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**SOT23-6W 120mA 2ch LDO  
REGULATORS  
R5322N SERIES**

**APPLICATION MANUAL**



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**SOT23-6W 120mA 2ch LDO REGULATORS**

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## R5322N SERIES

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### OUTLINE

The R5322N Series are voltage regulator ICs with high output voltage accuracy, low supply current, low dropout, and high ripple rejection by CMOS process. Each of these voltage regulator ICs consists of a voltage reference unit, an error amplifier, resistors for setting Output Voltage, a current limit circuit, and a chip enable circuit.

These ICs perform with low dropout voltage due to built-in transistor with low ON resistance, and a chip enable function and prolong the battery life of each system. The line transient response and load transient response of the R5322N Series are excellent, thus these ICs are very suitable for the power supply for hand-held communication equipment.

The output voltage of these ICs is internally fixed with high accuracy. Since the package for these ICs is SOT-23-6W package, and include 2ch LDO regulators each, high density mounting of the ICs on boards is possible.

### FEATURES

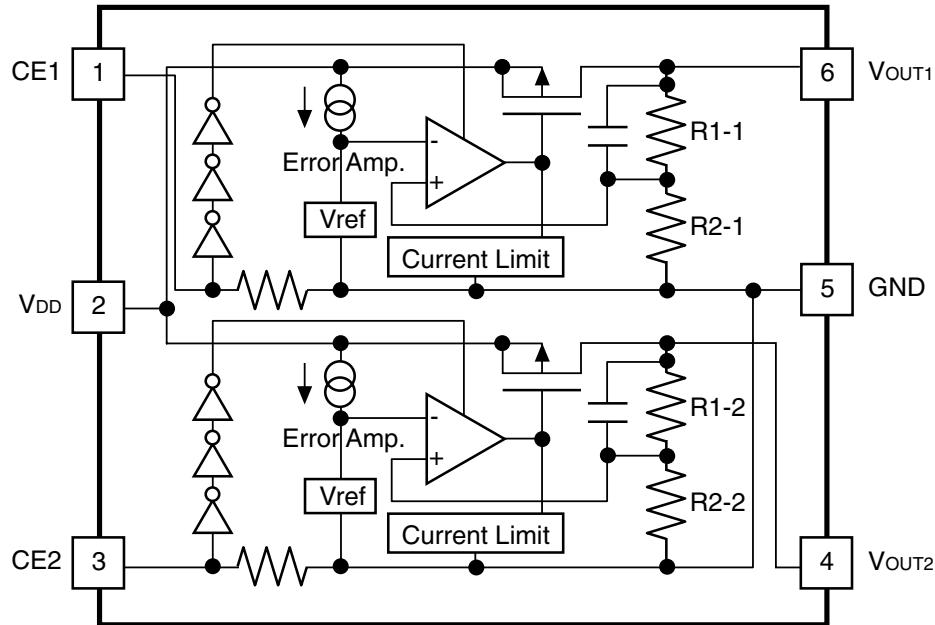
- Ultra-Low Supply Current ..... TYP. 75 $\mu$ A (VR1,VR2)
- Standby Mode ..... TYP. 0.1 $\mu$ A (VR1,VR2)
- Low Dropout Voltage ..... TYP. 0.15V (I<sub>OUT</sub>=100mA Output Voltage=3.0V Type)
- High Ripple Rejection ..... TYP. 75dB (f=1kHz)
- Low Temperature-Drift Coefficient of Output Voltage..... TYP.  $\pm$ 100ppm/ $^{\circ}$ C
- Excellent Line Regulation..... TYP. 0.05%/V
- High Output Voltage Accuracy .....  $\pm$ 2.0%
- Small Package ..... SOT-23-6W
- Output Voltage ..... Stepwise setting with a step of 0.1V in the range of 1.5V to 4.0V is possible
- Built-in chip enable circuit (A/B: active high)
- Built-in fold-back protection circuit..... TYP. 40mA (Current at short mode)

### APPLICATIONS

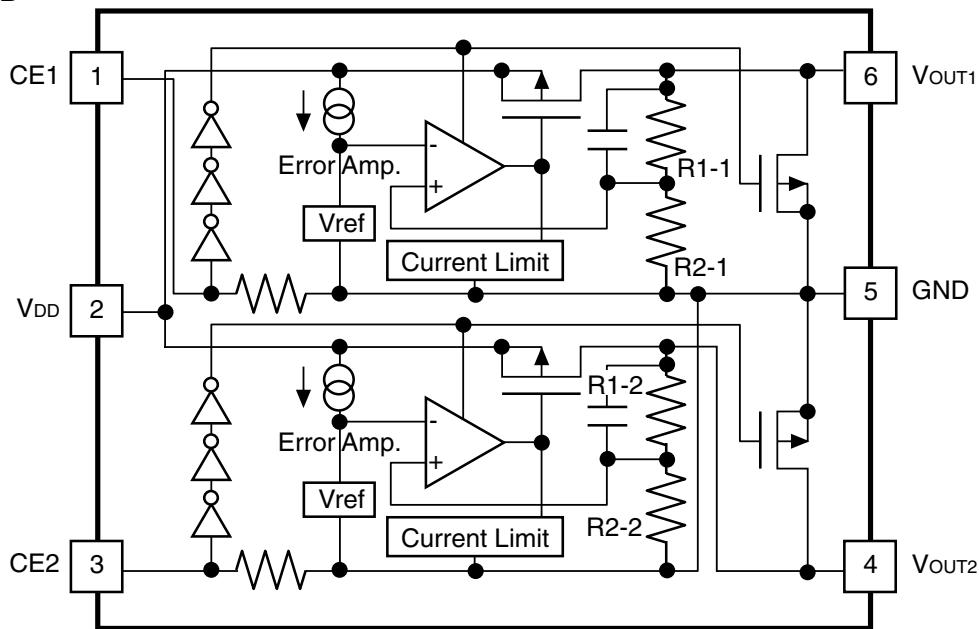
- Power source for cellular phones such as GSM, CDMA and various kinds of PCS.
- Power source for electrical appliances such as cameras, VCRs and camcorders.
- Power source for battery-powered equipment.

## BLOCK DIAGRAM

- R5322NxxxA



- R5322NxxxB



## SELECTION GUIDE

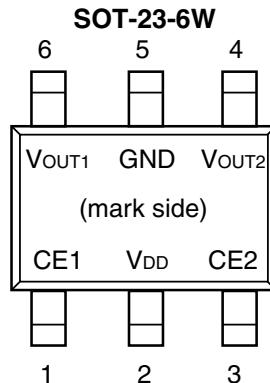
The output voltage, mask option, and the taping type for the ICs can be selected at the user's request. The selection can be made with designating the part number as shown below;

R5322Nxxxxx-xx ←Part Number

↑ ↑ ↑  
a b c

Code	Contents
a	Setting combination of 2ch Output Voltage ( $V_{OUT}$ ) : Serial Number for Voltage Setting, Stepwise setting with a step of 0.1V in the range of 1.5V to 4.0V is possible for each channel.
b	Designation of Mask Option : A version: without auto discharge function at OFF state. B version: with auto discharge function at OFF state.
c	Designation of Taping Type : Ex. TR (refer to Taping Specifications; TR type is the standard direction.)

## PIN CONFIGURATION



## PIN DESCRIPTION

Pin No.	Symbol	Description
1	CE1	Chip Enable Pin 1
2	V <sub>DD</sub>	Input Pin
3	CE2	Chip Enable Pin 2
4	V <sub>OUT2</sub>	Output Pin 2
5	GND	Ground Pin
6	V <sub>OUT1</sub>	Output Pin 1

**ABSOLUTE MAXIMUM RATINGS**

<b>Symbol</b>	<b>Item</b>	<b>Rating</b>	<b>Unit</b>
$V_{IN}$	Input Voltage	6.5	V
$V_{CE}$	Input Voltage (CE Pin)	$-0.3 \sim V_{IN}+0.3$	V
$V_{OUT}$	Output Voltage	$-0.3 \sim V_{IN}+0.3$	V
$I_{OUT1}$	Output Current 1	130	mA
$I_{OUT2}$	Output Current 2	130	mA
$P_D$	Power Dissipation	250	mW
$T_{opt}$	Operating Temperature Range	$-40 \sim 85$	°C
$T_{stg}$	Storage Temperature Range	$-55 \sim 125$	°C

## ELECTRICAL CHARACTERISTICS

- R5322NxxxA/B

Topt=25°C

Symbol	Item	Conditions	MIN.	TYP.	MAX.	Unit
V <sub>OUT</sub>	Output Voltage	V <sub>IN</sub> = Set V <sub>OUT</sub> +1V 1mA ≤ I <sub>OUT</sub> ≤ 30mA	V <sub>OUT</sub> ×0.98		V <sub>OUT</sub> ×1.02	V
I <sub>OUT</sub>	Output Current	V <sub>IN</sub> – V <sub>OUT</sub> = 1.0V	120			mA
ΔV <sub>OUT</sub> /ΔI <sub>OUT</sub>	Load Regulation	V <sub>IN</sub> = Set V <sub>OUT</sub> +1V 1mA ≤ I <sub>OUT</sub> ≤ 120mA		12	40	mV
V <sub>DIF</sub>	Dropout Voltage	Refer to the ELECTRICAL CHARACTERISTICS by OUTPUT VOLTAGE				
I <sub>SS</sub>	Supply Current	V <sub>IN</sub> = Set V <sub>OUT</sub> +1V		75	150	μA
I <sub>Standby</sub>	Supply Current (Standby)	V <sub>IN</sub> = V <sub>CE</sub> = Set V <sub>OUT</sub> +1V		0.1	1.0	μA
ΔV <sub>OUT</sub> /ΔV <sub>IN</sub>	Line Regulation	Set V <sub>OUT</sub> +0.5V ≤ V <sub>IN</sub> ≤ 6V I <sub>OUT</sub> = 30mA (In case that V <sub>OUT</sub> ≤ 1.6V, 2.2V ≤ V <sub>IN</sub> ≤ 6V)		0.05	0.20	%/V
R <sub>R</sub>	Ripple Rejection	f = 1kHz, Ripple 0.5Vp-p V <sub>IN</sub> = Set V <sub>OUT</sub> +1V, I <sub>OUT</sub> = 30mA		75		dB
V <sub>IN</sub>	Input Voltage		2.2		6.0	V
ΔV <sub>OUT</sub> /ΔT	Output Voltage Temperature Coefficient	I <sub>OUT</sub> = 30mA -40°C ≤ Topt ≤ 85°C		±100		ppm /°C
I <sub>lim</sub>	Short Current Limit	V <sub>OUT</sub> = 0V		40		mA
R <sub>PD</sub>	CE Pull-down Resistance		1.5	4	16	MΩ
V <sub>CEH</sub>	CE Input Voltage "H"		1.5		V <sub>IN</sub>	V
V <sub>CEL</sub>	CE Input Voltage "L"		0.0		0.3	V
en	Output Noise	BW=10Hz to 100kHz		30		μVrms
R <sub>LOW</sub>	Low Output Nch Tr. ON Resistance (of B version)	V <sub>CE</sub> =0V		70		Ω

- ELECTRICAL CHARACTERISTICS by OUTPUT VOLTAGE

Topt = 25°C

Output Voltage V <sub>OUT</sub> (V)	Dropout Voltage		
	V <sub>DIF</sub> (V)		
	Condition	TYP.	MAX.
1.5 ≤ V <sub>OUT</sub> ≤ 1.6	I <sub>OUT</sub> = 120mA	0.36	0.70
1.7 ≤ V <sub>OUT</sub> ≤ 1.8		0.30	0.50
1.9 ≤ V <sub>OUT</sub> ≤ 2.0		0.28	0.45
2.1 ≤ V <sub>OUT</sub> ≤ 2.7		0.24	0.40
2.8 ≤ V <sub>OUT</sub> ≤ 4.0		0.18	0.30

## TEST CIRCUITS

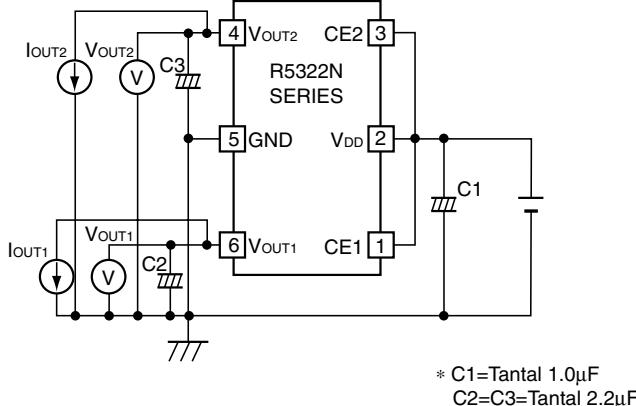


Fig.1 Standard test Circuit

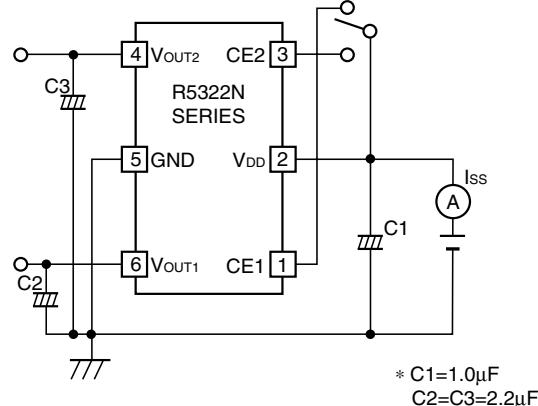


Fig.2 Supply Current Test Circuit

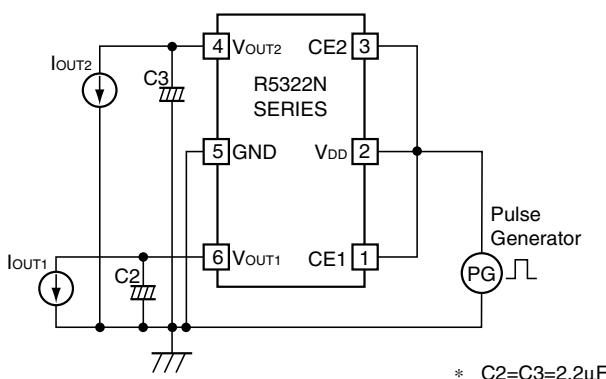
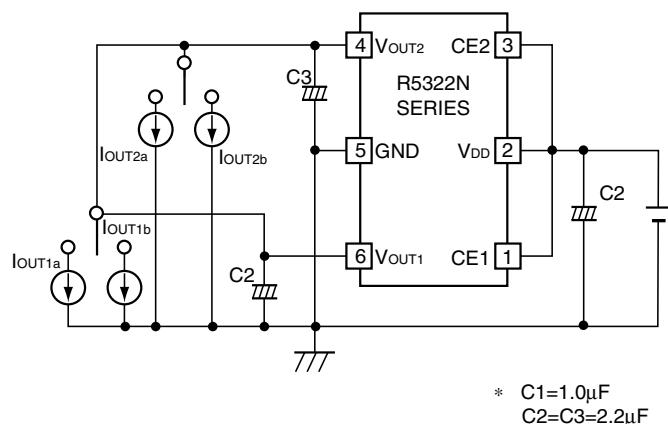
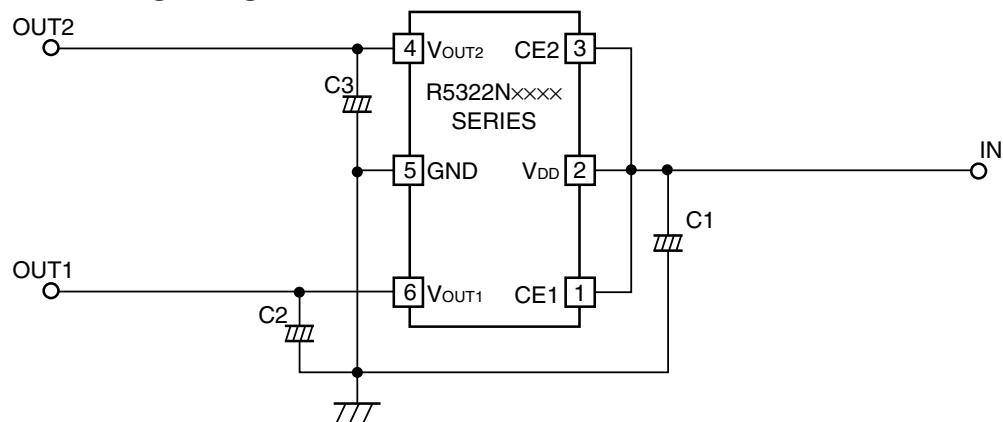
Fig.3 Ripple Rejection, Line Transient Response  
Test Circuit

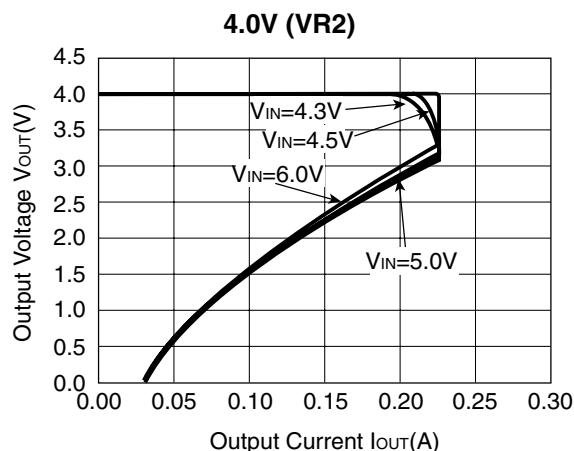
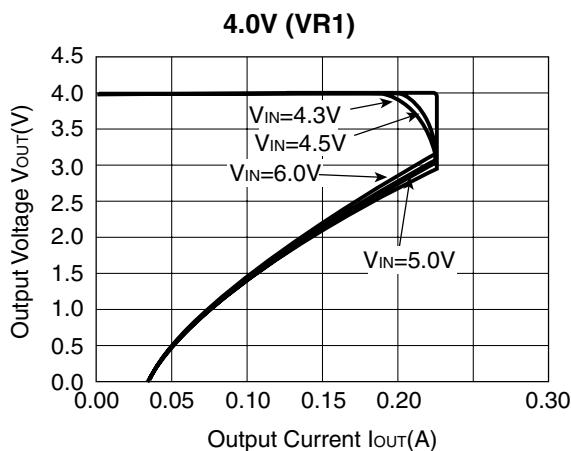
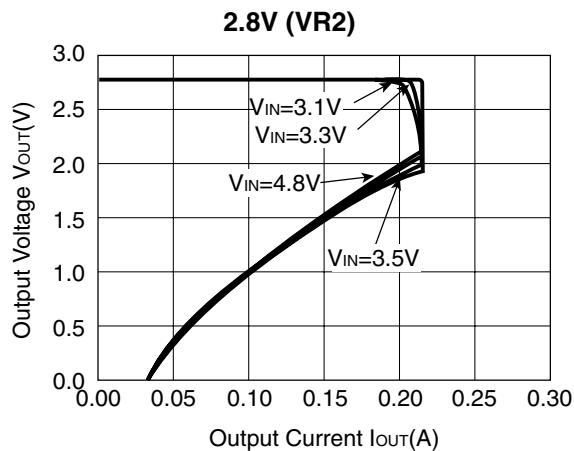
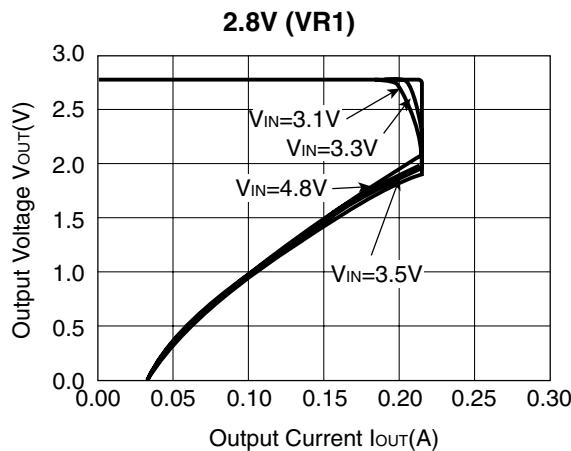
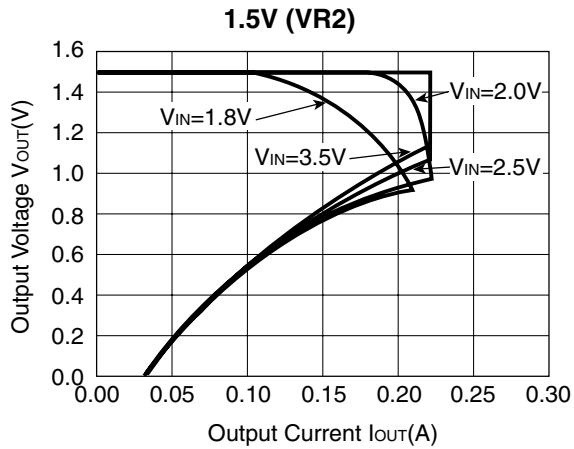
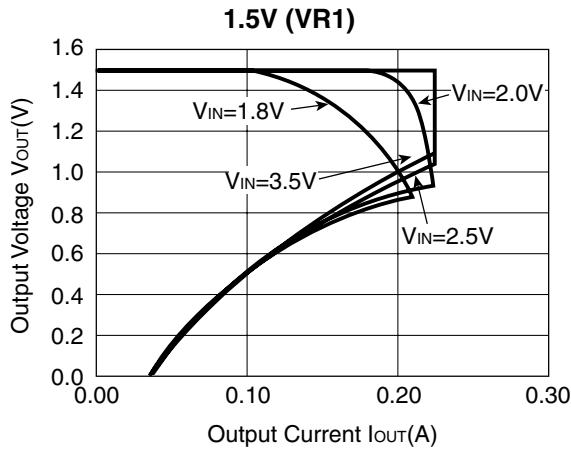
Fig.4 Load Transient Response Test Circuit

**TYPICAL APPLICATION**

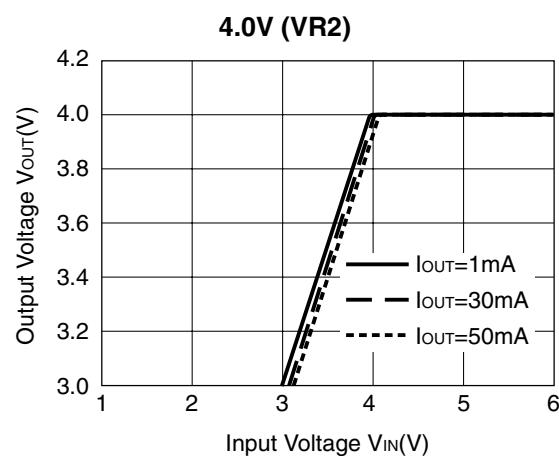
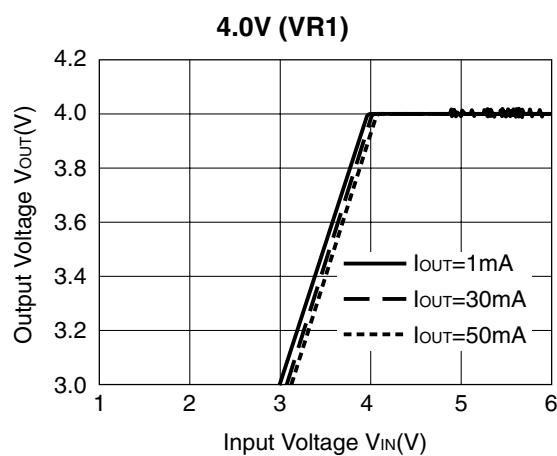
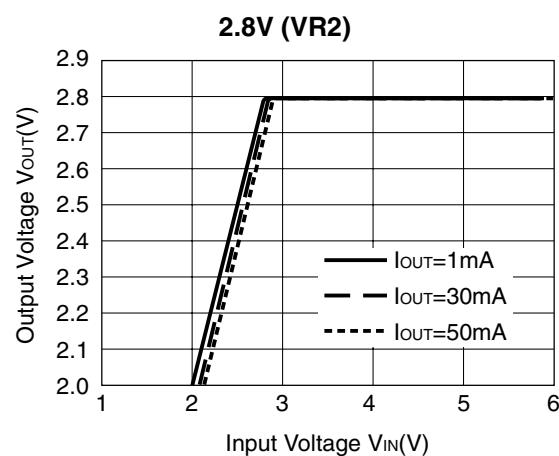
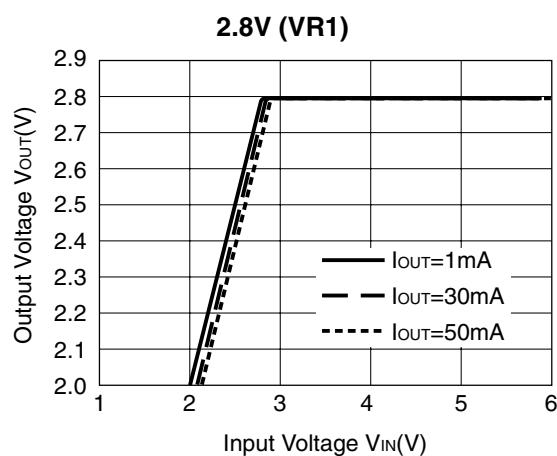
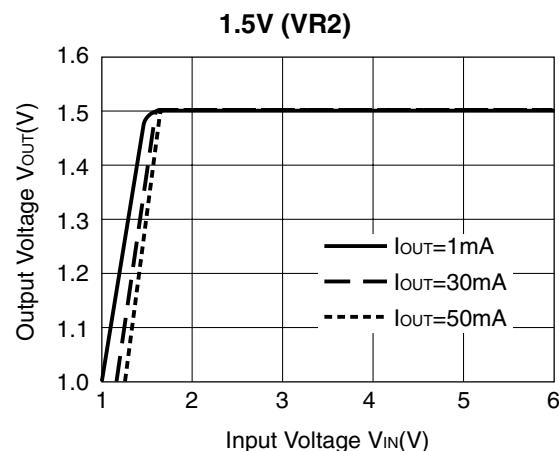
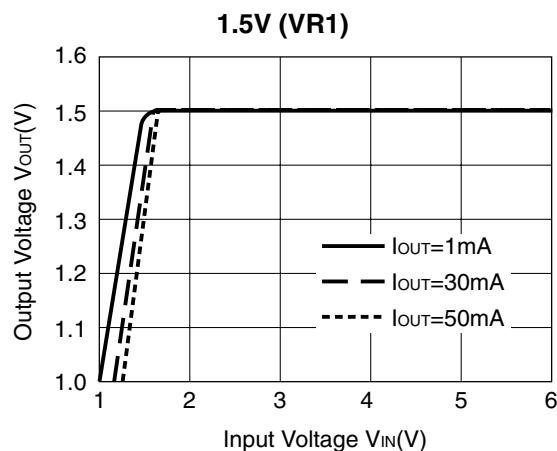
(External Components) Output Capacitor; Tantalum Type

## TYPICAL CHARACTERISTICS

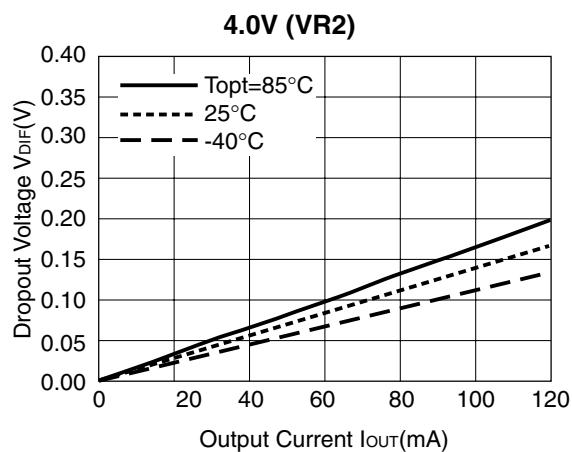
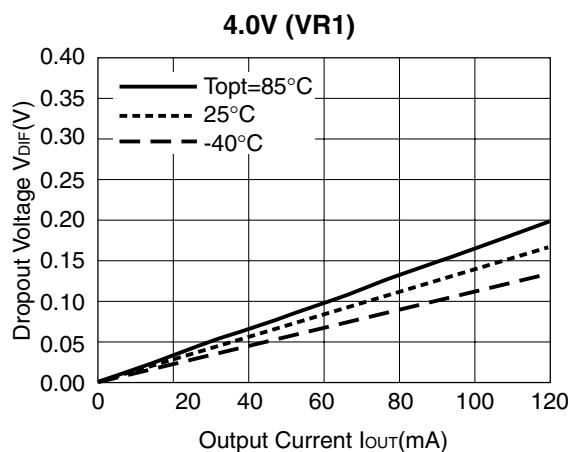
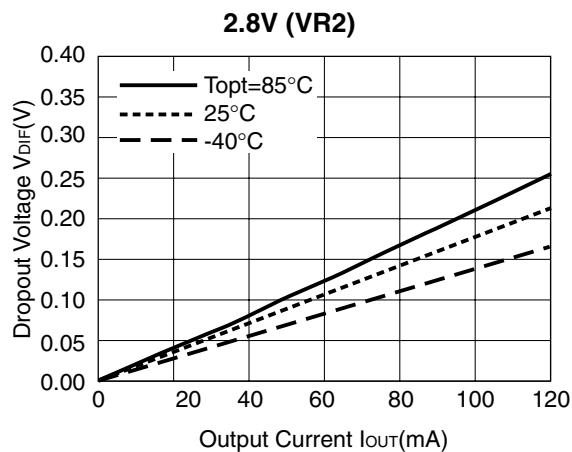
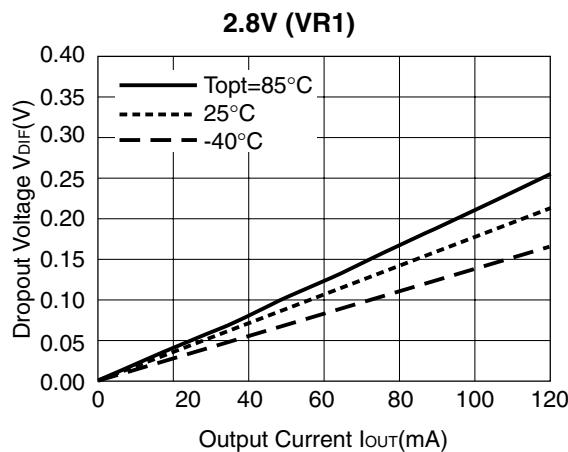
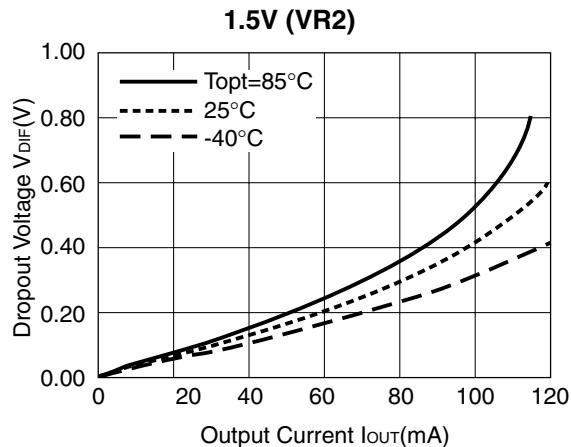
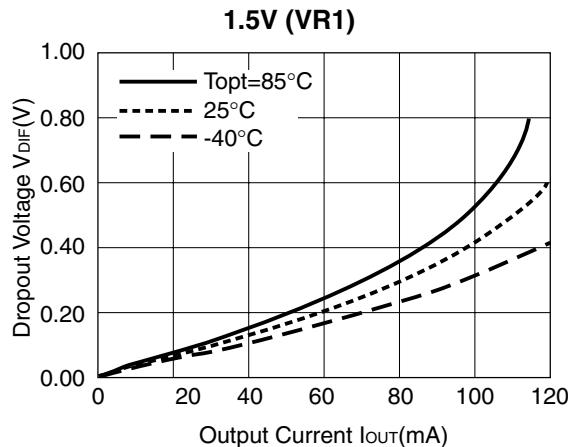
### 1) Output Voltage vs. Output Current



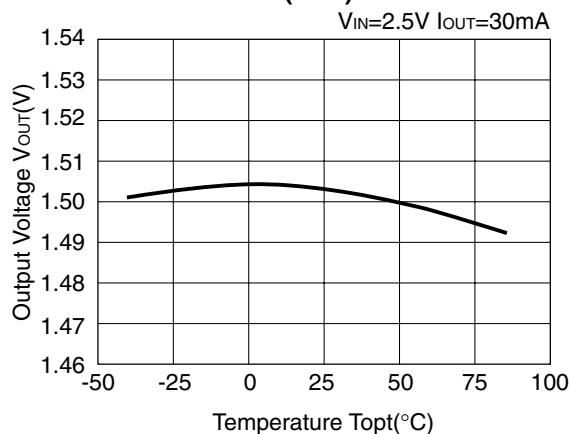
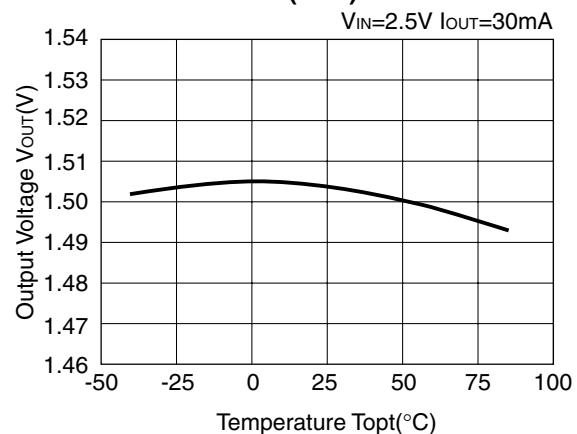
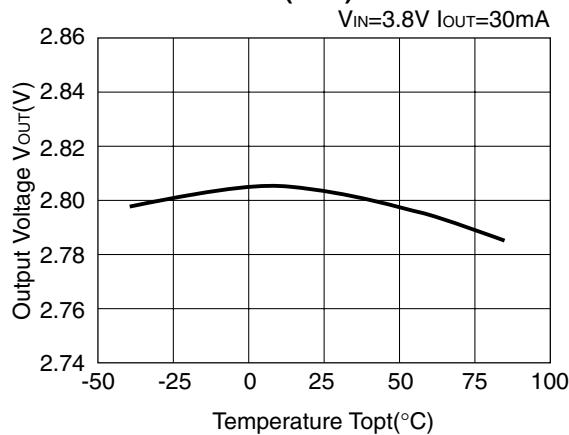
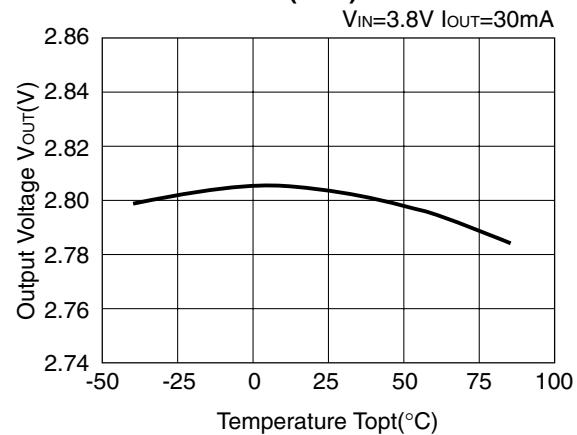
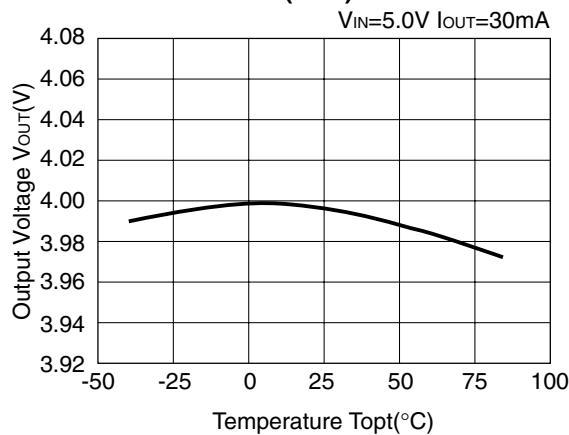
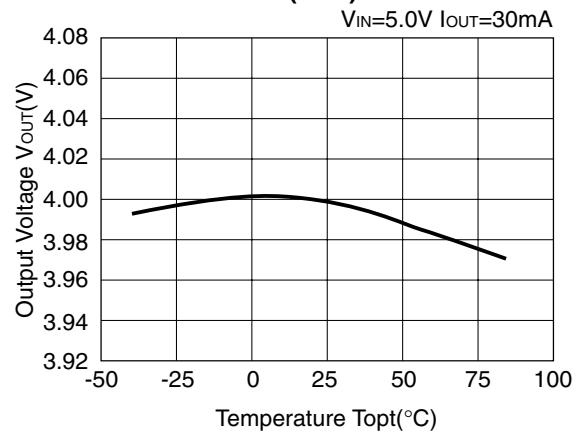
## 2) Output Voltage vs. Input Voltage



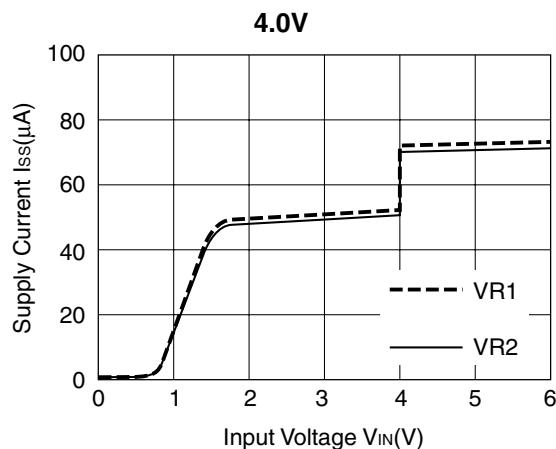
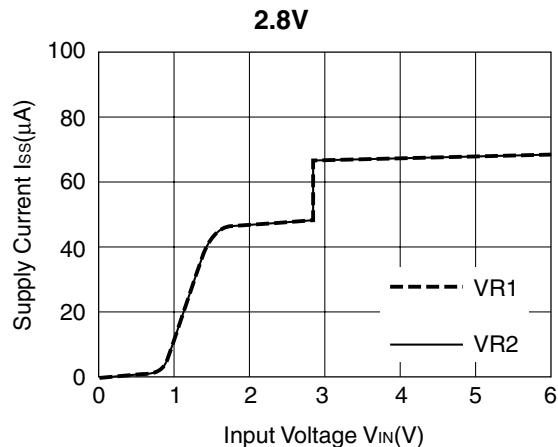
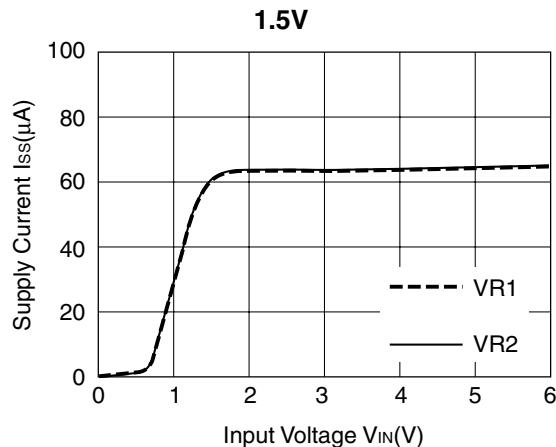
## 3) Dropout Voltage vs. Temperature



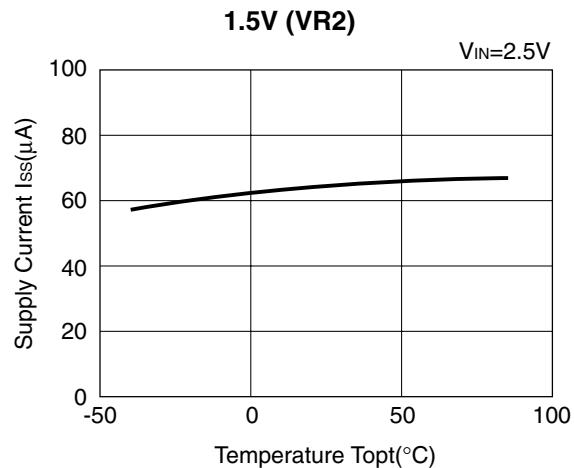
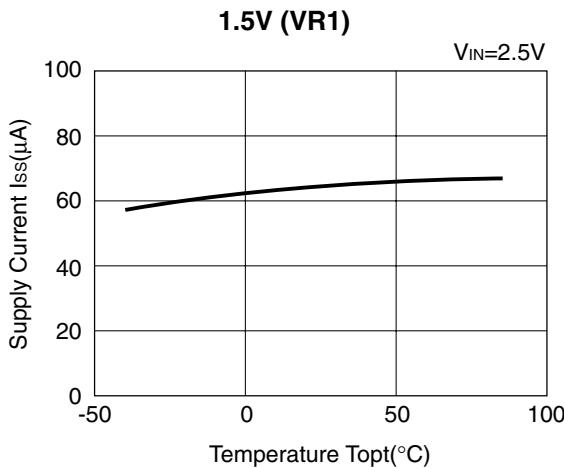
## 4) Output Voltage vs. Temperature

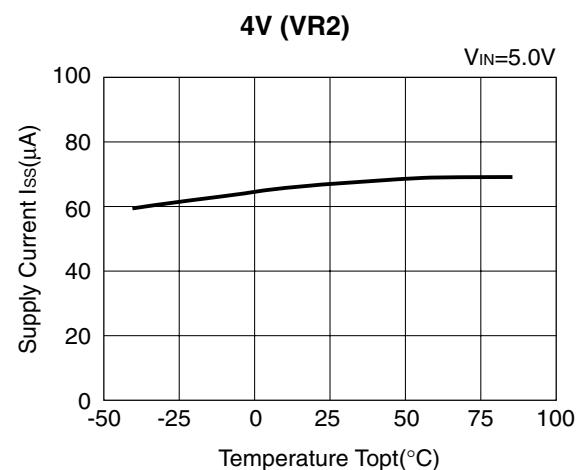
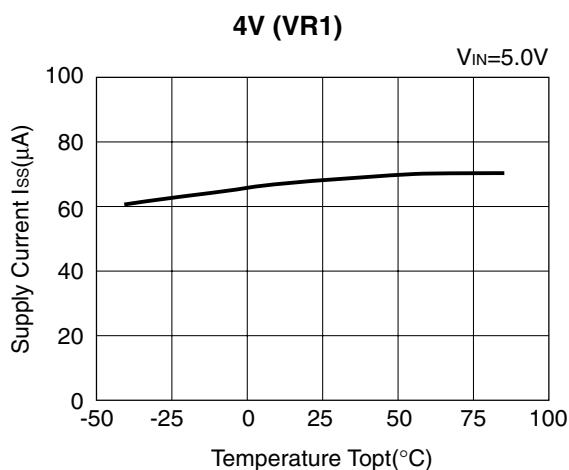
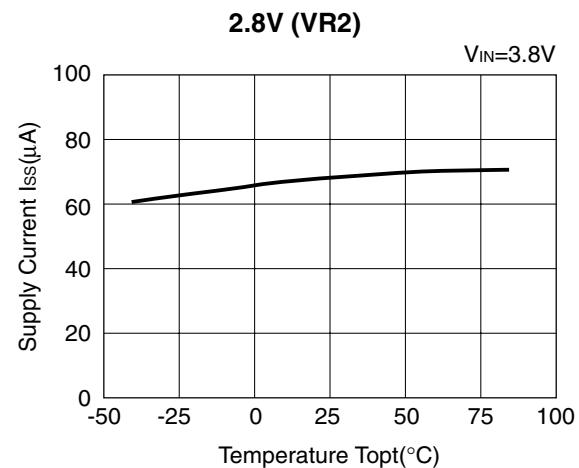
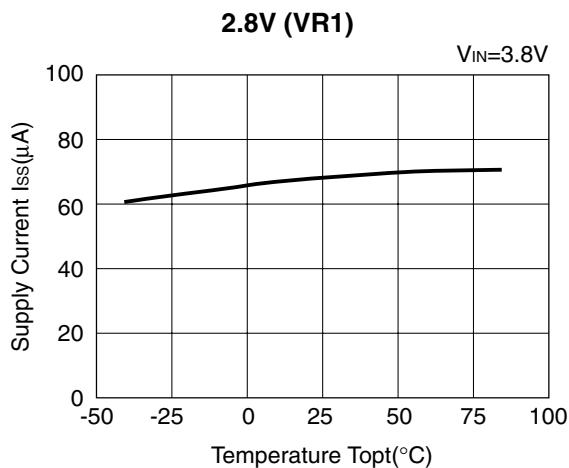
**1.5V (VR1)****1.5V (VR2)****2.8V (VR1)****2.8V (VR2)****4.0V (VR1)****4.0V (VR2)**

## 5) Supply Current vs. Input Voltage

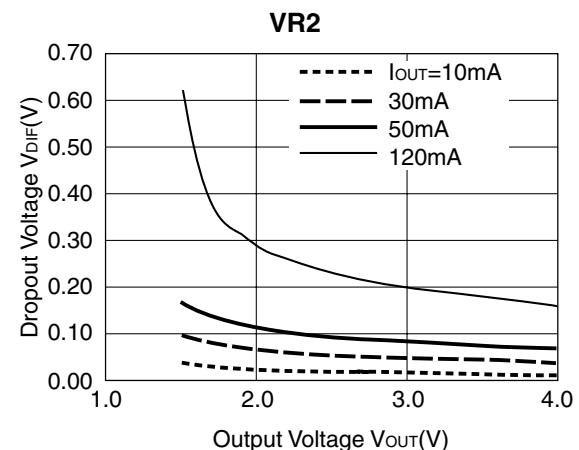
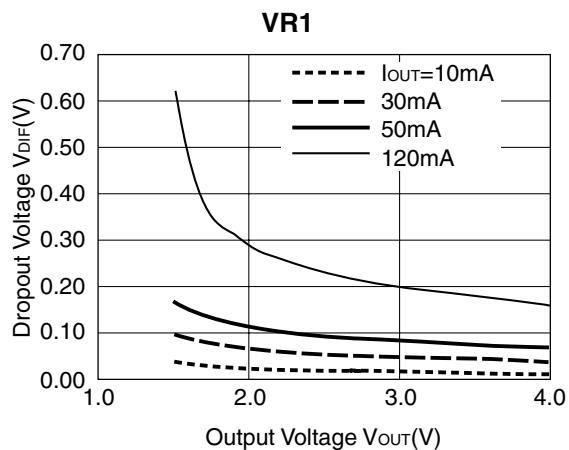


## 6) Supply Current vs. Temperature

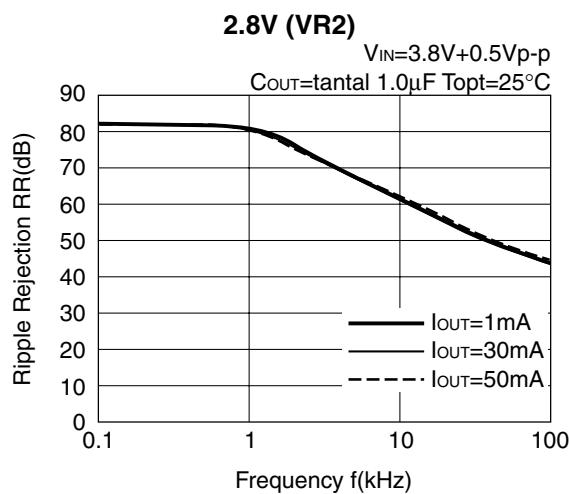
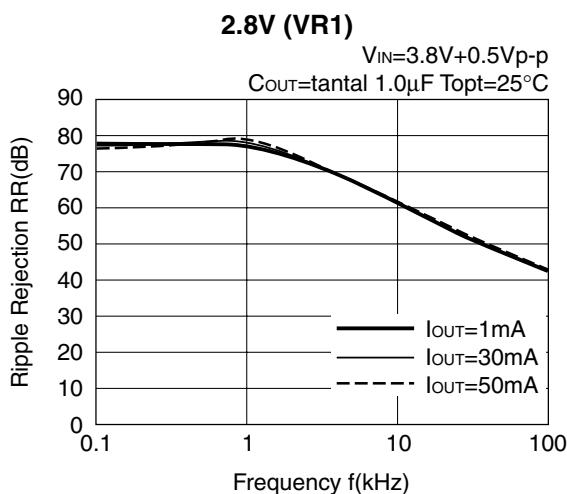
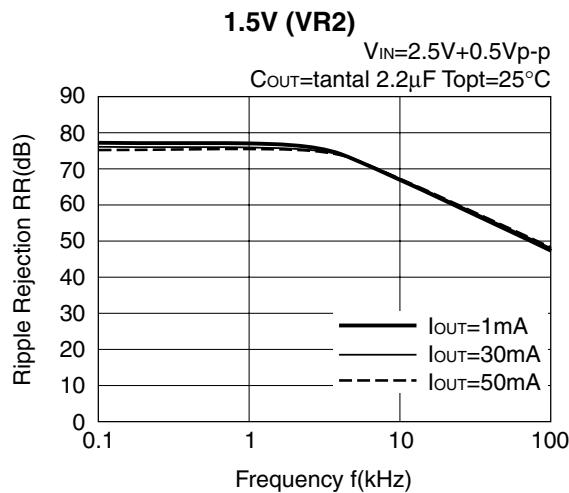
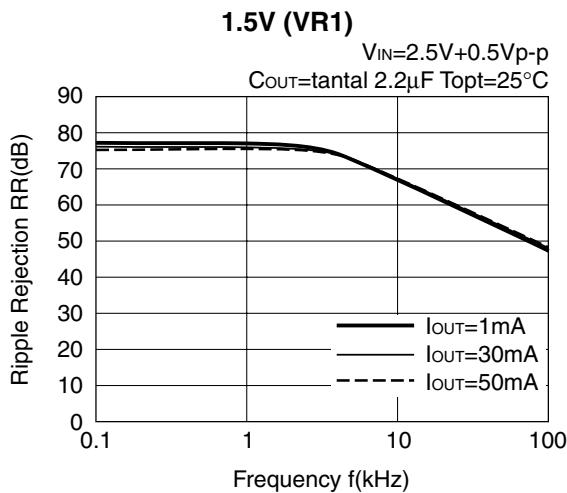
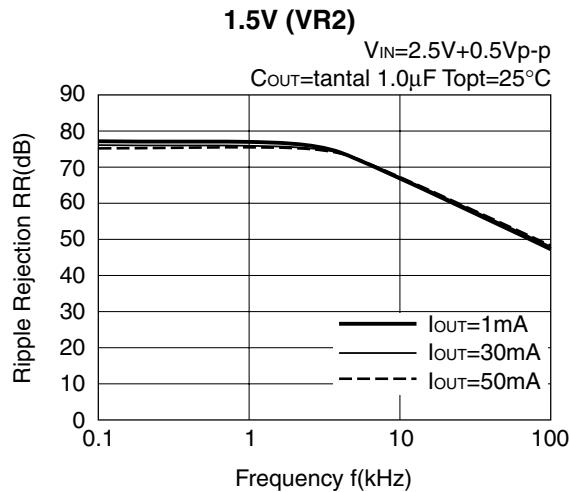
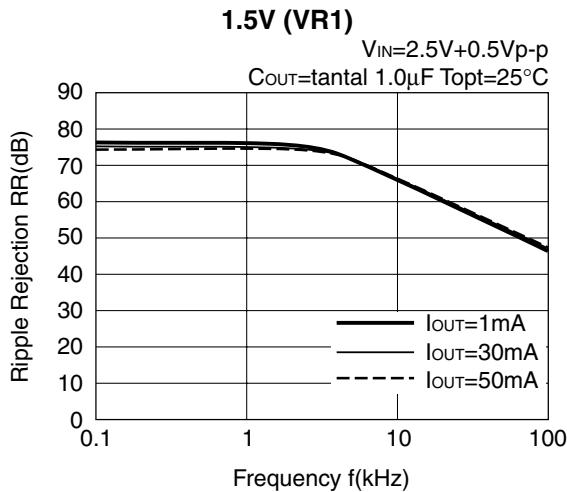


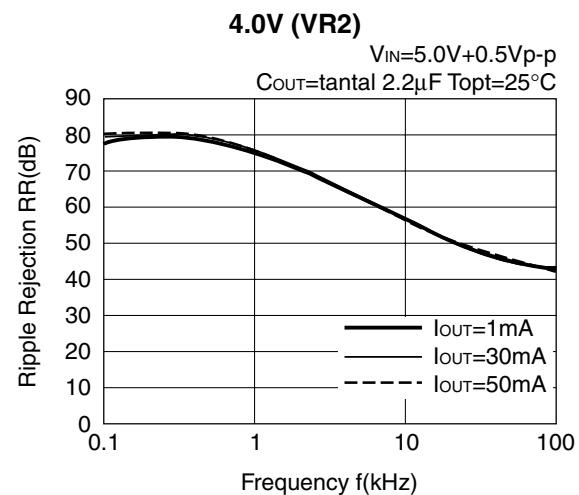
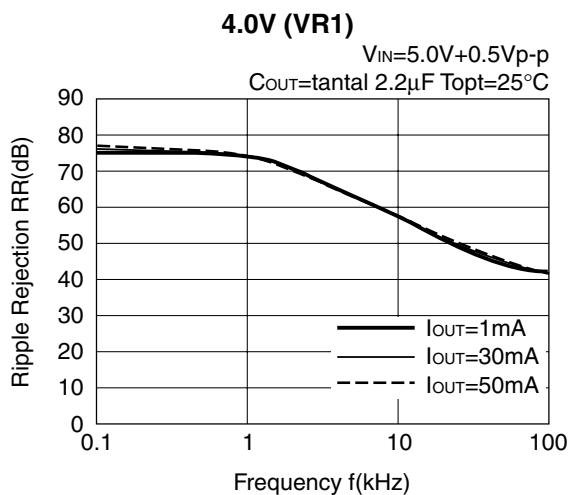
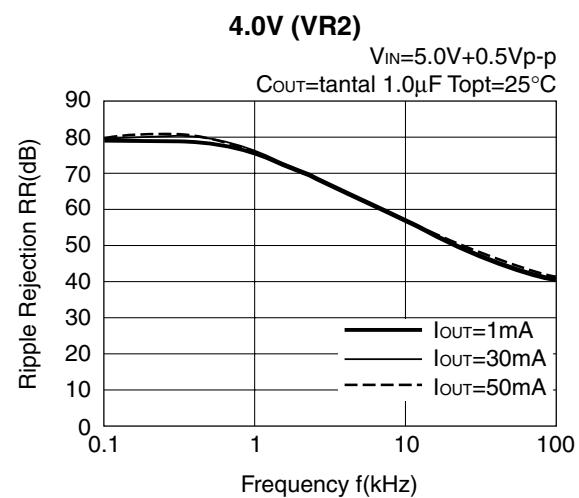
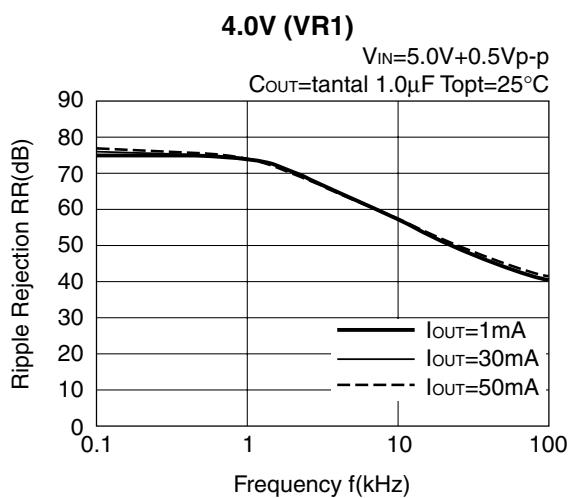
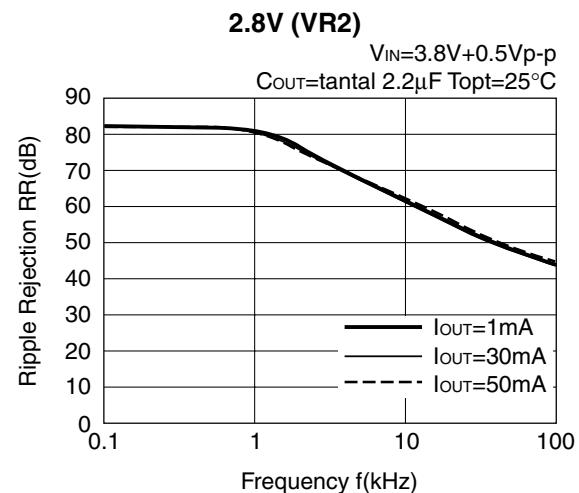
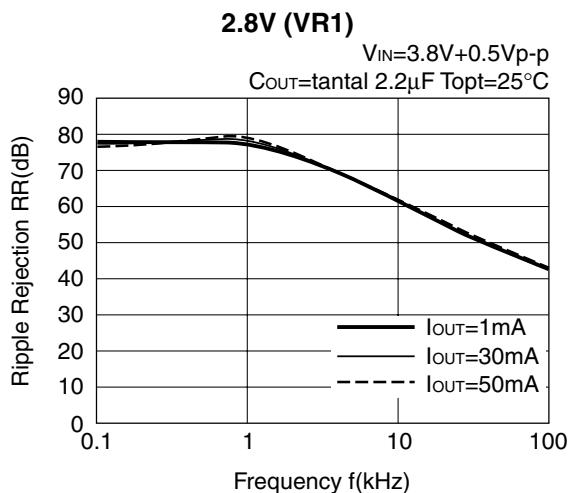


## 7) Dropout Voltage vs. Set Output Voltage

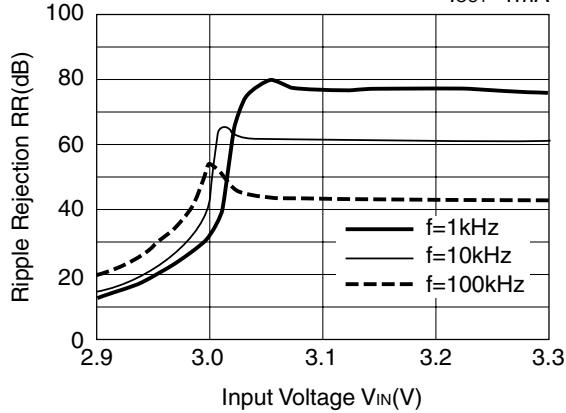
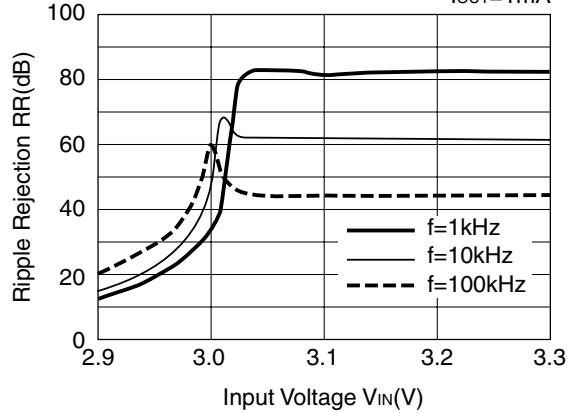
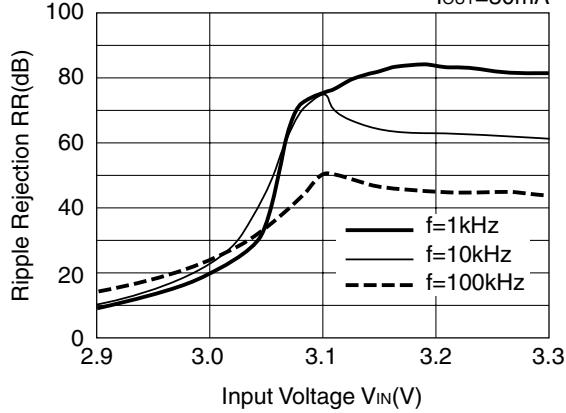
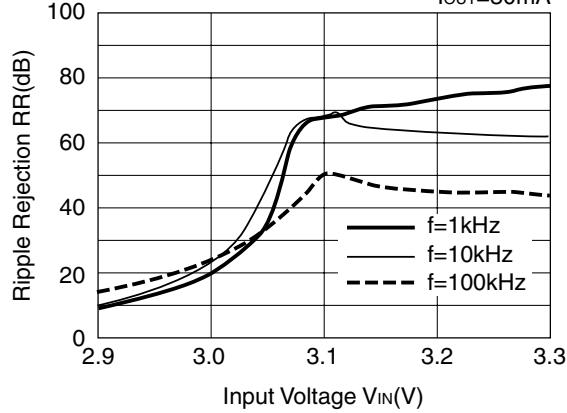
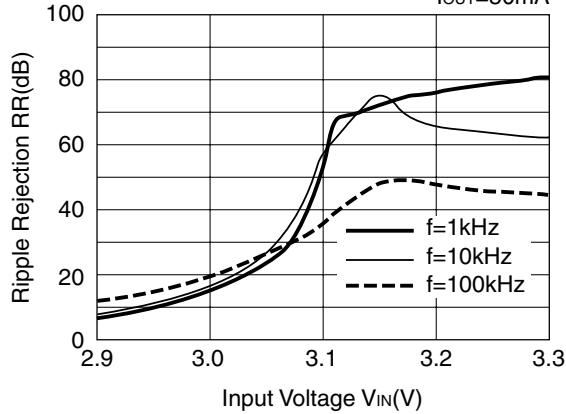
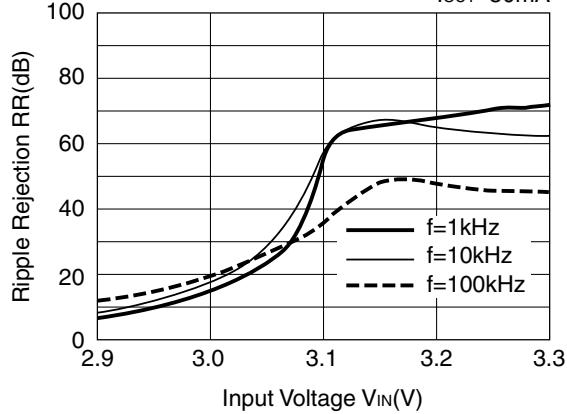


## 8) Ripple Rejection vs. Frequency





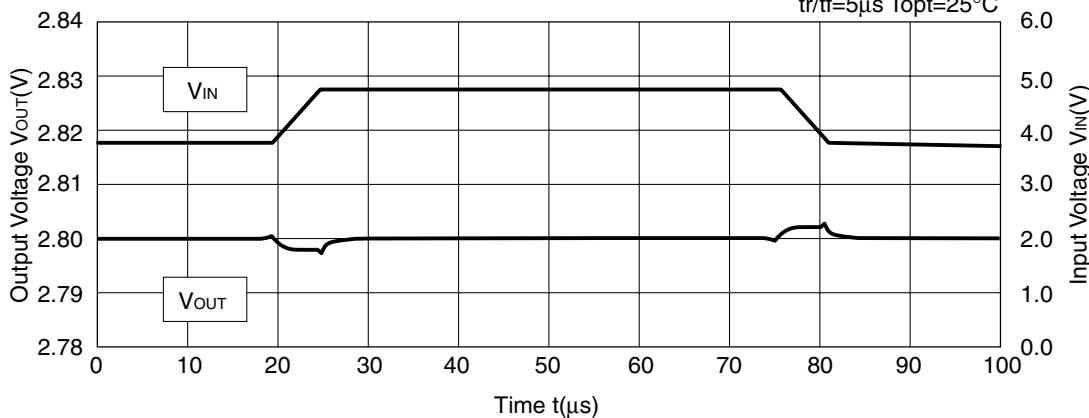
## 9) Ripple Rejection vs. Input Voltage (DC bias)

**2.8V (VR1)** $C_{OUT}$ =tantal  $2.2\mu F$   
 $I_{OUT}=1mA$ **2.8V (VR2)** $C_{OUT}$ =tantal  $2.2\mu F$   
 $I_{OUT}=1mA$ **2.8V (VR1)** $C_{OUT}$ =tantal  $2.2\mu F$   
 $I_{OUT}=30mA$ **2.8V (VR2)** $C_{OUT}$ =tantal  $2.2\mu F$   
 $I_{OUT}=30mA$ **2.8V (VR1)** $C_{OUT}$ =tantal  $2.2\mu F$   
 $I_{OUT}=50mA$ **2.8V (VR2)** $C_{OUT}$ =tantal  $2.2\mu F$   
 $I_{OUT}=50mA$ 

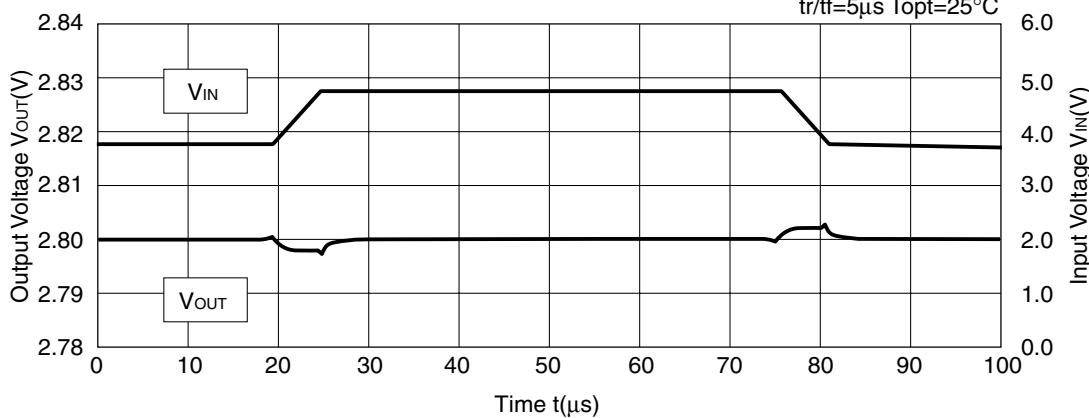
## 10) Input Transient Response

**R5322N001x (2.8V, VR1)**

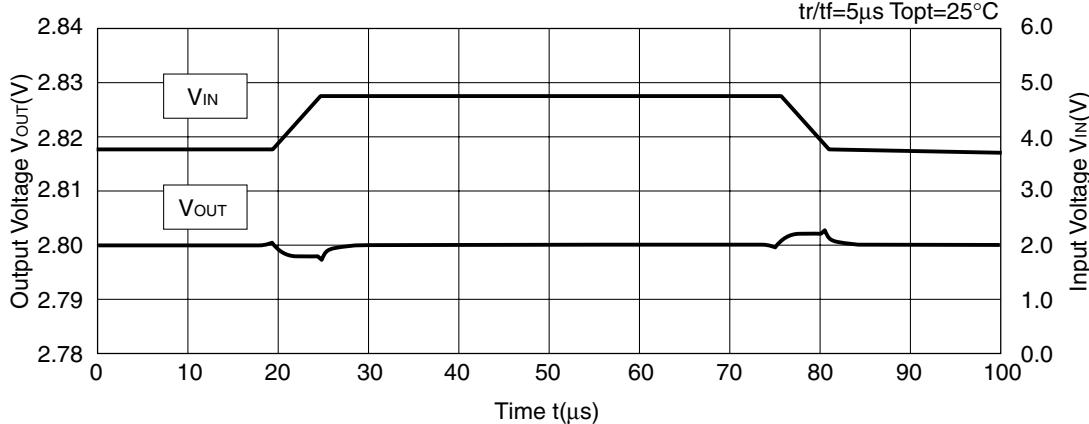
$I_{OUT}=30mA$   $C_{OUT}=\text{tantal } 1.0\mu F$   
 $t_r/t_f=5\mu s$   $T_{opt}=25^\circ C$

**R5322N001x (2.8V, VR1)**

$I_{OUT}=30mA$   $C_{OUT}=\text{tantal } 2.2\mu F$   
 $t_r/t_f=5\mu s$   $T_{opt}=25^\circ C$

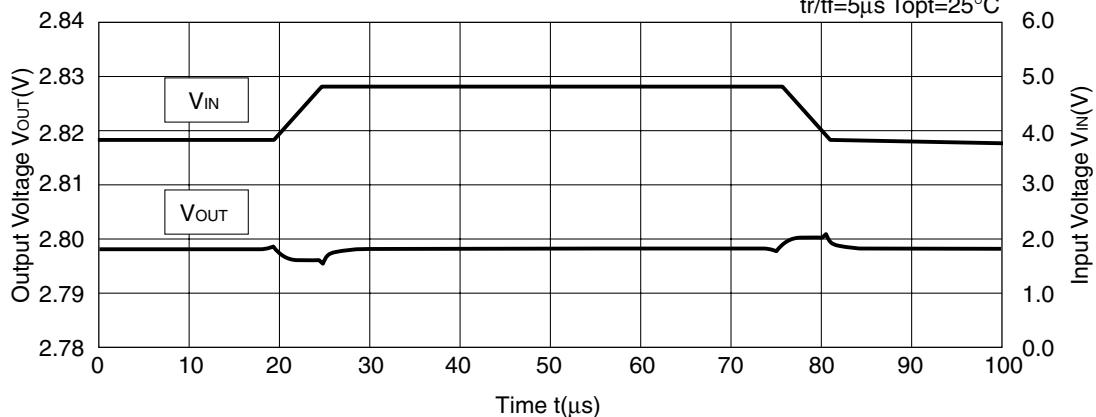
**R5322N001x (2.8V, VR1)**

$I_{OUT}=30mA$   $C_{OUT}=\text{tantal } 6.8\mu F$   
 $t_r/t_f=5\mu s$   $T_{opt}=25^\circ C$

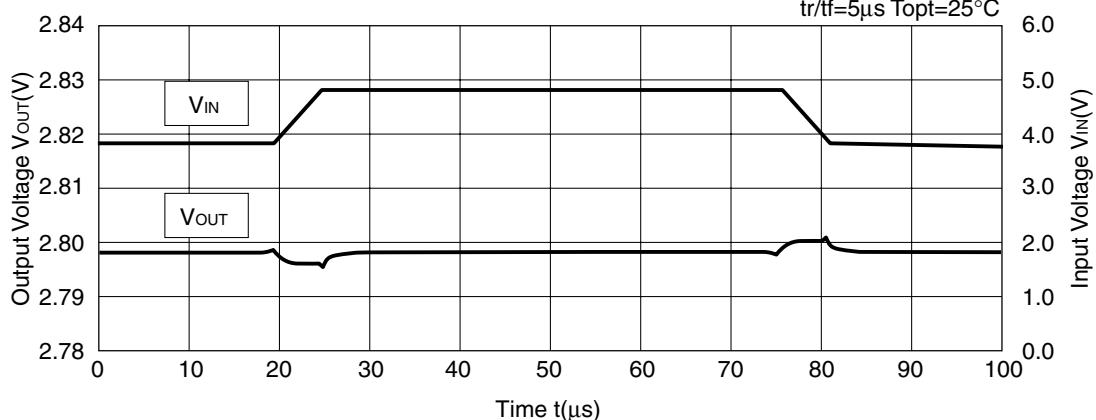


**R5322N001x (2.8V, VR2)**

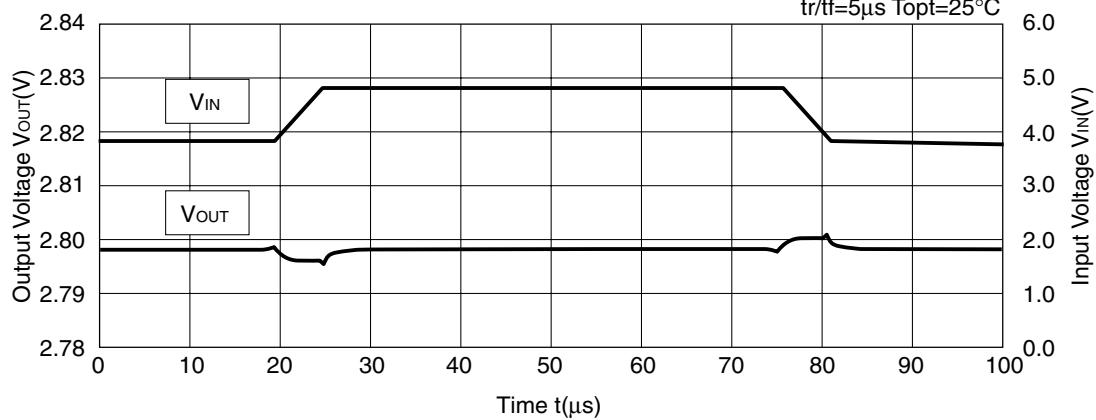
$I_{OUT}=30mA$   $C_{OUT}=\text{tantal } 1.0\mu F$   
 $t_r/t_f=5\mu s$   $T_{opt}=25^\circ C$

**R5322N001x (2.8V, VR2)**

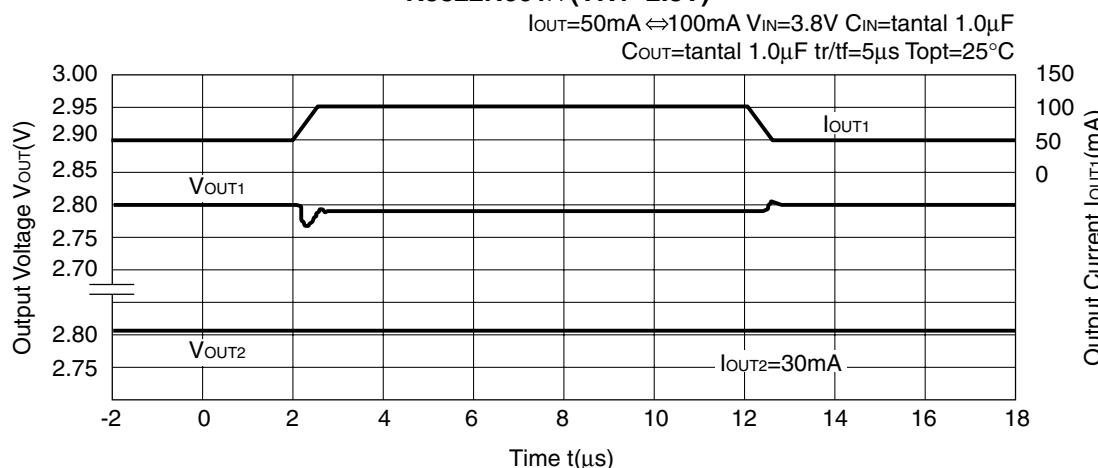
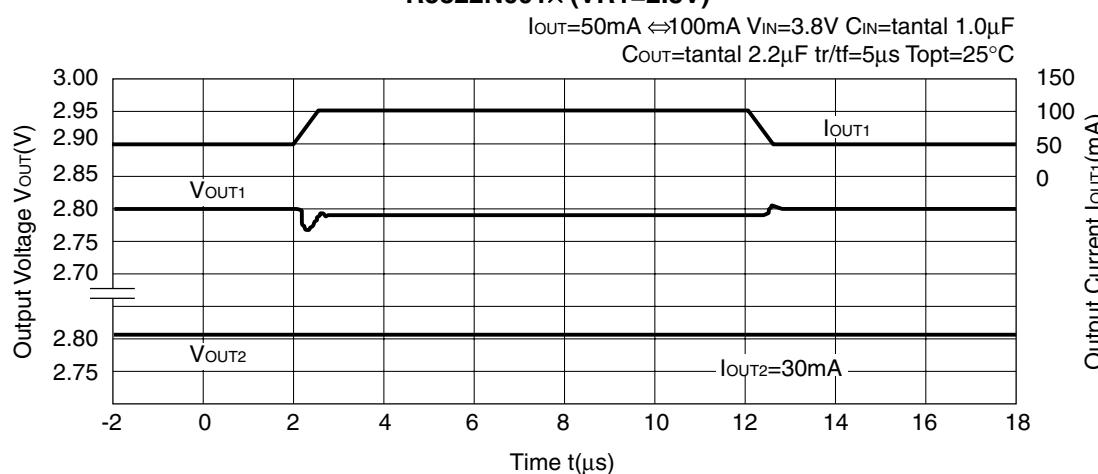
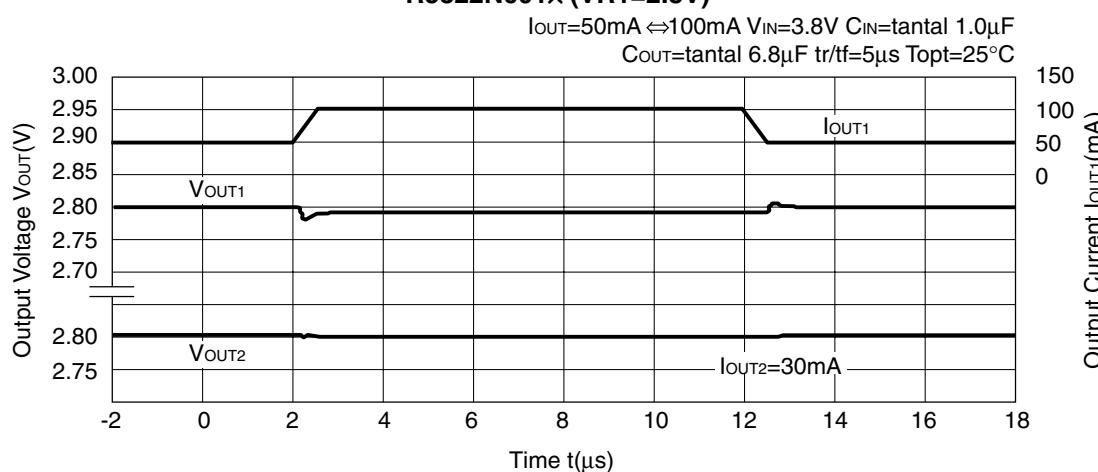
$I_{OUT}=30mA$   $C_{OUT}=\text{tantal } 2.2\mu F$   
 $t_r/t_f=5\mu s$   $T_{opt}=25^\circ C$

**R5322N001x (2.8V, VR2)**

$I_{OUT}=30mA$   $C_{OUT}=\text{tantal } 6.8\mu F$   
 $t_r/t_f=5\mu s$   $T_{opt}=25^\circ C$

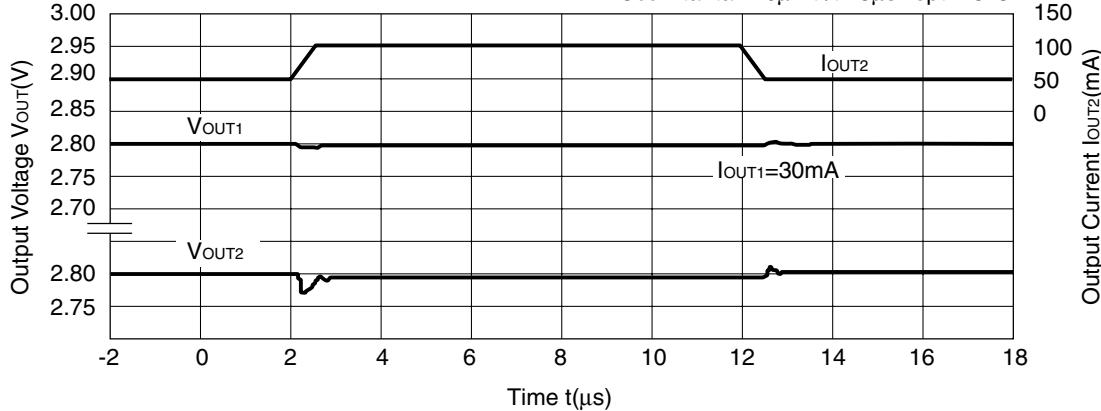


## 11) Load Transient Response

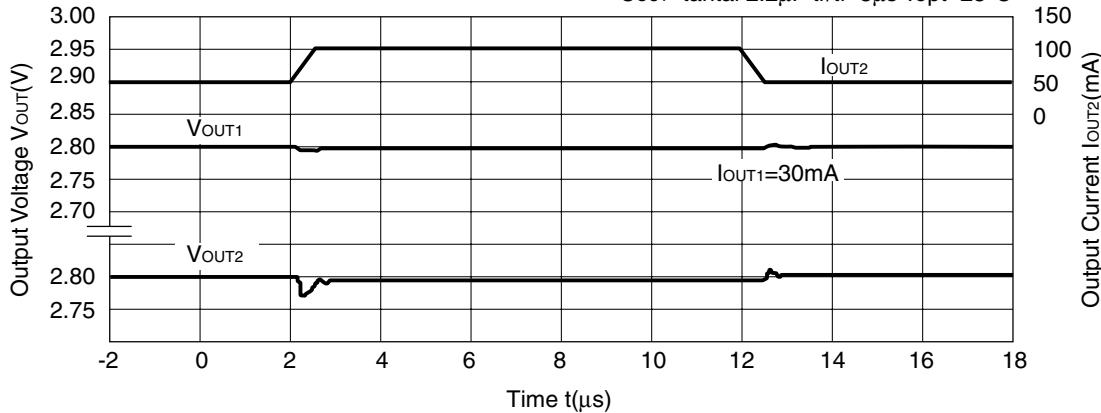
**R5322N001x (VR1=2.8V)****R5322N001x (VR1=2.8V)****R5322N001x (VR1=2.8V)**

**R5322N001x (VR2=2.8V)**

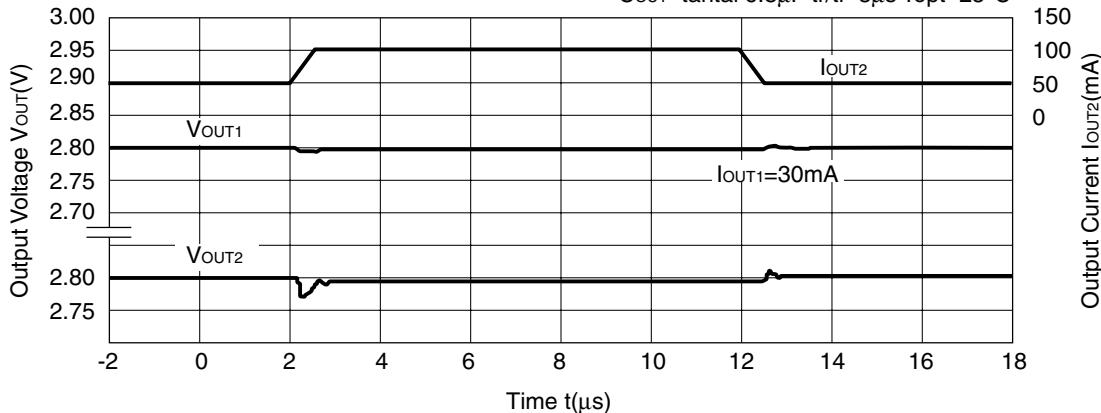
$I_{OUT}=50\text{mA} \leftrightarrow 100\text{mA}$   $V_{IN}=3.8\text{V}$   $C_{IN}=\text{tantal } 1.0\mu\text{F}$   
 $C_{OUT}=\text{tantal } 1.0\mu\text{F}$   $tr/tr=5\mu\text{s}$   $T_{opt}=25^\circ\text{C}$

**R5322N00x (VR2=2.8V)**

$I_{OUT}=50\text{mA} \leftrightarrow 100\text{mA}$   $V_{IN}=3.8\text{V}$   $C_{IN}=\text{tantal } 1.0\mu\text{F}$   
 $C_{OUT}=\text{tantal } 2.2\mu\text{F}$   $tr/tr=5\mu\text{s}$   $T_{opt}=25^\circ\text{C}$

**R5322N00x (VR2=2.8V)**

$I_{OUT}=50\text{mA} \leftrightarrow 100\text{mA}$   $V_{IN}=3.8\text{V}$   $C_{IN}=\text{tantal } 1.0\mu\text{F}$   
 $C_{OUT}=\text{tantal } 6.8\mu\text{F}$   $tr/tr=5\mu\text{s}$   $T_{opt}=25^\circ\text{C}$



## TECHNICAL NOTES

When using these ICs, consider the following points:

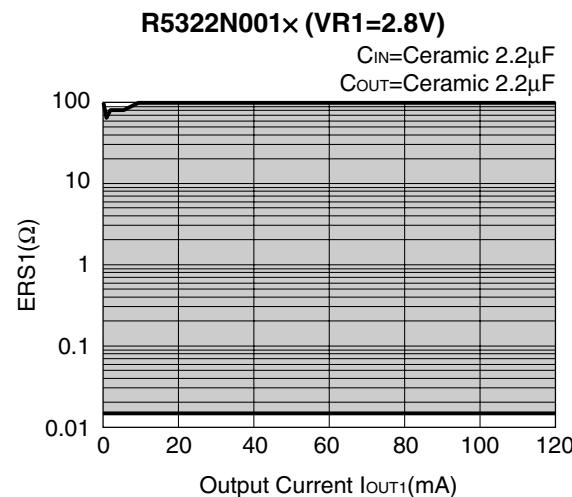
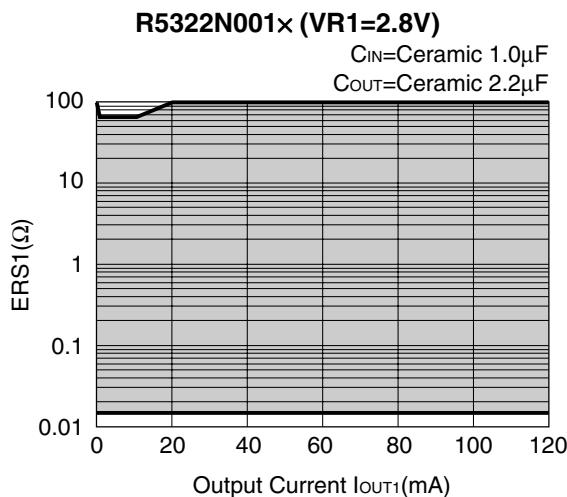
In these ICs, phase compensation is made for securing stable operation even if the load current is varied. For this purpose, be sure to use a  $2.2\mu\text{F}$  or more capacitance  $C_{\text{OUT}}$  with good frequency characteristics and ESR (Equivalent Series Resistance) of which is in the range described as follows:

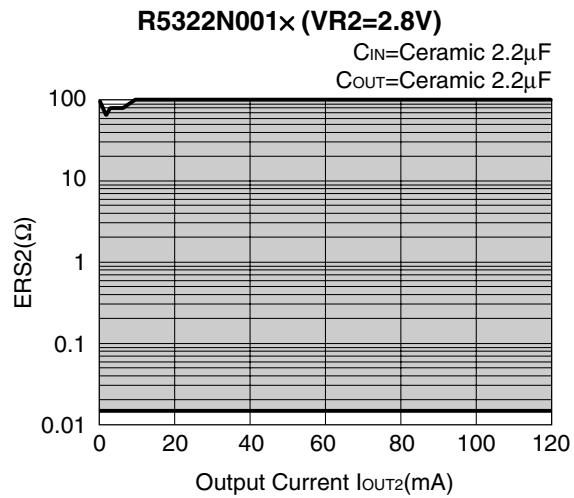
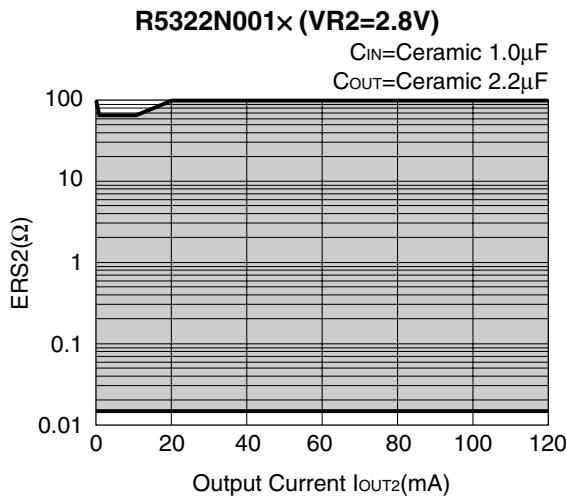
The relations between  $I_{\text{OUT}}$  (Output Current) and ESR of Output Capacitor are shown below. The conditions when the white noise level is under  $40\mu\text{V}$  (Avg.) are marked as the hatched area in the graph.

(Note: When a ceramic capacitor is connected to the Output Pin as Output capacitor for phase compensation, the operation might be unstable unless as much as  $1\Omega$  resistor is connected between the capacitor and GND instead of ESR. Test these ICs with same external components as ones to be used on the PCB.)

<Test conditions>

- (1)  $V_{\text{IN}}=3.8\text{V}$
- (2) Frequency band: 10Hz to 2MHz
- (3) Temperature:  $25^{\circ}\text{C}$





- Make V<sub>DD</sub> and GND line sufficient. When the impedance of these is high, the noise might be picked up or not work correctly.
- Connect the capacitor with a capacitance of 1μF or more between V<sub>DD</sub> and GND as close as possible.
- Set external components, especially Output Capacitor, as close as possible to the ICs and make wiring shortest.