

General Description

The AAT3242 is a dual low dropout linear regulator with Power OK (POK) outputs. Two integrated regulators provide a high power 300mA output and a lower power 150mA output, making this device ideal for use with microprocessors and DSP cores in portable products. Two POK pins provide open drain output signals when their respective regulator output is within regulation. The AAT3242 has independent input voltage and enable pins for increased design flexibility. This device features a very low quiescent current (140µA typical) and low dropout voltages (typically 200mV and 400mV at the full output current level), making it ideal for portable applications where extended battery life is critical. The AAT3242 has complete over-current/short-circuit and over-temperature protection circuits to guard against extreme operating conditions.

The AAT3242 is available in a space-saving, Pbfree, 12-pin TSOPJW package. This device is capable of operation over the -40°C to +85°C temperature range.

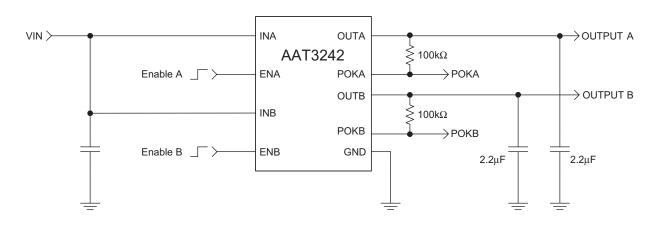
Features

PowerLinear[™]

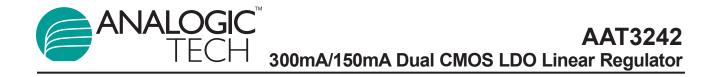
- High/Low Current Outputs, 300mA/150mA
- Low Dropout:
 - LDO A: 400mV at 300mA
 - LDO B: 200mV at 150mA
- High Output Voltage Accuracy: ±1.5%
- High PSRR: 65dB at 1kHz
- 70µA Quiescent Current for Each LDO
- Over-Current/Short-Circuit Protection
- Over-Temperature Protection
- Two POK Outputs
- Independent Power and Enable Inputs
- Uses Low Equivalent Series Resistance
 (ESR) Ceramic Capacitors
- 12-Pin TSOPJW Package
- -40°C to +85°C Temperature Range

Applications

- Cellular Phones
- Digital Cameras
- Handheld Instruments
- Microprocessor / DSP Core / I/O Power
- Notebook Computers
- PDAs and Handheld Computers
- Portable Communication Devices



Typical Application



Pin Descriptions

Pin #	Symbol	Function		
1	ENA	Enable Regulator A pin; this pin should not be left floating. When pulled low, the PMOS pass transistor turns off and the device enters shutdown mode, consuming less than 1μ A.		
2, 3, 8, 9	GND	Ground connection pins.		
4	POKA	Power OK pin with open drain output. It is pulled low when the OUTA pin is below the 10% regulation window.		
5	OUTB	Low current (150mA) regulator output pin; should be decoupled with a 2.2μ F or greater output low-ESR ceramic capacitor.		
6	INB	Input voltage pin for Regulator B; should be decoupled with $1\mu F$ or greater capacitor.		
7	ENB	Enable Regulator B; this pin should not be left floating. When pulled low, the PMOS pass transistor turns off and the device enters shutdown mode, consuming less than 1μ A.		
10	РОКВ	Power OK pin with open drain output. It is pulled low when the OUTB pin is below the 10% regulation window.		
11	OUTA	High-current (300mA) regulator output pin; should be decoupled with a 2.2µF or greater output low-ESR ceramic capacitor.		
12	INA	Input voltage pin for Regulator A; should be decoupled with 1µF or greater capacitor.		

Pin Configuration

TSOPJW-12 (Top View)

ENA 💷	0	12 INA
GND 2		
GND 🖪		
POKA 4		🧿 GND
OUTB 5		🔊 GND
INB 📑		7 ENB



Absolute Maximum Ratings¹

Symbol	Description	Value	Units
V _{IN}	Input Voltage	6.0	V
V _{ENIN(MAX)}	Maximum EN to Input Voltage	0.3	V
I _{OUT} ²	DC Output Current	$P_D/(V_{IN} - V_O)$	mA
TJ	Operating Junction Temperature Range	-40 to 150	°C
T _{LEAD}	Maximum Soldering Temperature (at leads, 10 sec)	300	°C

Thermal Information

Symbol	Description	Value	Units
θ _{JA}	Thermal Resistance ³	110	°C/W
P _D	Maximum Power Dissipation (T _A = 25°C) ⁴	909	mW

Stresses above those listed in Absolute Maximum Ratings may cause permanent damage to the device. Functional operation at conditions other than the operating conditions specified is not implied. Only one Absolute Maximum Rating should be applied at any one time.
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^{2.} Based on long-term current density limitation.

^{3.} Mounted on an FR4 board.

^{4.} Derate 9.1mW/°C above 25°C.



Electrical Characteristics¹

 $V_{\text{IN}} = V_{\text{OUT}(\text{NOM})} + 1.0V \text{ for } V_{\text{OUT}} \text{ options greater than } 1.5V. \quad V_{\text{IN}} = 2.5V \text{ for } V_{\text{OUT}} \leq 1.5V. \quad I_{\text{OUT}} = 1.0\text{mA}, \quad C_{\text{OUT}} = 2.2\mu\text{F}, \quad C_{\text{IN}} = 1.0\mu\text{F}, \quad T_{\text{A}} = -40^{\circ}\text{C} \text{ to } +85^{\circ}\text{C}, \text{ unless otherwise noted.} \quad \text{Typical values are } T_{\text{A}} = 25^{\circ}\text{C}.$

Symbol	Description	Conditions	S	Min	Тур	Max	Units
LDO A; I _{OUT} =	= 300mA			I			
V _{OUT}	Output Voltage Tolerance	I _{OUT} = 1mA to 300mA	$T_A = 25^{\circ}C$ $T_A = -40 \text{ to } 85^{\circ}C$	-1.5 -2.5		1.5 2.5	%
V _{IN}	Input Voltage		X	$V_{OUT} + V_{DO}^{5}$		5.5	V
V _{DO}	Dropout Voltage ^{2, 3}	I _{OUT} = 300m/	4		400	600	mV
ΔV _{OUT} / V _{OUT} *ΔV _{IN}	Line Regulation⁴	$V_{IN} = V_{OUT} +$				0.09	%/V
$\Delta V_{OUT(Line)}$	Dynamic Line Regulation	I _{OUT} = 300m/ to V _{OUT} + 2,	A, V _{IN} = V _{OUT} + 1 T _R /T _F = 2µS		5.0		mV
$\Delta V_{OUT(\text{Load})}$	Dynamic Load Regulation	$I_{OUT} = 1 \text{mA to}$ $T_R < 5 \mu \text{S}$			60		mV
V _{EN(L)}	Enable Threshold Low					0.6	V
V _{EN(H)}	Enable Threshold High			1.5			V
V _{POK}	Power OK Trip Threshold	V _{OUT} Rising,	T _A = 25°C	90	94	98	% of V_{OUT}
V _{POKHYS}	Power OK Hysteresis				1.0		% of V _{OUT}
V _{POK(LO)}	Power OK Output Voltage Low	I _{SINK} = 1mA				0.4	V
I _{POK}	POK Output Leakage Current	V_{POK} < 5.5V, V_{OUT} in Regulation				1.0	μA
I _{OUT}	Output Current	V _{OUT} > 1.2V		300			mA
I _{SC}	Short-Circuit Current	V _{OUT} < 0.4V			600		mA
IQ	Ground Current	$V_{IN} = 5V$, No Load; EN A = V_{IN}			70	125	μA
I _{SD}	Shutdown Current	V _{IN} = 5V, EN	A = 0V			1.0	μA
			1kHz		65		
PSRR	Power Supply Rejection Ratio	I _{OUT} =1 0mA	10kHz		45		dB
		1MHz			42		
T _{SD}	Over-Temperature Shutdown Threshold				145		C°
T _{HYS}	Over-Temperature Shutdown Hysteresis				12		C°
e _N	Output Noise	e _{NBW} = 300H	e _{NBW} = 300Hz to 50kHz		250		μVRMS
T _C	Output Voltage Temperature Coefficient				22		ppm/°C

^{1.} The AAT3242 is guaranteed to meet performance specifications over the -40°C to +85°C operating temperature range and is assured by design, characterization, and correlation with statistical process controls.

^{2.} V_{DO} is defined as V_{IN} - V_{OUT} when V_{OUT} is 98% of nominal.

^{3.} For $V_{\rm OUT}$ < 2.1V, $V_{\rm DO}$ = 2.5 - $V_{\rm OUT}.$

^{4.} C_{IN} = 10μF.

^{5.} To calculate minimum input voltage, use the following equation: $V_{IN(MIN)} = V_{OUT(MAX)} + V_{DO(MAX)}$ as long as $V_{IN} \ge 2.5V$.



Electrical Characteristics¹ (continued)

 $V_{IN} = V_{OUT(NOM)} + 1.0V$ for V_{OUT} options greater than 1.5V. $V_{IN} = 2.5V$ for $V_{OUT} \le 1.5V$. $I_{OUT} = 1.0$ mA, $C_{OUT} = 2.2\mu$ F, $C_{IN} = 1.0\mu$ F, $T_A = -40^{\circ}$ C to +85°C, unless otherwise noted. Typical values are $T_A = 25^{\circ}$ C.

Symbol	Description	Conditions		Min	Тур	Max	Units	
LDO B; I _{OUT} =	= 150mA							
V _{OUT}	Output Voltage Tolerance	I _{OUT} = 1mA to 150mA	$T_A = 25^{\circ}C$ $T_A = -40 \text{ to } 85^{\circ}C$	-1.5 -2.5		1.5 2.5	%	
V _{IN}	Input Voltage			$V_{OUT} + V_{DO}^{5}$		5.5	V	
V _{DO}	Dropout Voltage ^{2, 3}	I _{OUT} = 150mA		001 00	200	300	mV	
ΔV _{OUT} / V _{OUT} *ΔV _{IN}	Line Regulation⁴	$V_{IN} = V_{OUT} + T$				0.09	%/V	
$\Delta V_{OUT(Line)}$	Dynamic Line Regulation	I _{OUT} = 150mA to V _{OUT} + 2, T	$V_{\rm IN} = V_{\rm OUT} + 1$ $T_{\rm R}/T_{\rm F} = 2\mu {\rm S}$		5.0		mV	
$\Delta V_{OUT(Load)}$	Dynamic Load Regulation	I _{OUT} = 1mA to	150mA, T _R <5µS		60		mV	
V _{EN(L)}	Enable Threshold Low					0.6	V	
V _{EN(H)}	Enable Threshold High			1.5			V	
V _{POK}	Power OK Trip Threshold	V _{OUT} Rising, 1	Г _А = 25°С	90	94	98	% of V _{OUT}	
V _{POKHYS}	Power OK Hysteresis				1.0		% of V _{OUT}	
V _{POK(LO)}	Power OK Output Voltage Low	I _{SINK} = 1mA				0.4	V	
I _{POK}	POK Output Leakage Current	V _{POK} < 5.5V, V	V _{OUT} in Regulation			1.0	μA	
I _{OUT}	Output Current	V _{OUT} > 1.2V		150			mA	
I _{SC}	Short-Circuit Current	V _{OUT} < 0.4V			600		mA	
ا _Q	Ground Current	V _{IN} = 5V, No L	_oad; EN B = V _{IN}		70	125	μA	
	Power Supply Rejection Ratio		1kHz		65			
PSRR		I _{OUT} = 10mA	10kHz		45		dB	
			1MHz		42			
T _{SD}	Over-Temperature Shutdown Threshold				145		°C	
T _{HYS}	Over-Temperature Shutdown Hysteresis				12		°C	
e _N	Output Noise	e _{NBW} = 300Hz to 50kHz			250		μVRMS	
T _c	Output Voltage Temperature Coefficient				22		ppm/°C	

2. V_{DO} is defined as V_{IN} - V_{OUT} when V_{OUT} is 98% of nominal.

3. For $\rm V_{OUT}$ < 2.3V, $\rm V_{DO}$ = 2.5 - $\rm V_{OUT}.$

4. $C_{IN} = 10 \mu F.$

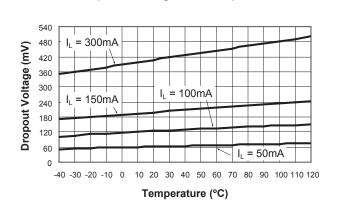
^{1.} The AAT3242 is guaranteed to meet performance specifications over the -40°C to +85°C operating temperature range and is assured by design, characterization, and correlation with statistical process controls.

^{5.} To calculate minimum input voltage, use the following equation: $V_{IN(MIN)} = V_{OUT(MAX)} + V_{DO(MAX)}$ as long as $V_{IN} \ge 2.5V$.



Typical Characteristics

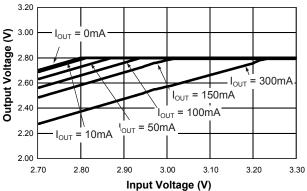
Unless otherwise noted, $V_{IN} = 5V$, $T_A = 25^{\circ}C$.



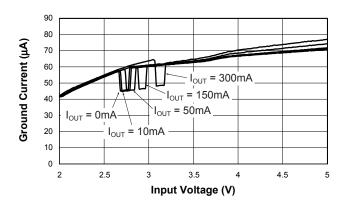
Dropout Voltage vs. Temperature

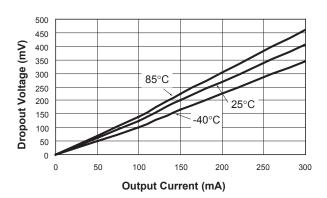
Dropout Voltage vs. Output Current

Dropout Characteristics

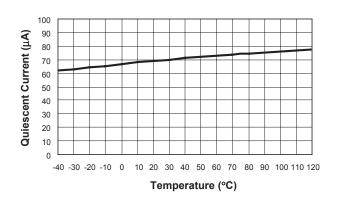


Ground Current vs. Input Voltage

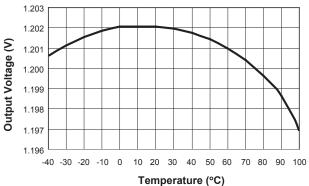




Quiescent Current vs. Temperature



Output Voltage vs. Temperature

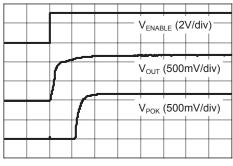




Typical Characteristics

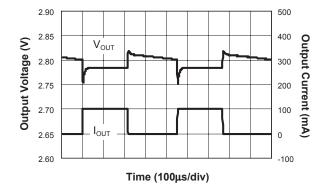
Unless otherwise noted, $V_{IN} = 5V$, $T_A = 25^{\circ}C$.

Turn-On Time and POK Delay

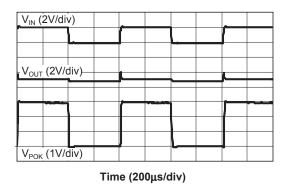


Time (10µs/div)

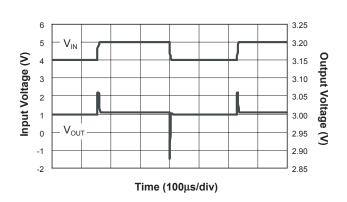
Load Transient Response



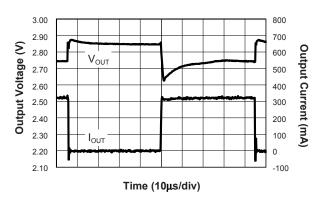
POK Output Response

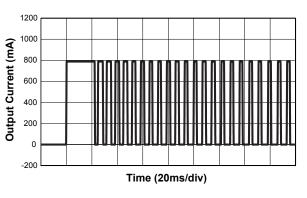


Line Transient Response

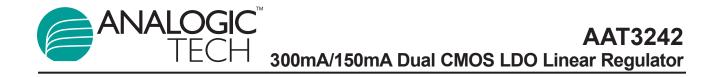


Load Transient Response 300mA



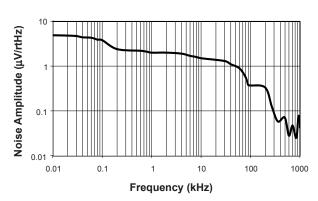


Over-Current Protection



 $\frac{\text{Typical Characteristics}}{\text{Unless otherwise noted, V}_{\text{IN}} = 5\text{V}, \text{T}_{\text{A}} = 25^{\circ}\text{C}.}$

Self Noise



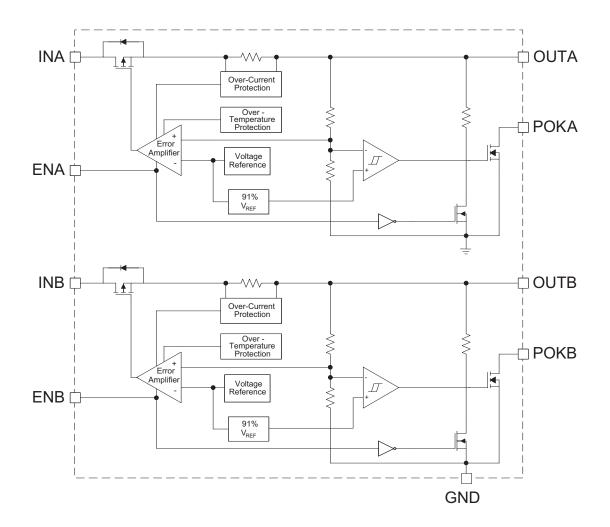
1.250 1.225 1.200 $V_{\text{EN(H)}}$ 1.175 V_{EN} (V) 1.150 1.125 V_{EN(L)} -1.100 1.075 1.050 3.0 3.5 4.0 4.5 5.0 5. 2.5 Input Voltage (V)

 $V_{\text{EN(H)}}$ and $V_{\text{EN(L)}}$ vs. V_{IN}

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Functional Block Diagram



Functional Description

The AAT3242 is a high performance dual LDO regulator with two Power OK pins. The first regulator (A) sources 300mA of current, while the second (B) regulator can deliver 150mA. Each regulator has an integrated Power OK comparator which indicates when the respective output is out of regulation. The POK pins are open drain outputs, and they are held low when the respective regulator is in shutdown mode. The device has independent enable pins to shut down each LDO regulator for power conservation in portable products. Forcing EN A/B low (<0.6V) powers down the regulators and draws a maximum of 1.0 μ A. The AAT3242 has short-circuit and thermal protection in case of adverse operating conditions. Device power dissipation is limited to the package type and thermal dissipation properties. Refer to the Thermal Considerations section of this datasheet for details on device operation at maximum output current loads.



Applications Information

To assure the maximum possible performance is obtained from the AAT3242, please refer to the following application recommendations.

Input Capacitor

A 1µF or larger capacitor is typically recommended for C_{IN} in most applications. A C_{IN} capacitor is not required for basic LDO regulator operation; however, if the AAT3242 is physically located more than three centimeters from an input power source, a C_{IN} capacitor will be needed for stable operation. C_{IN} should be located as closely to the device V_{IN} pin as practically possible. C_{IN} values greater than 1µF will offer superior input line transient response and will assist in maximizing the highest possible power supply ripple rejection.

Ceramic, tantalum, or aluminum electrolytic capacitors may be selected for C_{IN} . There is no specific capacitor ESR requirement for C_{IN} ; however, for 300mA LDO regulator output operation, ceramic capacitors are recommended for C_{IN} due to their inherent capability over tantalum capacitors to withstand input current surges from low impedance sources such as batteries in portable devices.

Output Capacitor

For proper load voltage regulation and operational stability, a capacitor is required between pins V_{OUT} and GND. The C_{OUT} capacitor connection to the LDO regulator ground pin should be made as direct as practically possible for maximum device performance. The AAT3242 has been specifically designed to function with very low ESR ceramic capacitors. For best performance, ceramic capacitors are recommended.

Typical output capacitor values for maximum output current conditions range from $1\mu F$ to $10\mu F.$

Applications utilizing the exceptionally low output noise and optimum power supply ripple rejection characteristics of the AAT3242 should use 2.2μ F or

greater for C_{OUT} . If desired, C_{OUT} may be increased without limit. In low output current applications where output load is less than 10mA, the minimum value for C_{OUT} can be as low as 0.47µF.

Capacitor Characteristics

Ceramic composition capacitors are highly recommended over all other types of capacitors for use with the AAT3242. Ceramic capacitors offer many advantages over their tantalum and aluminum electrolytic counterparts. A ceramic capacitor typically has very low ESR, is lower cost, has a smaller PCB footprint, and is non-polarized. Line and load transient response of the LDO regulator is improved by using low ESR ceramic capacitors. Since ceramic capacitors are non-polarized, they are not prone to incorrect connection damage.

Equivalent Series Resistance

ESR is a very important characteristic to consider when selecting a capacitor. ESR is the internal series resistance associated with a capacitor that includes lead resistance, internal connections, size and area, material composition, and ambient temperature. Typically, capacitor ESR is measured in milliohms for ceramic capacitors and can range to more than several ohms for tantalum or aluminum electrolytic capacitors.

Ceramic Capacitor Materials

Ceramic capacitors less than 0.1μ F are typically made from NPO or C0G materials. NPO and C0G materials generally have tight tolerance and are very stable over temperature. Larger capacitor values are usually composed of X7R, X5R, Z5U, or Y5V dielectric materials. These two material types are not recommended for use with LDO regulators since the capacitor tolerance can vary more than ±50% over the operating temperature range of the device. A 2.2µF Y5V capacitor could be reduced to 1µF over temperature; this could cause problems for circuit operation. X7R and X5R dielectrics are much more desirable. The temperature tolerance



of X7R dielectric is better than ±15%. Capacitor area is another contributor to ESR. Capacitors which are physically large in size will have a lower ESR when compared to a smaller sized capacitor of an equivalent material and capacitance value. These larger devices can improve circuit transient response when compared to an equal value capacitor in a smaller package size. Consult capacitor vendor datasheets carefully when selecting capacitors for LDO regulators.

POK Output

The AAT3242 features integrated Power OK comparators which can be used as an error flag. The POK open drain output goes low when output voltage is 6% (typ) below its nominal regulation voltage. Additionally, any time one of the regulators is in shutdown, the respective POK output is pulled low. Connect a pull-up resistor from POKA to OUTA, and POKB to OUTB.

Enable Function

The AAT3242 features an LDO regulator enable/disable function. Each LDO has its own dedicated enable pin. These pins (EN) are active high and are compatible with CMOS logic. To assure the LDO regulators will switch on, ENA/B must be greater than 1.6V. The LDO regulators will shut down when the voltage on the ENA/B pins falls below 0.6V. In shutdown, the AAT3242 will consume less than 1.0 μ A of current. If the enable function is not needed in a specific application, it may be tied to V_{IN} to keep the LDO regulator in a continuously on state.

When the LDO regulators are in shutdown mode, an internal 20Ω resistor is connected between V_{OUT} and GND. This is intended to discharge C_{OUT} when the LDO regulators are disabled. The internal 20Ω has no adverse effects on device turn-on time.

Short-Circuit Protection

The AAT3242 contains internal short-circuit protection that will trigger when the output load current exceeds the internal threshold limit. Under shortcircuit conditions, the output of the LDO regulator will be current limited until the short-circuit condition is removed from the output or LDO regulator package power dissipation exceeds the device thermal limit.

Thermal Protection

The AAT3242 has an internal thermal protection circuit which will turn on when the device die temperature exceeds 145°C. The LDO regulator output will remain in a shutdown state until the internal die temperature falls back below the 145°C trip point. The combination and interaction between the shortcircuit and thermal protection systems allows the LDO regulators to withstand indefinite short-circuit conditions without sustaining permanent damage.

No-Load Stability

The AAT3242 is designed to maintain output voltage regulation and stability under operational noload conditions. This is an important characteristic for applications where the output current may drop to zero.

Reverse Output-to-Input Voltage Conditions and Protection

Under normal operating conditions, a parasitic diode exists between the output and input of the LDO regulator. The input voltage should always remain greater than the output load voltage maintaining a reverse bias on the internal parasitic diode. Conditions where V_{OUT} might exceed V_{IN} should be avoided since this would forward bias the internal parasitic diode and allow excessive current flow into the V_{OUT} pin, possibly damaging the LDO regulator. In applications where there is a possibility of V_{OUT} exceeding V_{IN} for brief amounts of time during normal operation, the use of a larger value C_{IN} capacitor is highly recommended. A larger value of C_{IN} with respect to C_{OUT} will effect a slower C_{IN} decay rate during shutdown, thus preventing V_{OUT} from exceeding V_{IN} . In applications where there is a greater danger of V_{OUT} exceeding V_{IN} for extended periods of time, it is recommended to place a Schottky diode across V_{IN} to V_{OUT} (connecting the cathode to V_{IN} and anode to V_{OUT}). The Schottky diode forward voltage should be less than 0.45V.



Thermal Considerations and High Output Current Applications

The AAT3242 is designed to deliver continuous output load currents of 300mA and 150mA under normal operations, and can supply up to 500mA during circuit start-up conditions. This is desirable for circuit applications where there might be a brief high in-rush current during a power-on event.

The limiting characteristic for the maximum output load current safe operating area is essentially package power dissipation and the internal preset thermal limit of the device. In order to obtain high operating currents, careful device layout and circuit operating conditions need to be taken into account.

The following discussions will assume the LDO regulator is mounted on a printed circuit board utilizing the minimum recommended footprint as stated in the layout considerations section of this document. At any given ambient temperature (T_A), the maximum package power dissipation can be determined by the following equation:

$$\mathsf{P}_{\mathsf{D}(\mathsf{MAX})} = \frac{\mathsf{T}_{\mathsf{J}(\mathsf{MAX})} - \mathsf{T}_{\mathsf{A}}}{\theta_{\mathsf{J}\mathsf{A}}}$$

Constants for the AAT3242 are $T_{J(MAX)}$ (the maximum junction temperature for the device, which is 125°C) and $\theta_{JA} = 110^{\circ}$ C/W (the package thermal resistance). Typically, maximum conditions are calculated at the maximum operating temperature of $T_A = 85^{\circ}$ C and under normal ambient conditions where $T_A = 25^{\circ}$ C. Given $T_A = 85^{\circ}$ C, the maximum package power dissipation is 364mW. At $T_A = 25^{\circ}$ C, the maximum package power dissipation is 909mW.

The maximum continuous output current for the AAT3242 is a function of the package power dissipation and the input-to-output voltage drop across the LDO regulator. To determine the maximum output current for a given output voltage, refer to the following equation. This calculation accounts for the total power dissipation of the LDO regulator, including that caused by ground current.

$$\mathsf{P}_{\mathsf{D}(\mathsf{MAX})} = [(\mathsf{V}_{\mathsf{IN}} - \mathsf{V}_{\mathsf{OUTA}})\mathsf{I}_{\mathsf{OUTA}} + (\mathsf{V}_{\mathsf{IN}} \times \mathsf{I}_{\mathsf{GND}})] + [(\mathsf{V}_{\mathsf{IN}} - \mathsf{V}_{\mathsf{OUTB}})\mathsf{I}_{\mathsf{OUTB}} + (\mathsf{V}_{\mathsf{IN}} \times \mathsf{I}_{\mathsf{GND}})]$$

This formula can be solved for I_{OUTA} to determine the maximum output current for LDOA:

$$I_{OUTA(MAX)} = \frac{P_{D(MAX)} - (2 \times V_{IN} \times I_{GND}) - (V_{IN} - V_{OUTB}) \times I_{OUTB}}{V_{IN} - V_{OUTA}}$$



The following is an example for a 2.5V output:

V _{OUTA}	= 2.5V
V _{OUTB}	= 1.5V
I _{OUTB}	= 150mA
$V_{\rm IN}$	= 4.2V
I _{GND}	= 125μΑ
I _{OUTA(MAX)}	$= \frac{909\text{mW} - (2 \times 4.2\text{V} \times 125\mu\text{A}) - (4.2 - 1.5) \times 150\text{mA}}{4.2 - 2.5}$
I _{OUTA(MAX)}	= 296mA

From the discussion above, $P_{D(MAX)}$ was determined to equal 909mW at $T_A = 25^{\circ}C$.

Therefore, with Regulator B delivering 150mA at 1.5V, Regulator A can sustain a constant 2.5V output at a 296mA load current at an ambient temperature of 25°C. Higher input-to-output voltage differentials can be obtained with the AAT3242, while maintaining device functions within the thermal safe operating area. To accomplish this, the device thermal resistance must be reduced by increasing the heat sink area or by oper-ating the LDO regulator in a duty-cycled mode.

For example, an application requires $V_{IN} = 4.2V$ while $V_{OUT} = 1.5V$ at a 500mA load and $T_A = 25^{\circ}C$. To maintain this high input voltage and output current level, the LDO regulator must be operated in a duty-cycled mode. Refer to the following calculation for duty-cycle operation:

$$\begin{split} I_{GND} &= 125 \mu A \\ I_{OUT} &= 500 \text{mA} \\ V_{IN} &= 4.2 \text{V} \\ V_{OUT} &= 1.5 \text{V} \\ \% \text{DC} &= \frac{100(\text{P}_{\text{D}(\text{MAX})})}{[(\text{V}_{IN} - \text{V}_{\text{OUTA}})\text{I}_{\text{OUTA}} + (\text{V}_{IN} \times \text{I}_{\text{GND}})] + [(\text{V}_{IN} - \text{V}_{\text{OUTB}})\text{I}_{\text{OUTB}} + (\text{V}_{IN} \times \text{I}_{\text{GND}})]} \\ \% \text{DC} &= \frac{100(909 \text{mW})}{[(4.2 \text{V} - 1.5 \text{V})500 \text{mA} + (4.2 \text{V} \times 125 \mu \text{A})] + [(4.2 \text{V} - 1.5 \text{V})200 \text{mA} + (4.2 \text{V} \times 125 \mu \text{A})]} \\ \% \text{DC} &= 48.10\% \end{split}$$

 $\mathsf{P}_{\mathsf{D}(\mathsf{MAX})}$ is assumed to be 909mW

For a 500mA output current and a 2.7V drop across the AAT3242 at an ambient temperature of 25°C, the maximum on-time duty cycle for the device would be 48.10%.



Ordering Information

	Voltage					
Package	LDO A LDO B		LDO A LDO B		Marking ¹	Part Number (Tape and Reel) ²
TSOPJW-12	3.3V	2.5V	LSXYY	AAT3242ITP-WN-T1		
TSOPJW-12	3.3V	1.8V	PAXYY	AAT3242ITP-WI-T1		
TSOPJW-12	3.0V	2.85V	LPXYY	AAT3242ITP-TR-T1		
TSOPJW-12	3.0V	2.5V	LJXYY	AAT3242ITP-TN-T1		
TSOPJW-12	3.0V	1.8V	LHXYY	AAT3242ITP-TI-T1		
TSOPJW-12	3.0V	1.5V	NTXYY	AAT3242ITP-TG-T1		
TSOPJW-12	2.9V	1.5V	MOXYY	AAT3242ITP-SG-T1		
TSOPJW-12	2.8V	3.0V	LVXYY	AAT3242ITP-QT-T1		
TSOPJW-12	2.8V	2.8V	LDXYY	AAT3242ITP-QQ-T1		
TSOPJW-12	2.8V	2.6V	LQXYY	AAT3242ITP-QO-T1		
TSOPJW-12	2.8V	2.5V	LLXYY	AAT3242ITP-QN-T1		
TSOPJW-12	2.8V	1.9V	LRXYY	AAT3242ITP-QY-T1		
TSOPJW-12	2.8V	1.5V	MCXYY	AAT3242ITP-QG-T1		
TSOPJW-12	2.7V	2.7V	LOXYY	AAT3242ITP-PP-T1		
TSOPJW-12	2.6V	1.8V	MJXYY	AAT3242ITP-OI-T1		
TSOPJW-12	2.5V	1.8V	SGXYY	AAT3242ITP-IN-T1		
TSOPJW-12	1.8V	1.5V		AAT3242ITP-IG-T1		
TSOPJW-12	1.8V	2.7V	PZXYY	AAT3242ITP-IP-T1		
TSOPJW-12	1.8V	2.8V	RRXYY	AAT3242ITP-IQ-T1		



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Legend				
Voltage	Code			
1.5	G			
1.8	I			
1.9	Y			
2.5	N			
2.6	0			
2.7	Р			
2.8	Q			
2.85	R			
2.9	S			
3.0	Т			
3.3	W			

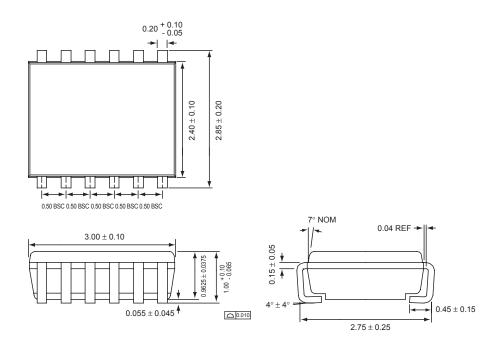
^{1.} XYY = assembly and date code.

^{2.} Sample stock is generally held on part numbers listed in **BOLD**.



Package Information

TSOPJW-12



All dimensions in millimeters.



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