$P_{D(MAX)} = (V^+ - V^-)(I_{S(MAX)}) + (V^+/2)^2/R_L$$

or

$$P_{D(MAX)} = (V^+ - V^-)(I_{S(MAX)}) + (V^+ - V_{O(MAX)})(V_{O(MAX)})/R_L$$

Example. Worst-case conditions are: both op amps in the LT1723IS8 are at $T_A = 85^\circ\text{C}$, $V_S = \pm 5\text{V}$, $R_L = 150\Omega$, $V_{OUT} = 2.5\text{V}$.

$$P_{D(MAX)} = 2 \cdot [(10\text{V})(5.95\text{mA}) + (2.5\text{V})^2/150\Omega] = 203\text{mW}$$

$$T_{J(MAX)} = 85^\circ\text{C} + (203\text{mW})(190^\circ\text{C/W}) = 124^\circ\text{C}$$

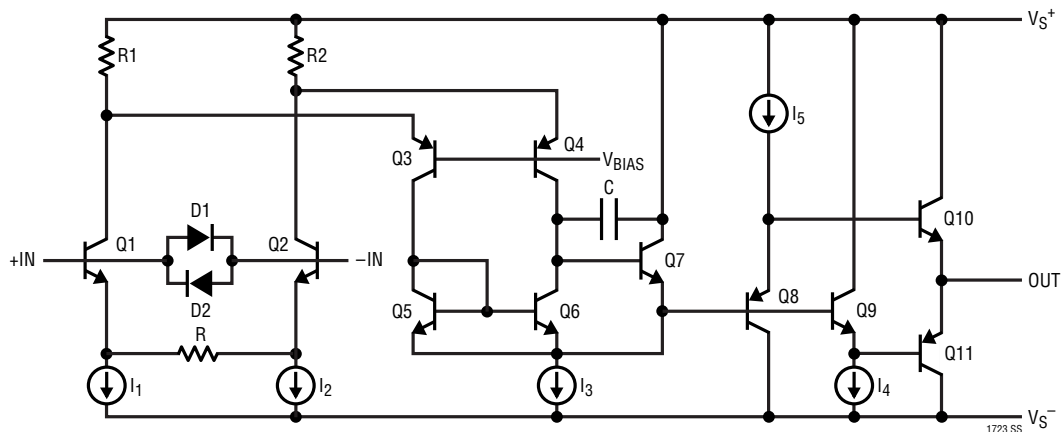
which is less than the absolute maximum rating at 150°C .

Circuit Operation

The LT1722/LT1723/LT1724 circuit topology is a voltage feedback amplifier. The operation of the circuit can be understood by referring to the Simplified Schematic. The first stage is a folded cascode formed by the transistors Q1 through Q4. A degeneration resistor, R, is used in the input stage. The current mirror Q5, Q6 is bootstrapped by Q7. The capacitor, C, assures the bandwidth and the slew rate performance. The output stage is formed by complementary emitter followers, Q8 through Q11. The diodes D1 and D2 protect against input reversed biasing. The remaining part of the circuit assures optimum voltage and current biases for all stages.

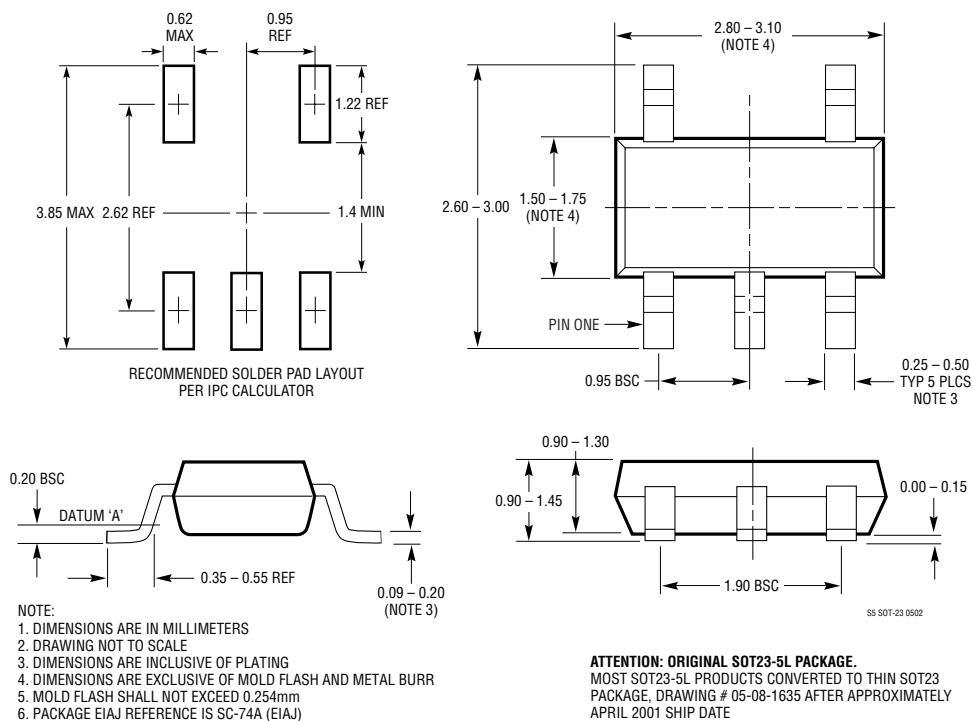
Low noise, reduced current supply, high speed and DC accurate parameters are distinctive features of the LT1722/LT1723/LT1724.

SIMPLIFIED SCHEMATIC

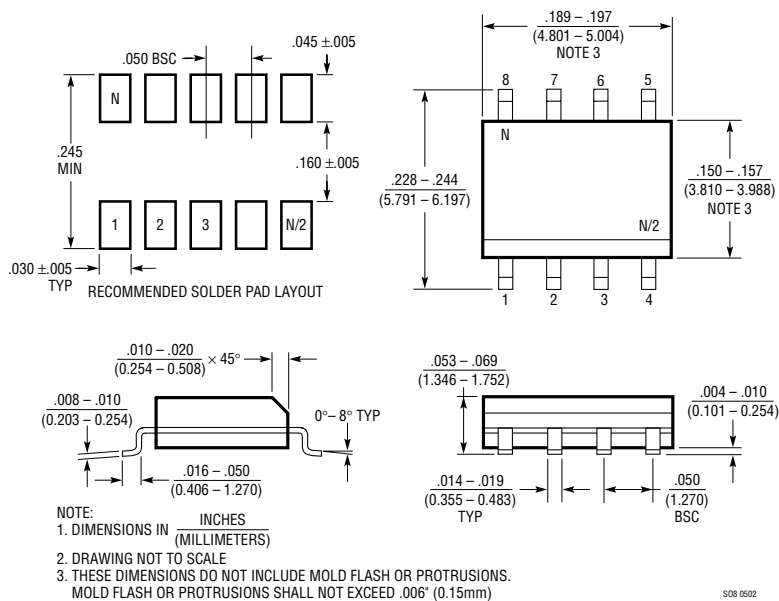


PACKAGE DESCRIPTION

S5 Package 5-Lead Plastic SOT-23 (Reference LTC DWG # 05-08-1633)

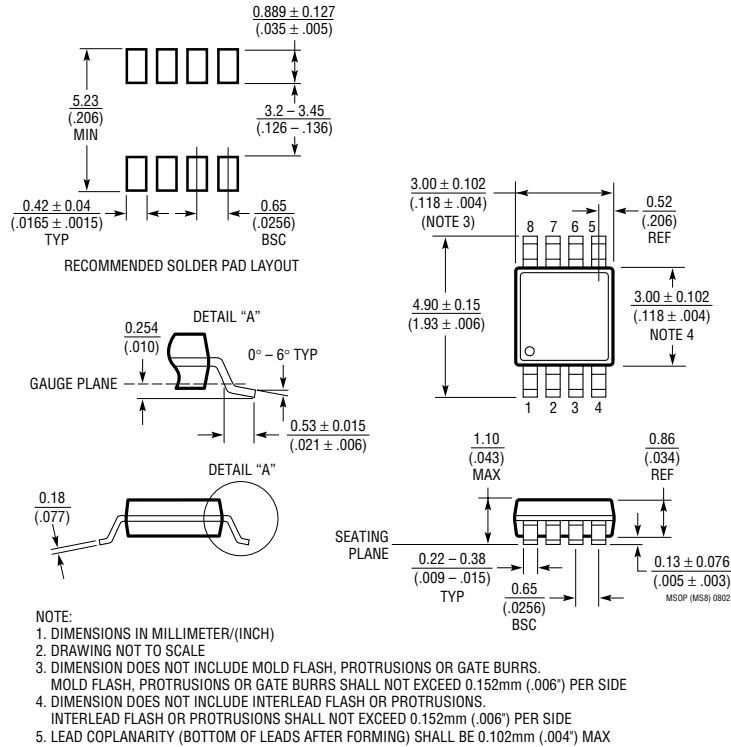


S8 Package 8-Lead Plastic Small Outline (Narrow .150 Inch) (Reference LTC DWG # 05-08-1610)

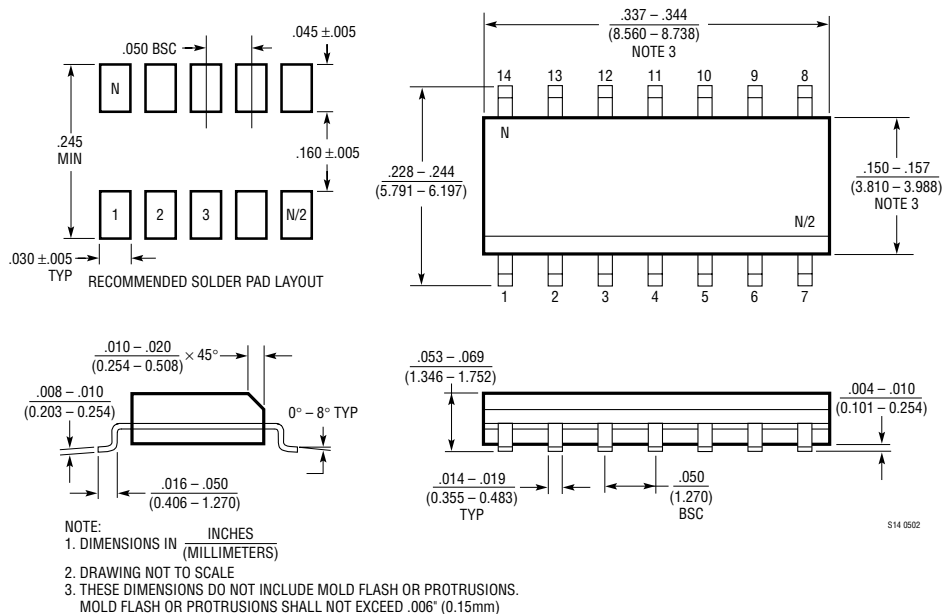


PACKAGE DESCRIPTION

MS8 Package 8-Lead Plastic MSOP (Reference LTC DWG # 05-08-1660)

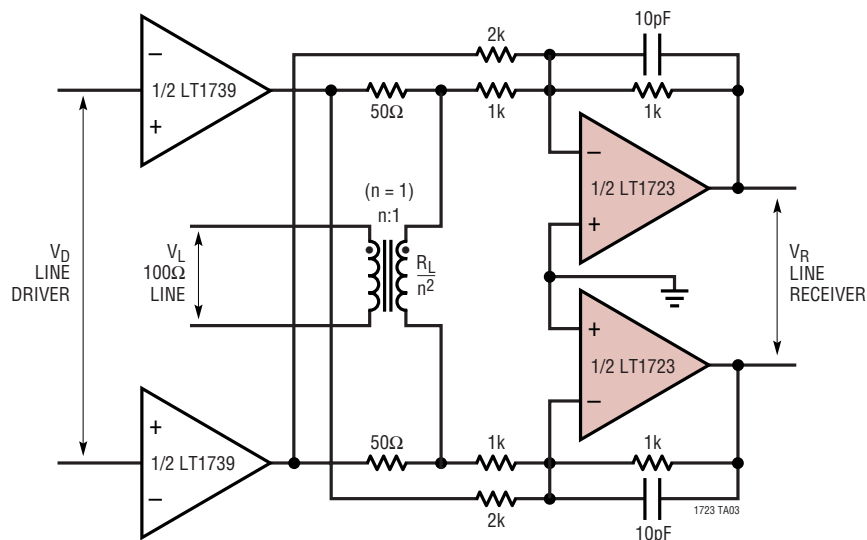


S Package 14-Lead Plastic Small Outline (Narrow .150 Inch) (Reference LTC DWG # 05-08-1610)



TYPICAL APPLICATION

4- to 2-Wire Local Echo Cancellation Differential Receiver Amplifier



RELATED PARTS

PART NUMBER	DESCRIPTION	COMMENTS
LT1677	Single, Low Noise Rail-to-Rail Amplifier	3V Operation, 2.5mA Supply Current, 4.5nV/√Hz Max e_n , 60μV Max V_{OS}
LT1800/LT1801/LT1802	Single/Dual/Quad, Low Power, 80MHz Rail-to-Rail Precision Amplifier	1.6mA Supply Current, 350μV V_{OS} , 2.3V Operation
LT1806/LT1807	Single/Dual, Low Noise 325MHz Rail-to-Rail Amplifiers	2.5V Operation, 550μV _{MAX} V_{OS} , 3.5nV/√Hz
LT1809/LT1810	Single/Dual, Low Distortion 180MHz Rail-to-Rail Amplifiers	2.5V Operation, -90dBc at 5MHz Distortion
LT1812/LT1813/LT1814	Single/Dual/Quad, 3mA, 750V/μs Amplifiers	5V Operation, 3.6mA Supply Current, 40mA Min Output Current
LT6202/LT6203/LT6204	Single/Dual/Quad, 100MHz, Low Noise Rail-to-Rail Op Amp	2nV/√Hz, 2.5mA on Single 3V Supply