



# **PCM1801**

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# 16-Bit, Stereo, Audio ANALOG-TO-DIGITAL CONVERTER

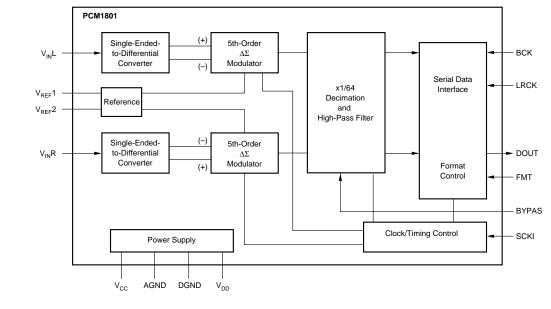
## FEATURES

- DUAL 16-BIT MONOLITHIC  $\Delta\Sigma$  ADC
- SINGLE-ENDED VOLTAGE INPUT
- 64X OVERSAMPLING DECIMATION FILTER: Passband Ripple: ±0.05dB Stopband Attenuation: -65dB
- HIGH PERFORMANCE: THD+N: -88dB (typ) SNR: 93dB (typ) Dynamic Range: 93dB (typ) Internal High-Pass Filter
- PCM AUDIO INTERFACE: Left Justified, I<sup>2</sup>S
- SAMPLING RATE: 4kHz to 48kHz
- SYSTEM CLOCK: 256f<sub>s</sub>, 384f<sub>s</sub>, or 512f<sub>s</sub>
- SINGLE +5V POWER SUPPLY
- SMALL 14-LEAD SOIC PACKAGE

# DESCRIPTION

PCM1801 is a low cost, single chip stereo analog-todigital converter with single-ended analog voltage inputs. The PCM1801 uses a delta-sigma modulator with 64x oversampling, including a digital decimation filter and serial interface which supports Slave mode operation and two data formats. The PCM1801 is suitable for a wide variety of cost-sensitive consumer applications where performance is required.

PCM1801 is fabricated on a highly-advanced CMOS process.



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## **SPECIFICATIONS**

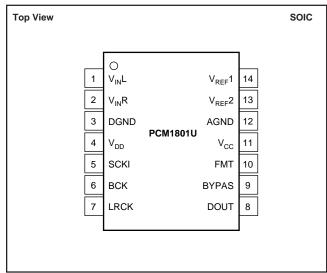
All specifications at +25°C, +V<sub>DD</sub> = +V<sub>CC</sub> = +5V,  $f_S$  = 44.1kHz, and 16-bit data, SYSCLK = 384 $f_S$ , unless otherwise noted.

			PCM1801U		
PARAMETER	CONDITIONS	MIN	TYP MAX		UNITS
RESOLUTION			16		
DIGITAL INPUT/OUTPUT					
Input Logic Level:					
$V_{\rm H}^{(1)}$		2.0			V
V <sub>IL</sub> <sup>(1)</sup> Input Logic Current:				0.8	V
$I_{\rm IN}^{(2)}$				±1	μA
IN (3)				+100	μΑ
Output Logic Level:					
V <sub>OH</sub> <sup>(4)</sup>	$I_{OH} = -1.6mA$	4.5			V
	$I_{OL} = +3.2mA$			0.5	V
Sampling Frequency System Clock Frequency	256f <sub>S</sub>	4 1.024	44.1 11.2896	48 12.2880	kHz MHz
System Clock Frequency	384f <sub>S</sub>	1.536	16.9344	18.4320	MHz
	512f <sub>S</sub>	2.024	22.5792	24.5760	MHz
DC ACCURACY					
Gain Mismatch Channel-to-Channel			±1.0	±2.5	% of FSR
Gain Error			±2.0	±5.0	% of FSR
Gain Drift			±20		ppm of FSR/°
Bipolar Zero Error	High-Pass Filter Bypass		±2.0		% of FSR
Bipolar Zero Drift	High-Pass Filter Bypass		±20		ppm of FSR/°
					-10
THD+N at FS (-0.5dB) THD+N at -60dB				-80	dB dB
Dynamic Range	EIAJ, A-weighted	90	93		dB
Signal-To-Noise Ratio	EIAJ, A-weighted	90	93		dB
Channel Separation	_	88	91		dB
ANALOG OUTPUT					
Input Range	FS (V <sub>IN</sub> = 0dB)		2.828		Vp-p
Center Voltage			2.1		V
Input Impedance Anti-Aliasing Filter Frequency Response	-3dB		30 170		kΩ kHz
DIGITAL FILTER PERFORMANCE	Jour Jour Jour Jour Jour Jour Jour Jour		170		IN 12
Passband				0.454f <sub>S</sub>	Hz
Stopband		0.583f <sub>S</sub>			Hz
Passband Ripple		Ű		±0.05	dB
Stopband Attenuation		-65			dB
Delay Time (Latency)			17.4/f <sub>S</sub>		sec
High Pass Frequency Response	–3dB			0.019f <sub>S</sub>	mHz
POWER SUPPLY REQUIREMENTS					1/00
Voltage Range	+V <sub>CC</sub> +V <sub>DD</sub>	+4.5	+5.0 +5.0	+5.5 +5.5	VDC VDC
Supply Current <sup>(6)</sup>	$+V_{DD}$ $+V_{CC} = +V_{DD} = +5V$	+4.0	+5.0	+5.5	mA
Power Dissipation	$+V_{CC} = +V_{DD} = +5V$		90	125	mW
TEMPERATURE RANGE					
Operation		-25		+85	°C
Storage		-55		+125	°C
Thermal Resistance, $\theta_{JA}$			100		°C/W

NOTES: (1) Pins 5, 6, 7, 9, and 10 (SCKI, BCK, LRCK, BYPAS, FMT). (2) Pins 5, 6, 7 (SCKI, BCK, LRCK) Schmitt-Trigger input. (3) Pins 9, 10 (BYPAS, FMT) Schmitt-Trigger input with 100k $\Omega$  typical pull-down resistor). (4) Pin 8 (DOUT). (5)  $f_{IN}$  = 1kHz, using Audio Precisions System II, rms Mode with 20kHz LPF and 400Hz HPF enabled. (6) No load on DOUT (pin 8).



#### **PIN CONFIGURATION**



#### **ABSOLUTE MAXIMUM RATINGS**

Supply Voltage: +V <sub>DD</sub> , +V <sub>CC</sub> Supply Voltage Differences	
GND Voltage Differences	
Digital Input Voltage	–0.3V to (V <sub>DD</sub> + 0.3V)
Analog Input Voltage	–0.3V to (V <sub>CC</sub> + 0.3V)
Input Current (any pin except supplies)	±10mA
Power Dissipation	
Operating Temperature Range	–25°C to +85°C
Storage Temperature	–55°C to +125°C
Lead Temperature (soldering, 5s)	+260°C
(reflow, 10s)	+235°C

#### **PIN ASSIGNMENTS**

PIN	NAME	I/O	DESCRIPTION	
1	V <sub>IN</sub> L	IN	Analog Input, Lch.	
2	V <sub>IN</sub> R	IN	Analog Input, Rch.	
3	DGND	—	Digital Ground	
4	V <sub>DD</sub>	—	Digital Power Supply	
5	SCKI	IN	System Clock Input; 256 $f_S$ , 384 $f_S$ , or 512 $f_S$ .	
6	BCK	IN	Bit Clock Input	
7	LRCK	IN	Sampling Clock Input	
8	DOUT	OUT	Audio Data Output	
9	BYPAS	IN	HPF Bypass Control <sup>(1)</sup> L: HPF Enabled	
			H: HPF Disabled	
10	FMT	IN	Audio Data Format <sup>(1)</sup> L: MSB-First, Left-Justified	
			H: MSB-First, I <sup>2</sup> S	
11	V <sub>CC</sub>	—	Analog Power Supply	
12	AGND	_ _	Analog Ground	
13	V <sub>REF</sub> 2	—	Reference 2 Decoupling Capacitor	
14	$V_{REF}1$	—	Reference 1 Decoupling Capacitor	

NOTE: (1) With  $100k\Omega$  typical pull-down resistor.

## ELECTROSTATIC DISCHARGE SENSITIVITY

This integrated circuit can be damaged by ESD. Burr-Brown recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

#### PACKAGE/ORDERING INFORMATION

PRODUCT	PACKAGE	PACKAGE DRAWING NUMBER	SPECIFIED TEMPERATURE RANGE	PACKAGE MARKING	ORDERING NUMBER <sup>(1)</sup>	TRANSPORT MEDIA
PCM1801U	SO-14	235	–25°C to +85°C	PCM1801U	PCM1801U	Rails
"	"	"	"	"	PCM1801U/2K	Tape and Reel

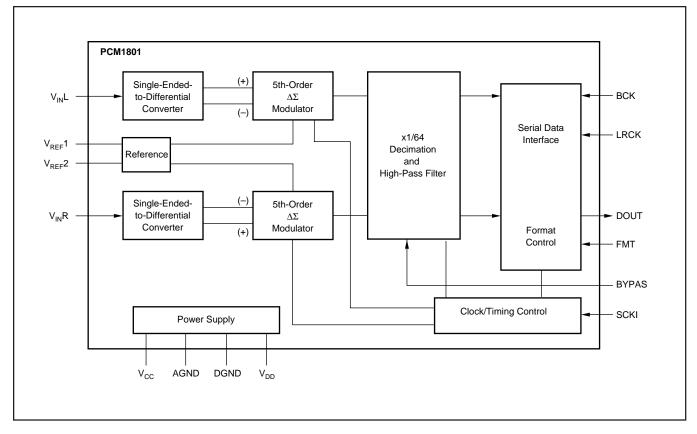
NOTE: (1) Models with a slash (/) are available only in Tape and Reel in the quantities indicated (e.g., /2K indicates 2000 devices per reel). Ordering 2000 pieces of "PCM1801U/2K" will get a single 2000-piece Tape and Reel.

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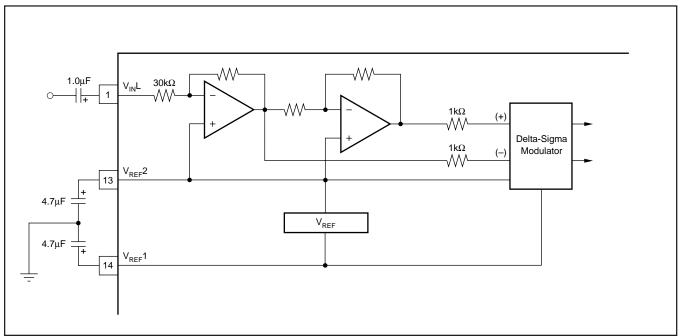


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#### **BLOCK DIAGRAM**



#### ANALOG FRONT-END (Single-Channel)

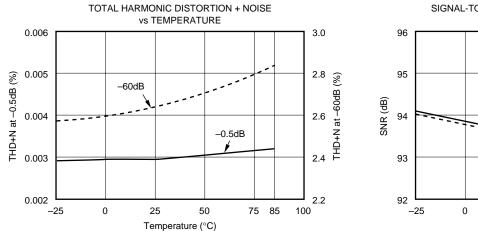


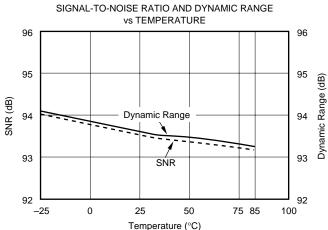
BURR-BROWN® PCM1801

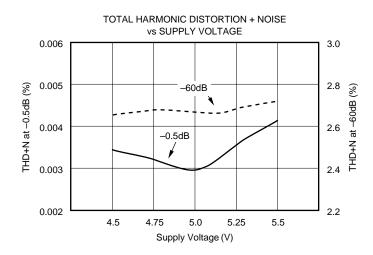
## **TYPICAL PERFORMANCE CURVES**

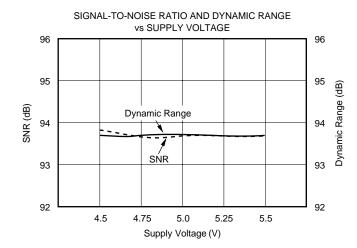
At  $T_A = +25^{\circ}C$ ,  $+V_{DD} = +V_{CC} = +5V$ ,  $f_S = 44.1 \text{kHz}$ , and SYSCLK =  $384f_S$ , unless otherwise noted.

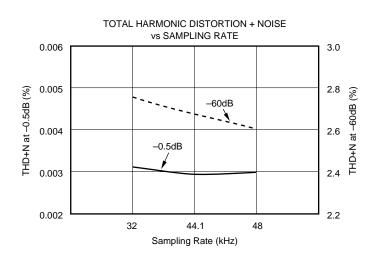
#### ANALOG DYNAMIC PERFORMANCE

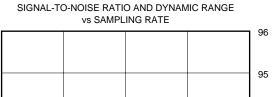


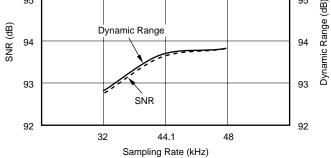












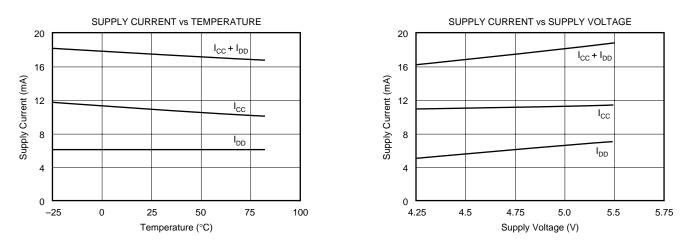


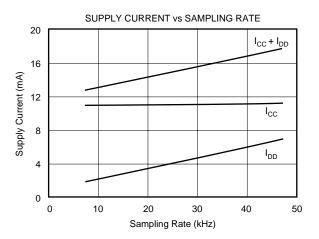
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At  $T_A = +25^{\circ}C$ ,  $+V_{DD} = +V_{CC} = +5V$ ,  $f_S = 44.1$ kHz, and SYSCLK = 384 $f_S$ , unless otherwise noted.

#### ANALOG DYNAMIC PERFORMANCE (cont.)

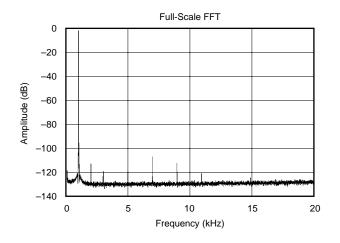


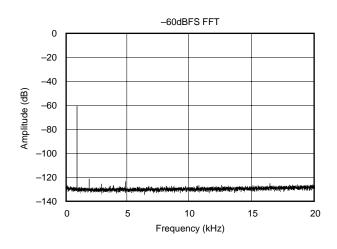


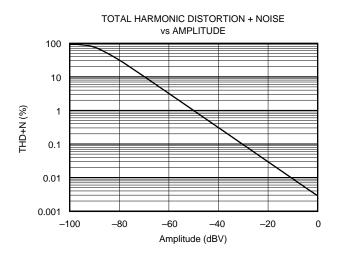


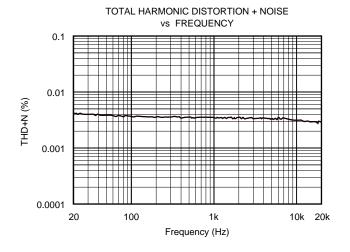
At  $T_A = +25^{\circ}C$ ,  $+V_{DD} = +V_{CC} = +5V$ ,  $f_S = 44.1 \text{kHz}$ , and SYSCLK =  $384f_S$ , unless otherwise noted.

#### **OUTPUT SPECTRUM**







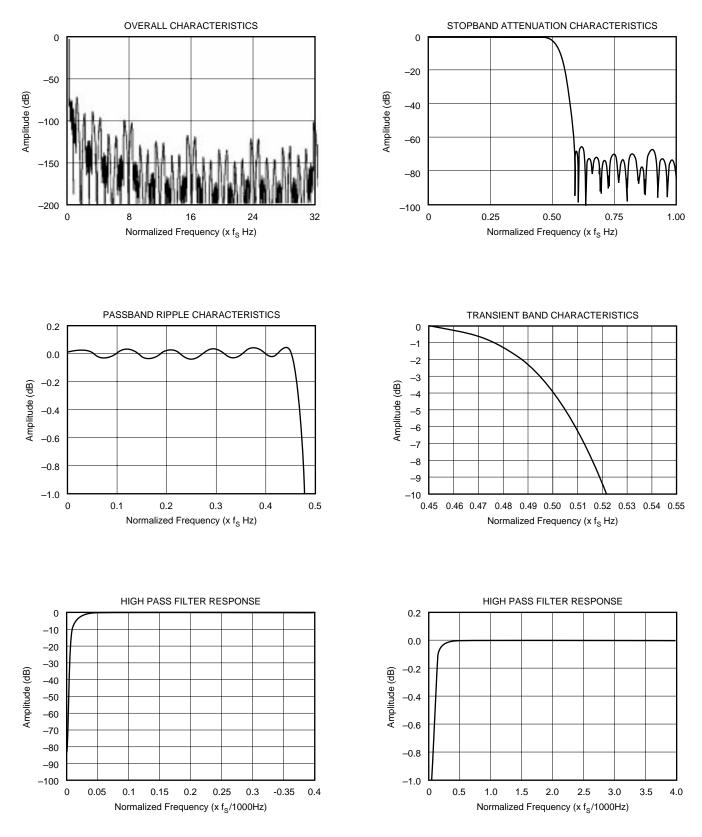


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At  $T_A = +25^{\circ}C$ ,  $+V_{DD} = +V_{CC} = +5V$ ,  $f_S = 44.1$ kHz, and SYSCLK =  $384f_S$ , unless otherwise noted.

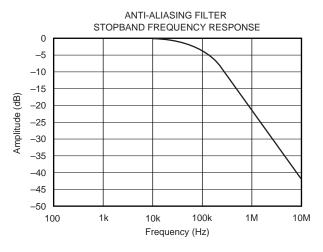
#### **DIGITAL FILTER**

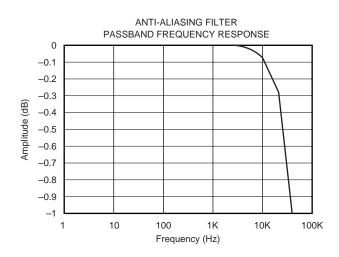




At  $T_A = +25^{\circ}C$ ,  $+V_{DD} = +V_{CC} = +5V$ ,  $f_S = 44.1 \text{kHz}$ , and SYSCLK =  $384f_S$ , unless otherwise noted.

#### ANTI-ALIASING







## THEORY OF OPERATION

PCM1801 consists of a bandgap reference, two channels of a single-to-differential converter, a fully differential 5thorder delta-sigma modulator, a decimation filter (including digital high pass), and a serial interface circuit. The Block Diagram illustrates the total architecture of PCM1801, the Analog Front-End diagram illustrates the architecture of the single-to-differential converter, and the anti-aliasing filter is illustrated in the Block Diagram. Figure 1 illustrates the architecture of the 5th-order delta-sigma modulator and transfer functions.

An internal high precision reference with two external capacitors provides all reference voltages which are required by the converter, and defines the full-scale voltage range of both channels. The internal single-ended to differential voltage converter saves the design, space and extra parts needed for external circuitry required by many delta-sigma converters. The internal full differential architecture provides a wide dynamic range and excellent power supply rejection performance.

The input signal is sampled at 64x oversampling rate, eliminating the need for a sample-and-hold circuit, and simplifying anti-alias filtering requirements. The 5th-order delta-sigma noise shaper consists of five integrators which use a switched-capacitor topology, a comparator and a feedback loop consisting of a 1-bit DAC. The delta-sigma modulator shapes the quantization noise, shifting it out of the audio band in the frequency domain. The high order of the modulator enables it to randomize the modulator outputs, reducing idle tone levels. The  $64f_S$ , 1-bit stream from the modulator is converted to  $1f_S$ , 16-bit digital data by the decimation filter, which also acts as a low-pass filter to remove the shaped quantization noise. The DC components are removed by a high-pass filter, and the filtered output is converted to time-multiplexed serial signals through a serial interface which provides flexible serial formats and Master/Slave Modes.

#### SYSTEM CLOCK

The system clock for PCM1801 must be either  $256f_S$ ,  $384f_S$ , or  $512f_S$ , where  $f_S$  is the audio sampling frequency. The system clock must be supplied on SCKI (pin 5).

PCM1801 also has a system clock detection circuit which automatically senses if the system clock is operating at  $256f_{s}$ ,  $384f_{s}$ , or  $512f_{s}$ .

When  $384f_S$  and  $512f_S$  system clock are used, the PCM1801 automatically divides these clocks down to  $256f_S$  internally. This  $256f_S$  clock is used to operate the digital filter and the modulator. Table I lists the relationship of typical sampling frequencies and system clock frequencies. Figure 2 illustrates the system clock timing.

SAMPLING RATE FREQUENCY	SYSTEM CLOCK FREQUENCY (MHz)			
(kHz)	256f <sub>S</sub>	384f <sub>S</sub>	512f <sub>S</sub>	
32	8.1920	12.2880	16.3840	
44.1	11.2896	16.9340	22.5792	
48	12.2880	18.4320	24.5760	

TABLE I. System Clock Frequencies.

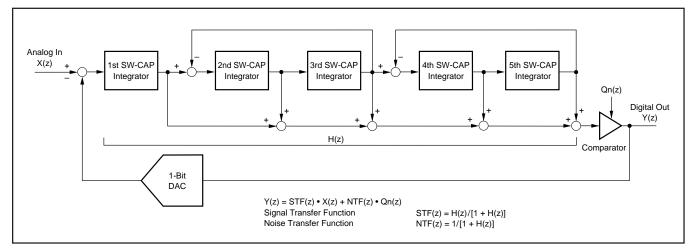


FIGURE 1. Simplified Diagram of the PCM1801 5th-Order Delta-Sigma Modulator.

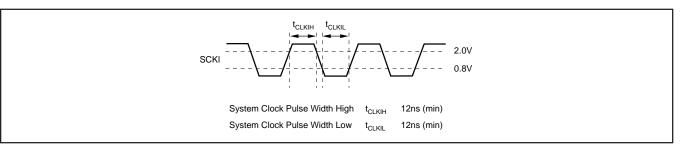


FIGURE 2. System Clock Timing.

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#### RESET

PCM1801 has an internal power-on reset circuit, which initializes (resets) when the supply voltage ( $V_{CC}/V_{DD}$ ) exceeds 4.0V (typ). The PCM1801 stays in the reset state and the digital output is forced to zero. The digital output is valid after reset state release and 18436f<sub>S</sub> periods. During reset, the logic circuits and the digital filter stop operating. Figure 3 illustrates the internal power-on reset and external reset timing.

## SERIAL AUDIO DATA INTERFACE

The PCM1801 interfaces the audio system through BCK (pin 6), LRCK (pin 7), and DOUT (pin 8).

#### DATA FORMAT

PCM1801 supports four audio data formats in both Master and Slave Modes, and are selected by FMT (pin 10) as shown in Table II.

FMT	DATA FORMAT	
0 (L)	16-Bit, Left-Justified	
1 (H)	16-Bit, I <sup>2</sup> S	

TABLE II. Data Format.

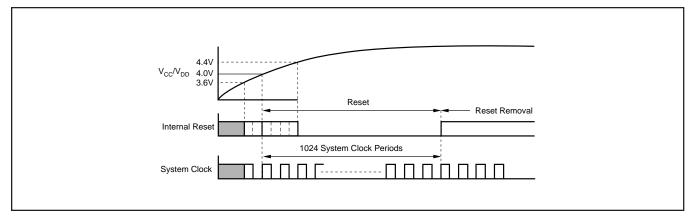


FIGURE 3. Internal Power-On Reset Timing.

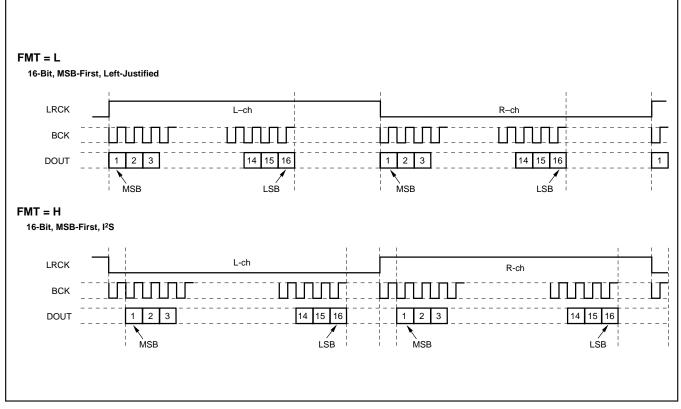


FIGURE 4. Audio Data Format (Slave Mode: LRCK, and BCK are inputs).



# SYNCHRONIZATION WITH DIGITAL AUDIO SYSTEM

PCM1801 operates with LRCK synchronized to the system clock (SCKI). PCM1801 does not require a specific phase relationship between LRCK and SCKI, but does require the synchronization of LRCK and SCKI. If the relationship between LRCK and SCKI changes more than 6 bit clocks (BCK) during one sample period due to LRCK or SCKI jitter, internal operation of the ADC halts within  $1/f_s$  and digital output is forced into BPZ code until resynchronization between LRCK and SCKI is completed. In case of changes less than 5 bit clocks (BCK), resynchronization does not occur and above digital output control and discontinuity does not occur.

#### ADC DATA OUTPUT AT RESET

Figures 6 and 7 illustrate the ADC digital output when the reset operation is done and synchronization is lost. During undefined data, it may generate some noise in the audio signal. Also, the transition of normal to undefined data and undefined or zero data to normal makes a discontinuity of data on the digital output, and may generate some noise in the audio signal.

### BOARD DESIGN AND LAYOUT CONSIDERATIONS

#### V<sub>CC</sub>, V<sub>DD</sub> PINS

The digital and analog power supply lines to the PCM1801 should be bypassed to the corresponding ground pins with both  $0.1\mu$ F ceramic and  $10\mu$ F tantalum capacitors as close to the pins as possible to maximize the dynamic performance of the ADC. Although PCM1801 has two power lines to maximize the potential of dynamic performance, using one common power

supply is recommended to avoid unexpected power supply problems, such as latch-up or power supply sequence.

#### AGND, DGND PINS

To maximize the dynamic performance of the PCM1801, the analog and digital grounds are not internally connected. These points should have very low impedance to avoid digital noise feedback into the analog ground. They should be connected directly to each other under the parts to reduce potential noise problems.

#### V<sub>IN</sub> PINS

A  $1.0\mu$ F tantalum capacitor is recommended as an ACcoupling capacitor which establishes a 5.3Hz cut-off frequency. If a higher full-scale input voltage is required, the input voltage range can be increased by adding a series resistor to the V<sub>IN</sub> pins.

#### V<sub>REF</sub> INPUTS

A 4.7 $\mu$ F tantalum capacitor is recommended between V<sub>REF</sub>1 and V<sub>REF</sub>2 to ensure low source impedance for the ADC's references. These capacitors should be located as close as possible to the V<sub>REF</sub>1 or V<sub>REF</sub>2 pin to reduce dynamic errors on the ADC's references.

#### SYSTEM CLOCK

The quality of the system clock can influence dynamic performance in the PCM1801. The duty cycle, jitter, and threshold voltage at the system clock input pin must be carefully managed. When power is supplied to the part, the system clock, bit clock (BCK), and a word clock (LRCK) should also be supplied simultaneously. Failure to supply the audio clocks will result in a power dissipation increase of up to three times normal dissipation and may degrade long-term reliability if the maximum power dissipation limit is exceeded.

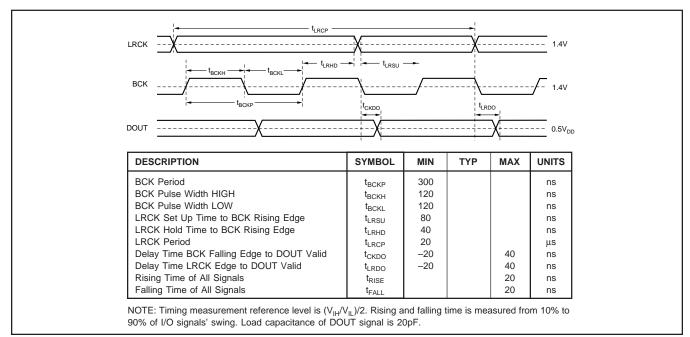


FIGURE 5. Audio Data Interface Timing (LRCK and BCK are inputs).

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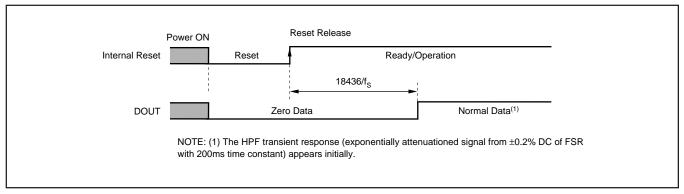
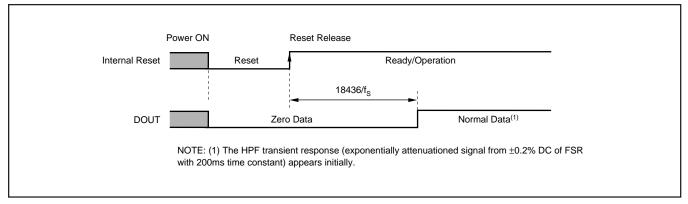


FIGURE 6. ADC Output for Power-On Reset and RSTB Control.





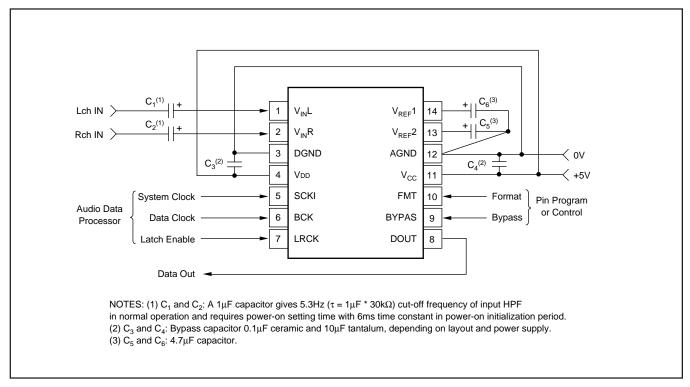


FIGURE 8. Typical Circuit Connection.