

# MPC5553 Microcontroller Data Sheet

by: Microcontroller Division

This document provides electrical specifications, pin assignments, and package diagrams for the MPC5553 microcontroller device. For functional characteristics, refer to the MPC5553/MPC5554 *Microcontroller Reference Manual*.

## 1 Overview

The MPC5553 microcontroller (MCU) is a member of the MPC5500 family of microcontrollers built on the Power Architecture™ embedded technology. This family of parts contains many new features coupled with high performance CMOS technology to provide substantial reduction of cost per feature and significant performance improvement over the MPC500 family.

The host processor core of this device complies with the Power Architecture embedded category that is 100% user-mode compatible (with floating point library) with the original Power PC™ user instruction set architecture (UISA). The embedded architecture has enhancements that improve the performance in embedded applications. This core also has additional instructions, including digital signal processing (DSP) instructions, beyond the original Power PC instruction set. This family of parts

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## Overview

contains many new features coupled with high performance CMOS technology to provide significant performance improvement over the MPC565.

The MPC5553 of the MPC5500 family has two levels of memory hierarchy. The fastest accesses are to the 8-kilobyte unified cache. The next level in the hierarchy contains the 64-kilobyte on-chip internal SRAM and 1.5 Mbyte internal Flash memory. The internal SRAM and flash memory can hold instructions and data. The external bus interface has been designed to support most of the standard memories used with the MPC5xx family.

The complex input/output timer functions of the MPC5500 family are performed by an enhanced time processor unit engine (eTPU). The eTPU engine controls 32 hardware channels. The eTPU has been enhanced over the TPU by providing 24-bit timers, double-action hardware channels, variable number of parameters per channel, angle clock hardware, and additional control and arithmetic instructions. The eTPU can be programmed using a high-level programming language.

The less complex timer functions of the MPC5500 family are performed by the enhanced modular input/output system (eMIOS). The eMIOS' 24 hardware channels are capable of single-action, double-action, pulse-width modulation (PWM), and modulus-counter operations. Motor control capabilities include edge-aligned and center-aligned PWM.

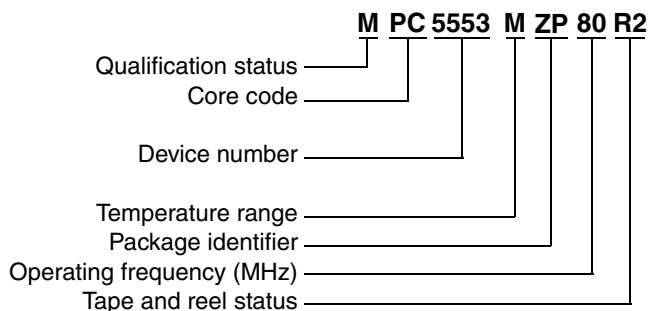
Off-chip communication is performed by a suite of serial protocols including controller area networks (FlexCANs), enhanced deserial/serial peripheral interfaces (DSPI), and enhanced serial communications interfaces (eSCIs). The DSPIs support pin reduction through hardware serialization and deserialization of timer channels and general-purpose input/output (GPIO) signals.

The MCU of the MPC5553 has an on-chip 40-channel enhanced queued dual analog-to-digital converter (eQADC).

The system integration unit (SIU) performs several chip-wide configuration functions. Pad configuration and general-purpose input and output (GPIO) are controlled from the SIU. External interrupts and reset control are also determined by the SIU. The internal multiplexer submodule (SIU\_DISR) provides multiplexing of eQADC trigger sources, daisy chaining the DSPIs, and external interrupt signal multiplexing.

The Fast Ethernet (FEC) module is a RISC-based controller that supports both 10 and 100 Mbps Ethernet/IEEE® 802.3 networks and is compatible with three different standard MAC (media access controller) PHY (physical) interfaces to connect to an external Ethernet bus. The FEC supports the 10 or 100 Mbps MII (media independent interface), and the 10 Mbps-only with a seven-wire interface, which uses a subset of the MII signals. The upper 16-bits of the 32-bit external bus interface (EBI) are used to connect to an external Ethernet device. The FEC contains built-in transmit and receive message FIFOs and DMA support.

## 2 Ordering Information



**Temperature Range**  
M = -40° C to 125° C

**Package Identifier**  
ZP = 416PBGA SnPb  
VR = 416PBGA Pb-free  
VF = 208MAPBGA SnPb  
VM = 208MAPBGA Pb-free  
ZQ = 324PBGA SnPb  
VZ = 324PBGA Pb-free

**Operating Frequency**  
80 = 80 MHz  
112 = 112 MHz  
132 = 132 MHz

**Tape and Reel Status**  
R2 = Tape and seal  
(blank) = Trays

**Qualification Status**  
P = Pre qualification  
M = Full spec qualified

**Note:** Not all options are available on all devices. Refer to [Table 1](#).

**Figure 1. MPC5500 Family Part Number Example**

**Table 1. Orderable Part Numbers**

Freescale Part Number <sup>1</sup>	Package Description	Speed (MHz)		Operating Temperature <sup>2</sup>	
		Nominal	Max <sup>3</sup> (f <sub>MAX</sub> )	Min (T <sub>L</sub> )	Max (T <sub>H</sub> )
MPC5553MVR132	MPC5553 Lead-free 416 package	132	132	-40° C	125° C
MPC5553MVR112		112	114		
MPC5553MVR80		80	82		
MPC5553MVZ132	MPC5553 Lead-free 324 package	132	132	-40° C	125° C
MPC5553MVZ112		112	114		
MPC5553MVZ80		80	82		
MPC5553MVM132	MPC5553 Lead-free 208 package	132	132	-40° C	125° C
MPC5553MVM112		112	114		
MPC5553MVM80		80	82		
MPC5553MZP132	MPC5553 Lead 416 package	132	132	-40° C	125° C
MPC5553MZP112		112	114		
MPC5553MZP80		80	82		
MPC5553MZQ132	MPC5553 Lead 324 package	132	132	-40° C	125° C
MPC5553MZQ112		112	114		
MPC5553MZQ80		80	82		

Table 1. Orderable Part Numbers (continued)

Freescale Part Number <sup>1</sup>	Package Description	Speed (MHz)		Operating Temperature <sup>2</sup>	
		Nominal	Max <sup>3</sup> (f <sub>MAX</sub> )	Min (T <sub>L</sub> )	Max (T <sub>H</sub> )
MPC5553MVF132	MPC5553 Lead 208 package	132	132	-40° C	125° C
MPC5553MVF112		112	114		
MPC5553MVF80		80	82		

<sup>1</sup> All devices are PPC5553, rather than MPC5553, until the product qualifications. Not all configurations are available in the PPC parts.

<sup>2</sup> The lowest operating temperature is referenced by T<sub>L</sub>; the highest operating temperature is referenced by T<sub>H</sub>.

<sup>3</sup> Speed is the nominal maximum frequency. Max speed is the maximum speed allowed including any frequency modulation. 80 MHz parts allow for 80 MHz + 2% modulation. However, 132 MHz devices allow 128 MHz plus two percent frequency modulation only.

### 3 Electrical Characteristics

This section contains detailed information on power considerations, DC/AC electrical characteristics, and AC timing specifications for the MCU.

#### 3.1 Maximum Ratings

Table 2. Absolute Maximum Ratings<sup>1</sup>

Spec	Characteristic	Symbol	Min	Max <sup>2</sup>	Unit
1	1.5 V core supply voltage <sup>3</sup>	V <sub>DD</sub>	-0.3	1.7	V
2	Flash program/erase voltage	V <sub>PP</sub>	-0.3	6.5	V
3	Flash core voltage	V <sub>DDF</sub>	-0.3	1.7	V
4	Flash read voltage	V <sub>FLASH</sub>	-0.3	4.6	V
5	SRAM standby voltage	V <sub>STBY</sub>	-0.3	1.7	V
6	Clock synthesizer voltage	V <sub>DDSYN</sub>	-0.3	4.6	V
7	3.3 V I/O buffer voltage	V <sub>DD33</sub>	-0.3	4.6	V
8	Voltage regulator control input voltage	V <sub>RC33</sub>	-0.3	4.6	V
9	Analog supply voltage (reference to V <sub>SSA</sub> )	V <sub>DDA</sub>	-0.3	5.5	V
10	I/O supply voltage (fast I/O pads) <sup>4</sup>	V <sub>DDE</sub>	-0.3	4.6	V
11	I/O supply voltage (slow and medium I/O pads) <sup>4</sup>	V <sub>DDEH</sub>	-0.3	6.5	V
12	DC input voltage <sup>5</sup> V <sub>DDEH</sub> powered I/O pads V <sub>DDE</sub> powered I/O pads	V <sub>IN</sub>	-1.0 <sup>6</sup> -1.0 <sup>6</sup>	6.5 <sup>7</sup> 4.6 <sup>8</sup>	V
13	Analog reference high voltage (reference to V <sub>RL</sub> )	V <sub>RH</sub>	-0.3	5.5	V
14	V <sub>SS</sub> differential voltage	V <sub>SS</sub> - V <sub>SSA</sub>	-0.1	0.1	V
15	V <sub>DD</sub> differential voltage	V <sub>DD</sub> - V <sub>DDA</sub>	-V <sub>DDA</sub>	V <sub>DD</sub>	V
16	V <sub>REF</sub> differential voltage	V <sub>RH</sub> - V <sub>RL</sub>	-0.3	5.5	V

Table 2. Absolute Maximum Ratings<sup>1</sup> (continued)

Spec	Characteristic	Symbol	Min	Max <sup>2</sup>	Unit
17	$V_{RH}$ to $V_{DDA}$ differential voltage	$V_{RH} - V_{DDA}$	-5.5	5.5	V
18	$V_{RL}$ to $V_{SSA}$ differential voltage	$V_{RL} - V_{SSA}$	-0.3	0.3	V
19	$V_{DDEH}$ to $V_{DDA}$ differential voltage	$V_{DDEH} - V_{DDA}$	$-V_{DDA}$	$V_{DDEH}$	V
20	$V_{DDF}$ to $V_{DD}$ differential voltage	$V_{DDF} - V_{DD}$	-0.3	0.3	V
21	This spec has been moved to Table 9, spec 43a.				
22	$V_{SSSYN}$ to $V_{SS}$ differential voltage	$V_{SSSYN} - V_{SS}$	-0.1	0.1	V
23	$V_{RCVSS}$ to $V_{SS}$ differential voltage	$V_{RCVSS} - V_{SS}$	-0.1	0.1	V
24	Maximum DC digital input current <sup>9</sup> (per pin, applies to all digital pins) <sup>5</sup>	$I_{MAXD}$	-2	2	mA
25	Maximum DC analog input current <sup>10</sup> (per pin, applies to all analog pins)	$I_{MAXA}$	-3	3	mA
26	Maximum operating temperature range <sup>11</sup> Die junction temperature	$T_J$	$T_L$	150.0	°C
27	Storage temperature range	$T_{STG}$	-55.0	150.0	°C
28	Maximum solder temperature <sup>12</sup>	$T_{SDR}$	—	260.0	°C
29	Moisture sensitivity level <sup>13</sup>	MSL	—	3	

<sup>1</sup> Functional operating conditions are given in the DC electrical specifications. Absolute maximum ratings are stress ratings only, and functional operation at the maxima is not guaranteed. Stress beyond the listed maxima can affect device reliability or cause permanent damage to the device.

<sup>2</sup> Absolute maximum voltages are currently maximum burn-in voltages. Absolute maximum specifications for device stress have not yet been determined.

<sup>3</sup> 1.5 V +/- 10% for proper operation. This parameter is specified at a maximum junction temperature of 150 °C.

<sup>4</sup> All functional non-supply I/O pins are clamped to  $V_{SS}$  and  $V_{DDE}$ , or  $V_{DDEH}$ .

<sup>5</sup> AC signal overshoot and undershoot of up to +/- 2.0 V of the input voltages is permitted for an accumulative duration of 60 hours over the complete lifetime of the device (injection current not limited for this duration).

<sup>6</sup> Internal structures hold the voltage greater than -1.0 V if the injection current limit of 2 mA is met. Keep the negative DC current greater than -0.6 V on eTPUB[15] and SINB during the internal power-on reset (POR) state.

<sup>7</sup> Internal structures hold the input voltage less than the maximum voltage on all pads powered by  $V_{DDEH}$  supplies, if the maximum injection current specification is met (2 mA for all pins) and  $V_{DDEH}$  is within the operating voltage specifications.

<sup>8</sup> Internal structures hold the input voltage less than the maximum voltage on all pads powered by  $V_{DDE}$  supplies, if the maximum injection current specification is met (2 mA for all pins) and  $V_{DDE}$  is within the operating voltage specifications.

<sup>9</sup> Total injection current for all pins (including both digital and analog) must not exceed 25 mA.

<sup>10</sup> Total injection current for all analog input pins must not exceed 15 mA.

<sup>11</sup> Lifetime operation at these specification limits is not guaranteed.

<sup>12</sup> Solder profile per CDF-AEC-Q100.

<sup>13</sup> Moisture sensitivity per JEDEC test method A112.

## 3.2 Thermal Characteristics

The shaded rows in the following table indicate information specific to a four-layer device.

**Table 3. Thermal Characteristics**

Spec	MPC5553 Thermal Characteristic	Symbol	Package			Unit
			208 MAPBGA	324 PBGA	416 PBGA	
1	Junction to ambient <sup>1,2</sup> , natural convection (one-layer board)	$R_{\theta JA}$	41	30	29	°C/W
2	Junction to ambient <sup>1,3</sup> , natural convection (four-layer board 2s2p)	$R_{\theta JA}$	25	21	21	°C/W
3	Junction to ambient (@200 ft./min., one-layer board)	$R_{\theta JMA}$	33	24	23	°C/W
4	Junction to ambient (@200 ft./min., four-layer board 2s2p)	$R_{\theta JMA}$	22	17	18	°C/W
5	Junction to board <sup>4</sup> (four-layer board 2s2p)	$R_{\theta JB}$	15	12	13	°C/W
6	Junction to case <sup>5</sup>	$R_{\theta JC}$	7	8	9	°C/W
7	Junction to package top <sup>6</sup> , natural convection	$\Psi_{JT}$	2	2	2	°C/W

<sup>1</sup> Junction temperature is a function of on-chip power dissipation, package thermal resistance, mounting site (board) temperature, ambient temperature, air flow, power dissipation of other components on the board, and board thermal resistance.

<sup>2</sup> Per SEMI G38-87 and JEDEC JESD51-2 with the single-layer board horizontal.

<sup>3</sup> Per JEDEC JESD51-6 with the board horizontal.

<sup>4</sup> Thermal resistance between the die and the printed circuit board per JEDEC JESD51-8. Board temperature is measured on the top surface of the board near the package.

<sup>5</sup> Indicates the average thermal resistance between the die and the case top surface as measured by the cold plate method (MIL SPEC-883 Method 1012.1) with the cold plate temperature used for the case temperature.

<sup>6</sup> Thermal characterization parameter indicating the temperature difference between package top and the junction temperature per JEDEC JESD51-2.

### 3.2.1 General Notes for Specifications at Maximum Junction Temperature

An estimation of the device junction temperature,  $T_J$ , can be obtained from the equation:

$$T_J = T_A + (R_{\theta JA} \times P_D)$$

where:

$T_A$  = ambient temperature for the package (°C)

$R_{\theta JA}$  = junction to ambient thermal resistance (°C/W)

$P_D$  = power dissipation in the package (W)

The thermal resistance values used are based on the JEDEC JESD51 series of standards to provide consistent values for estimations and comparisons. The difference between the values determined for the single-layer (1s) board compared to a four-layer board that has two signal layers, a power and a ground plane (2s2p), demonstrate that the effective thermal resistance is not a constant. The thermal resistance depends on the:

- Construction of the application board (number of planes)
- Effective size of the board which cools the component
- Quality of the thermal and electrical connections to the planes
- Power dissipated by adjacent components

Connect all the ground and power balls to the respective planes with one via per ball. Using fewer vias to connect the package to the planes reduces the thermal performance. Thinner planes also reduce the thermal performance. When the clearance between the vias leave the planes virtually disconnected, the thermal performance is also greatly reduced.

As a general rule, the value obtained on a single-layer board is within the normal range for the tightly packed printed circuit board. The value obtained on a board with the internal planes is usually within the normal range if the application board has:

- One oz (35 micron nominal thickness) internal planes
- Components are well separated
- Overall power dissipation on the board is less than  $0.02 \text{ W/cm}^2$

The thermal performance of any component depends on the power dissipation of the surrounding components. In addition, the ambient temperature varies widely within the application. For many natural convection and especially closed box applications, the board temperature at the perimeter (edge) of the package is approximately the same as the local air temperature near the device. Specifying the local ambient conditions explicitly as the board temperature provides a more precise description of the local ambient conditions that determine the temperature of the device.

At a known board temperature, the junction temperature is estimated using the following equation:

$$T_J = T_B + (R_{\theta JB} \times P_D)$$

where:

$T_J$  = junction temperature ( $^{\circ}\text{C}$ )

$T_B$  = board temperature at the package perimeter ( $^{\circ}\text{C/W}$ )

$R_{\theta JB}$  = junction-to-board thermal resistance ( $^{\circ}\text{C/W}$ ) per JESD51-8

$P_D$  = power dissipation in the package (W)

When the heat loss from the package case to the air does not factor into the calculation, an acceptable value for the junction temperature is predictable. Ensure the application board is similar to the thermal test condition, with the component soldered to a board with internal planes.

The thermal resistance is expressed as the sum of a junction-to-case thermal resistance plus a case-to-ambient thermal resistance:

$$R_{\theta JA} = R_{\theta JC} + R_{\theta CA}$$

where:

$R_{\theta JA}$  = junction-to-ambient thermal resistance ( $^{\circ}\text{C/W}$ )

$R_{\theta JC}$  = junction-to-case thermal resistance ( $^{\circ}\text{C/W}$ )

$R_{\theta CA}$  = case-to-ambient thermal resistance ( $^{\circ}\text{C/W}$ )

## Electrical Characteristics

$R_{\theta JC}$  is device related and is not affected by other factors. The thermal environment can be controlled to change the case-to-ambient thermal resistance,  $R_{\theta CA}$ . For example, change the air flow around the device, add a heat sink, change the mounting arrangement on printed circuit board, or change the thermal dissipation on the printed circuit board surrounding the device. This description is most useful for packages with heat sinks where 90% of the heat flow is through the case to the heat sink to ambient. For most packages, a better model is required.

A more accurate two-resistor thermal model can be constructed from the junction-to-board thermal resistance and the junction-to-case thermal resistance. The junction-to-case thermal resistance describes when using a heat sink or where a substantial amount of heat is dissipated from the top of the package. The junction-to-board thermal resistance describes the thermal performance when most of the heat is conducted to the printed circuit board. This model can be used to generate simple estimations and for computational fluid dynamics (CFD) thermal models.

To determine the junction temperature of the device in the application on a prototype board, use the thermal characterization parameter ( $\Psi_{JT}$ ) to determine the junction temperature by measuring the temperature at the top center of the package case using the following equation:

$$T_J = T_T + (\Psi_{JT} \times P_D)$$

where:

$T_T$  = thermocouple temperature on top of the package ( $^{\circ}\text{C}$ )

$\Psi_{JT}$  = thermal characterization parameter ( $^{\circ}\text{C}/\text{W}$ )

$P_D$  = power dissipation in the package (W)

The thermal characterization parameter is measured in compliance with the JESD51-2 specification using a 40-gauge type T thermocouple epoxied to the top center of the package case. Position the thermocouple so that the thermocouple junction rests on the package. A small amount of epoxy is placed on the thermocouple junction and approximately 1 mm of wire extending from the junction. The thermocouple wire is placed flat against the package case to avoid measurement errors caused by the cooling effects of the thermocouple wire.

## References:

Semiconductor Equipment and Materials International  
805 East Middlefield Rd.  
Mountain View, CA., 94043  
(415) 964-5111

MIL-SPEC and EIA/JESD (JEDEC) specifications are available from Global Engineering Documents at 800-854-7179 or 303-397-7956.

JEDEC specifications are available on the web at <http://www.jedec.org>.

- 1. C.E. Triplett and B. Joiner, "An Experimental Characterization of a 272 PBGA Within an Automotive Engine Controller Module," Proceedings of SemiTherm, San Diego, 1998, pp. 47–54.
- 2. G. Kromann, S. Shidore, and S. Addison, "Thermal Modeling of a PBGA for Air-Cooled Applications," Electronic Packaging and Production, pp. 53–58, March 1998.
- 3. B. Joiner and V. Adams, "Measurement and Simulation of Junction to Board Thermal Resistance and Its Application in Thermal Modeling," Proceedings of SemiTherm, San Diego, 1999, pp. 212–220.



### 3.3 Package

The MPC5553 is available in packaged form. Package options are listed in [Section 2, “Ordering Information.”](#)

Refer to [Section 4, “Mechanicals,”](#) for pinouts and package drawings.

### 3.4 EMI (Electromagnetic Interference) Characteristics

Table 4. EMI Testing Specifications<sup>1</sup>

Spec	Characteristic	Minimum	Typical	Maximum	Unit
1	Scan range	0.15	—	1000	MHz
2	Operating frequency	—	—	132	MHz
3	V <sub>DD</sub> operating voltages	—	1.5	—	V
4	V <sub>DDSYN</sub> , V <sub>RC33</sub> , V <sub>DD33</sub> , V <sub>FLASH</sub> , V <sub>DDE</sub> operating voltages	—	3.3	—	V
5	V <sub>PP</sub> , V <sub>DDEH</sub> , V <sub>DDA</sub> operating voltages	—	5.0	—	V
6	Maximum amplitude	—	—	14 <sup>2</sup> 32 <sup>3</sup>	dBuV
7	Operating temperature	—	—	25	°C

<sup>1</sup> EMI testing and I/O port waveforms per SAE J1752/3 issued 1995-03. Qualification testing was performed on the MPC5554 and applied to the MPC5500 family as generic EMI performance data.

<sup>2</sup> Measured with single-chip EMI program.

<sup>3</sup> Measured with expanded EMI program.

### 3.5 ESD Characteristics

Table 5. ESD Ratings<sup>1, 2</sup>

Characteristic	Symbol	Value	Unit
ESD for Human Body Model (HBM)		2000	V
HBM circuit description	R1	1500	Ω
	C	100	pF
ESD for Field Induced Charge Model (FDCM)		500 (all pins)	V
		750 (corner pins)	
Number of pulses per pin:			
Positive pulses (HBM)	—	1	—
Negative pulses (HBM)	—	1	—
Interval of pulses	—	1	second

<sup>1</sup> All ESD testing conforms to CDF-AEC-Q100 Stress Test Qualification for Automotive Grade Integrated Circuits.

<sup>2</sup> Device failure is defined as: if after exposure to ESD pulses, the device no longer meets the device specification requirements. Complete DC parametric and functional testing will be performed per applicable device specification at room temperature followed by hot temperature, unless specified otherwise in the device specification.

### 3.6 Voltage Regulator Controller ( $V_{RC}$ ) and Power-On Reset (POR) Electrical Specifications

Table 6. VRC/POR Electrical Specifications

Spec	Characteristic	Symbol	Min	Max	Units
1	1.5 V ( $V_{DD}$ ) POR negated (ramp up) 1.5 V ( $V_{DD}$ ) POR asserted (ramp down)	$V_{POR15}$	1.1 1.1	1.35 1.35	V
2	3.3 V ( $V_{DDSYN}$ ) POR negated (ramp up) 3.3 V ( $V_{DDSYN}$ ) POR asserted (ramp down)	$V_{POR33}$	2.0 2.0	2.85 2.85	V
3	RESET pin supply ( $V_{DDEH6}$ ) POR negated (ramp up) <sup>1</sup> RESET pin supply ( $V_{DDEH6}$ ) POR asserted (ramp down) <sup>1</sup>	$V_{POR5}$	2.0 2.0	2.85 2.85	V
4	$V_{RC33}$ voltage before the regulator controller allows the pass transistor to start turning on	$V_{TRANS\_START}$	1.0	2.0	V
5	$V_{RC33}$ voltage when the regulator controller allows the pass transistor to completely turn on <sup>2, 3</sup>	$V_{TRANS\_ON}$	2.0	2.85	V
6	$V_{RC33}$ voltage greater than the voltage at which the $V_{RC}$ keeps the 1.5 V supply in regulation <sup>4, 5</sup>	$V_{VRC33REG}$	3.0	—	V
7	Current can be sourced by $V_{RCCTL}$ – 40° C 25° C 150° C ( $T_j$ )	$I_{VRCCTL}$ <sup>6</sup>	11.0 9.0 7.5	— — —	mA mA mA
8	Voltage differential during power up such that: $V_{DD33}$ can lag $V_{DDSYN}$ or $V_{DDEH6}$ , before $V_{DDSYN}$ and $V_{DDEH6}$ reach the $V_{POR33}$ and $V_{POR5}$ minimums respectively.	$V_{DD33\_LAG}$	—	1.0	V
9	Absolute value of slew rate on power supply pins		—	50	V/ms
10	Required gain: $I_{DD} / I_{VRCCTL}$ (@ $V_{DD} = 1.35$ V, $f_{sys} = f_{MAX}$ ) <sup>5, 7</sup> – 40° C 25° C 150° C ( $T_j$ )	$BETA$ <sup>8</sup>	55.0 <sup>9</sup> 58.0 <sup>9</sup> 70.0 <sup>9</sup>	— — 500	— — —

<sup>1</sup>  $V_{IL\_S}$  (Table 9, Spec15) is guaranteed to scale with  $V_{DDEH6}$  down to  $V_{POR5}$ .

<sup>2</sup> Supply full operating current for the 1.5 V supply when the 3.3 V supply reaches this range.

<sup>3</sup> It is possible to reach the current limit during ramp up—do not treat this event as short circuit current.

<sup>4</sup> At peak current for device.

<sup>5</sup> Requires compliance with Freescale's recommended board requirements and transistor recommendations. Board signal traces/routing from the  $V_{RCCTL}$  package signal to the base of the external pass transistor and between the emitter of the pass transistor to the  $V_{DD}$  package signals must have a maximum of 100 nH inductance and minimal resistance (less than 1  $\Omega$ ).  $V_{RCCTL}$  must have a nominal 1  $\mu$ F phase compensation capacitor to ground.  $V_{DD}$  must have a 20  $\mu$ F (nominal) bulk capacitor (greater than 4  $\mu$ F over all conditions, including lifetime). Place high-frequency bypass capacitors consisting of eight 0.01  $\mu$ F, two 0.1  $\mu$ F, and one 1  $\mu$ F capacitors around the package on the  $V_{DD}$  supply signals.

<sup>6</sup>  $I_{VRCCTL}$  is measured at the following conditions:  $V_{DD} = 1.35$  V,  $V_{RC33} = 3.1$  V,  $V_{VRCCTL} = 2.2$  V.

<sup>7</sup> Values are based on  $I_{DD}$  from high-use applications as explained in the  $I_{DD}$  Electrical Specification.

<sup>8</sup>  $BETA$  is measured on a per-part basis and is calculated as  $(I_{DD} \div I_{VRCCTL})$ , and represents the worst-case external transistor  $BETA$ .

<sup>9</sup> Preliminary value. Final specification pending characterization.

### 3.7 Power-Up/Down Sequencing

Power sequencing between the 1.5 V power supply and  $V_{DDSYN}$  or the  $\overline{RESET}$  power supplies is required if using an external 1.5 V power supply with  $V_{RC33}$  tied to ground (GND). To avoid power-sequencing,  $V_{RC33}$  must be powered up within the specified operating range, even if the on-chip voltage regulator controller is not used. Refer to [Section 3.7.2, “Power-Up Sequence \(VRC33 Grounded\),”](#) and [Section 3.7.3, “Power-Down Sequence \(VRC33 Grounded\).”](#)

Power sequencing requires that  $V_{DD33}$  must reach a certain voltage where the values are read as ones before the POR signal negates. Refer to [Section 3.7.1, “Input Value of Pins During POR Dependent on VDD33.”](#)

Although power sequencing is not required between  $V_{RC33}$  and  $V_{DDSYN}$  during power up,  $V_{RC33}$  must not lead  $V_{DDSYN}$  by more than 600 mV or lag by more than 100 mV for the  $V_{RC}$  stage turn-on to operate within specification. Higher spikes in the emitter current of the pass transistor occur if  $V_{RC33}$  leads or lags  $V_{DDSYN}$  by more than these amounts. The value of that higher spike in current depends on the board power supply circuitry and the amount of board level capacitance.

Furthermore, when all of the PORs negate, the system clock starts to toggle, adding another large increase of the current consumed by  $V_{RC33}$ . If  $V_{RC33}$  lags  $V_{DDSYN}$  by more than 100 mV, the increase in current consumed can drop  $V_{DD}$  low enough to assert the 1.5 V POR again. Oscillations are possible when the 1.5 V POR asserts and stops the system clock, causing the voltage on  $V_{DD}$  to rise until the 1.5 V POR negates again. All oscillations stop when  $V_{RC33}$  is powered sufficiently.

When powering down,  $V_{RC33}$  and  $V_{DDSYN}$  have no delta requirement to each other, because the bypass capacitors internal and external to the device are already charged. When not powering up or down, no delta between  $V_{RC33}$  and  $V_{DDSYN}$  is required for the  $V_{RC}$  to operate within specification.

There are no power up/down sequencing requirements to prevent issues such as latch-up, excessive current spikes, and so on. Therefore, the state of the I/O pins during power up/down varies depending on which supplies are powered.

[Table 7](#) gives the pin state for the sequence cases for all pins with pad type pad\_fc (fast type).

**Table 7. Power Sequence Pin Status for Fast Pads**

$V_{DDE}$	$V_{DD33}$	$V_{DD}$	POR	Pin Status for Fast Pad Output Driver
				pad_fc (fast)
Low	—	—	Asserted	Low
$V_{DDE}$	Low	Low	Asserted	High
$V_{DDE}$	Low	$V_{DD}$	Asserted	High
$V_{DDE}$	$V_{DD33}$	Low	Asserted	High impedance (Hi-Z)
$V_{DDE}$	$V_{DD33}$	$V_{DD}$	Asserted	Hi-Z
$V_{DDE}$	$V_{DD33}$	$V_{DD}$	Negated	Functional

Table 8 gives the pin state for the sequence cases for all pins with pad type pad\_mh (medium type) and pad\_sh (slow type).

**Table 8. Power Sequence Pin Status for Medium / Slow Pads**

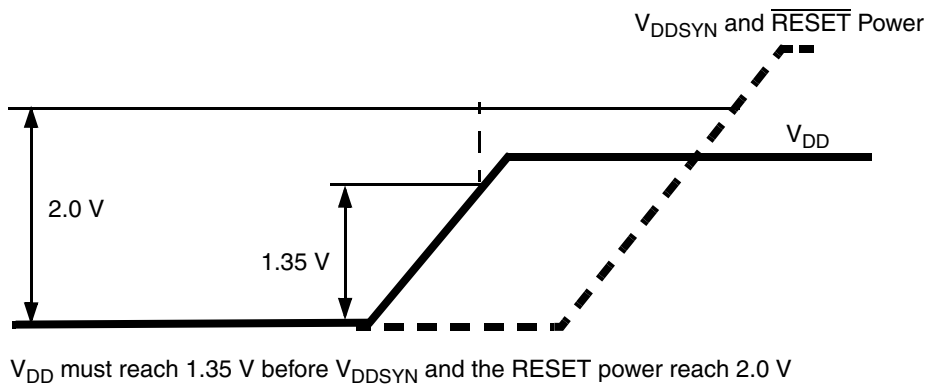
V <sub>DDEH</sub>	V <sub>DD</sub>	POR	Pin Status for Medium and Slow Pad Output Driver	
			pad_mh (medium)	pad_sh (slow)
Low	—	Asserted	Low	
V <sub>DDEH</sub>	Low	Asserted	High impedance (Hi-Z)	
V <sub>DDEH</sub>	V <sub>DD</sub>	Asserted	Hi-Z	
V <sub>DDEH</sub>	V <sub>DD</sub>	Negated	Functional	

### 3.7.1 Input Value of Pins During POR Dependent on V<sub>DD33</sub>

When powering up the device, V<sub>DD33</sub> must not lag the latest V<sub>DDSYN</sub> or  $\overline{\text{RESET}}$  power pin (V<sub>DDEH6</sub>) by more than the V<sub>DD33</sub> lag specification listed in Table 6, spec 8. This avoids accidentally selecting the bypass clock mode because the internal versions of PLLCFG[0:1] and RSTCFG are not powered and therefore cannot read the default state when POR negates. V<sub>DD33</sub> can lag V<sub>DDSYN</sub> or the  $\overline{\text{RESET}}$  power pin (V<sub>DDEH6</sub>), but cannot lag both by more than the V<sub>DD33</sub> lag specification. This V<sub>DD33</sub> lag specification applies during power up only. V<sub>DD33</sub> has no lead or lag requirements when powering down.

### 3.7.2 Power-Up Sequence (V<sub>RC33</sub> Grounded)

The 1.5 V V<sub>DD</sub> power supply must rise to 1.35 V before the 3.3 V V<sub>DDSYN</sub> power supply and the  $\overline{\text{RESET}}$  power supply rises above 2.0 V. This ensures that digital logic in the PLL for the 1.5 V power supply does not begin to operate below the specified operation range lower limit of 1.35 V. Because the internal 1.5 V POR is disabled, the internal 3.3 V POR or the  $\overline{\text{RESET}}$  power POR must hold the device in reset. Since they can negate as low as 2.0 V, V<sub>DD</sub> must be within specification before the 3.3 V POR and the  $\overline{\text{RESET}}$  POR negate.



**Figure 2. Power-Up Sequence (V<sub>RC33</sub> Grounded)**

### 3.7.3 Power-Down Sequence ( $V_{RC33}$ Grounded)

The only requirement for the power-down sequence when  $V_{RC33}$  is grounded is that if  $V_{DD}$  decreases to less than its operating range,  $V_{DDSYN}$  or the RESET power must decrease to less than 2.0 V before the  $V_{DD}$  power is allowed to increase to its operating range. This ensures that the digital 1.5 V logic, which is reset by the ORed POR only and can cause the 1.5 V supply to decrease below its specification, is reset properly.

## 3.8 DC Electrical Specifications

Table 9. DC Electrical Specifications

Spec	Characteristic	Symbol	Min	Max	Unit
1	Core supply voltage (average DC RMS voltage)	$V_{DD}$	1.35	1.65	V
2	I/O supply voltage (fast I/O)	$V_{DDE}$	1.62	3.6	V
3	I/O supply voltage (slow / medium I/O)	$V_{DDEH}$	3.0	5.25	V
4	3.3 V I/O buffer voltage	$V_{DD33}$	3.0	3.6	V
5	Voltage regulator control input voltage	$V_{RC33}$	3.0	3.6	V
6	Analog supply voltage <sup>1</sup>	$V_{DDA}$	4.5	5.25	V
8	Flash programming voltage <sup>2</sup>	$V_{PP}$	4.5	5.25	V
9	Flash read voltage	$V_{FLASH}$	3.0	3.6	V
10	SRAM standby voltage <sup>3</sup>	$V_{STBY}$	0.8	1.2	V
11	Clock synthesizer operating voltage	$V_{DDSYN}$	3.0	3.6	V
12	Fast I/O input high voltage	$V_{IH\_F}$	$0.65 \times V_{DDE}$	$V_{DDE} + 0.3$	V
13	Fast I/O input low voltage	$V_{IL\_F}$	$V_{SS} - 0.3$	$0.35 \times V_{DDE}$	V
14	Medium / slow I/O input high voltage	$V_{IH\_S}$	$0.65 \times V_{DDEH}$	$V_{DDEH} + 0.3$	V
15	Medium / slow I/O input low voltage	$V_{IL\_S}$	$V_{SS} - 0.3$	$0.35 \times V_{DDEH}$	V
16	Fast I/O input hysteresis	$V_{HYS\_F}$	$0.1 \times V_{DDE}$		V
17	Medium / slow I/O input hysteresis	$V_{HYS\_S}$	$0.1 \times V_{DDEH}$		V
18	Analog input voltage	$V_{INDC}$	$V_{SSA} - 0.3$	$V_{DDA} + 0.3$	V
19	Fast I/O output high voltage ( $I_{OH\_F} = -2.0$ mA )	$V_{OH\_F}$	$0.8 \times V_{DDE}$	—	V
20	Slow / medium I/O output high voltage ( $I_{OH\_S} = -2.0$ mA )	$V_{OH\_S}$	$0.8 \times V_{DDEH}$	—	V
21	Fast I/O output low voltage ( $I_{OL\_F} = 2.0$ mA )	$V_{OL\_F}$	—	$0.2 \times V_{DDE}$	V
22	Slow / medium I/O output low voltage ( $I_{OL\_S} = 2.0$ mA )	$V_{OL\_S}$	—	$0.2 \times V_{DDEH}$	V
23	Load capacitance (fast I/O) <sup>4</sup> DSC (SIU_PCR[8:9] ) = 0b00 DSC (SIU_PCR[8:9] ) = 0b01 DSC (SIU_PCR[8:9] ) = 0b10 DSC (SIU_PCR[8:9] ) = 0b11	$C_L$	— — — —	10 20 30 50	pF pF pF pF

## Electrical Characteristics

**Table 9. DC Electrical Specifications (continued)**

Spec	Characteristic	Symbol	Min	Max	Unit
24	Input capacitance (digital pins)	$C_{IN}$	—	7	pF
25	Input capacitance (analog pins)	$C_{IN\_A}$	—	10	pF
26	Input capacitance (Shared digital and analog pins AN[12]_MA[0]_SDS, AN[12]_MA[1]_SDO, AN[14]_MA[2]_SDI, and AN[15]_FCK)	$C_{IN\_M}$	—	12	pF
27a	Operating current <sup>5</sup> 1.5 V supplies @ 132 MHz:  $V_{DD}$ (including $V_{DDF}$ max current) <sup>6,7</sup> @ 1.65 V typical use $V_{DD}$ (including $V_{DDF}$ max current) <sup>6,7</sup> @ 1.35 V typical use $V_{DD}$ (including $V_{DDF}$ max current) <sup>7,8</sup> @ 1.65 V high use $V_{DD}$ (including $V_{DDF}$ max current) <sup>7,8</sup> @ 1.35 V high use	  $I_{DD}$ $I_{DD}$ $I_{DD}$ $I_{DD}$	  — — — —	  550 <sup>9</sup> 450 <sup>9</sup> 600 <sup>9</sup> 490 <sup>9</sup>	  mA mA mA mA
27b	Operating current <sup>5</sup> 1.5 V supplies @ 114 MHz:  $V_{DD}$ (including $V_{DDF}$ max current) <sup>6,7</sup> @ 1.65 V typical use $V_{DD}$ (including $V_{DDF}$ max current) <sup>6,7</sup> @ 1.35 V typical use $V_{DD}$ (including $V_{DDF}$ max current) <sup>7,8</sup> @ 1.65 V high use $V_{DD}$ (including $V_{DDF}$ max current) <sup>7,8</sup> @ 1.35 V high use	  $I_{DD}$ $I_{DD}$ $I_{DD}$ $I_{DD}$	  — — — —	  460 <sup>9</sup> 380 <sup>9</sup> 520 <sup>9</sup> 420 <sup>9</sup>	  mA mA mA mA
27c	Operating current <sup>5</sup> 1.5 V supplies @ 82 MHz:  $V_{DD}$ (including $V_{DDF}$ max current) <sup>6,7</sup> @ 1.65 V typical use $V_{DD}$ (including $V_{DDF}$ max current) <sup>6,7</sup> @ 1.35 V typical use $V_{DD}$ (including $V_{DDF}$ max current) <sup>7,8</sup> @ 1.65 V high use $V_{DD}$ (including $V_{DDF}$ max current) <sup>7,8</sup> @ 1.35 V high use	  $I_{DD}$ $I_{DD}$ $I_{DD}$ $I_{DD}$	  — — — —	  350 <sup>9</sup> 290 <sup>9</sup> 400 <sup>9</sup> 330 <sup>9</sup>	  mA mA mA mA
27d	Refer to <a href="#">Figure 3</a> for an interpolation of this data. <sup>10</sup> $I_{DD\_STBY}$ @ 25° C $V_{STBY}$ @ 0.8 V $V_{STBY}$ @ 1.0 V $V_{STBY}$ @ 1.2 V  $I_{DD\_STBY}$ @ 60° C $V_{STBY}$ @ 0.8 V $V_{STBY}$ @ 1.0 V $V_{STBY}$ @ 1.2 V  $I_{DD\_STBY}$ @ 150° C (Tj) $V_{STBY}$ @ 0.8 V $V_{STBY}$ @ 1.0 V $V_{STBY}$ @ 1.2 V	  $I_{DD\_STBY}$ $I_{DD\_STBY}$ $I_{DD\_STBY}$  $I_{DD\_STBY}$ $I_{DD\_STBY}$ $I_{DD\_STBY}$  $I_{DD\_STBY}$ $I_{DD\_STBY}$ $I_{DD\_STBY}$	  — — —  — — —  — — —	  20 30 50  70 100 200  1200 1500 2000	  μA μA μA  μA μA μA  μA μA μA
28	Operating current 3.3 V supplies @ 132 MHz  $V_{DD33}$ <sup>11</sup>  $V_{FLASH}$ $V_{DDSYN}$	  $I_{DD33}$  $I_{VFLASH}$ $I_{DDSYN}$	  —  — —	  2 + (values derived from procedure of Footnote <sup>11</sup> )  10 15	  mA  mA mA

Table 9. DC Electrical Specifications (continued)

Spec	Characteristic	Symbol	Min	Max	Unit
29	Operating current 5.0 V supplies (12 MHz ADCLK):				
	$V_{DDA}$ ( $V_{DDA0} + V_{DDA1}$ )	$I_{DDA}$	—	20.0	mA
	Analog reference supply current ( $V_{RH}$ , $V_{RL}$ )	$I_{REF}$	—	1.0	mA
	$V_{PP}$	$I_{PP}$	—	25.0	mA
30	Operating current $V_{DDE}^{12}$ supplies:				
	$V_{DDEH1}$	IDD1	—	Refer to Footnote <sup>12</sup>	mA
	$V_{DDE2}$	IDD2	—		mA
	$V_{DDE3}$	IDD3	—		mA
	$V_{DDEH4}$	IDD4	—		mA
	$V_{DDE5}$	IDD5	—		mA
	$V_{DDEH6}$	IDD6	—		mA
	$V_{DDE7}$	IDD7	—		mA
	$V_{DDEH8}$	IDD8	—		mA
	$V_{DDEH9}$	IDD9	—		mA
31	Fast I/O weak pullup current <sup>13</sup>	$I_{ACT\_F}$			
	1.62–1.98 V		10	110	$\mu$ A
	2.25–2.75 V		20	130	$\mu$ A
	3.00–3.60 V		20	170	$\mu$ A
	Fast I/O weak pulldown current <sup>13</sup>	$I_{ACT\_F}$			
	1.62–1.98 V		10	100	$\mu$ A
	2.25–2.75 V		20	130	$\mu$ A
	3.00–3.60 V		20	170	$\mu$ A
32	Slow / medium I/O weak pullup/down current <sup>14</sup>	$I_{ACT\_S}$			
	3.0–3.6 V 4.5–5.5 V		10 20	150 170	$\mu$ A $\mu$ A
33	I/O input leakage current <sup>15</sup>	$I_{INACT\_D}$	–2.5	2.5	$\mu$ A
34	DC injection current (per pin)	$I_{IC}$	–2.0	2.0	mA
35	Analog input current, channel off <sup>16</sup>	$I_{INACT\_A}$	–150	150	nA
35a	Analog input current, shared analog / digital pins (AN[12], AN[13], AN[14], AN[15])	$I_{INACT\_AD}$	–2.5	2.5	$\mu$ A
36	$V_{SS}$ differential voltage <sup>17</sup>	$V_{SS} - V_{SSA}$	–100	100	mV
37	Analog reference low voltage	$V_{RL}$	$V_{SSA} - 0.1$	$V_{SSA} + 0.1$	V
38	$V_{RL}$ differential voltage	$V_{RL} - V_{SSA}$	–100	100	mV
39	Analog reference high voltage	$V_{RH}$	$V_{DDA} - 0.1$	$V_{DDA} + 0.1$	V
40	$V_{REF}$ differential voltage	$V_{RH} - V_{RL}$	4.5	5.25	V
41	$V_{SSSYN}$ to $V_{SS}$ differential voltage	$V_{SSSYN} - V_{SS}$	–50	50	mV
42	$V_{RCVSS}$ to $V_{SS}$ differential voltage	$V_{RCVSS} - V_{SS}$	–50	50	mV
43	$V_{DDF}$ to $V_{DD}$ differential voltage <sup>2</sup>	$V_{DDF} - V_{DD}$	–100	100	mV
43a	$V_{RC33}$ to $V_{DDSYN}$ differential voltage	$V_{RC33} - V_{DDSYN}$	–0.1	0.1 <sup>18</sup>	V

## Electrical Characteristics

**Table 9. DC Electrical Specifications (continued)**

Spec	Characteristic	Symbol	Min	Max	Unit
44	Analog input differential signal range (with common mode 2.5 V)	$V_{IDIFF}$	- 2.5	2.5	V
45	Operating temperature range, ambient (packaged)	$T_A = (T_L \text{ to } T_H)$	$T_L$	$T_H$	°C
46	Slew rate on power-supply pins	—	—	50	V/ms

<sup>1</sup>  $|V_{DDA0} - V_{DDA1}|$  must be < 0.1 V.

<sup>2</sup>  $V_{PP}$  can drop to 3.0 V during read operations.

<sup>3</sup> During standby operation, if standby operation is not required, connect  $V_{STBY}$  to ground.

<sup>4</sup> Applies to CLKOUT, external bus pins, and Nexus pins.

<sup>5</sup> Maximum average RMS DC current.

<sup>6</sup> Average current measured on Automotive benchmark.

<sup>7</sup> Peak currents can be higher on specialized code.

<sup>8</sup> High use current measured while running optimized SPE assembly code with all code and data 100% locked in cache (0% miss rate) with all channels of the eMIOS and eTPU running autonomously, plus the eDMA transferring data continuously from SRAM to SRAM. Higher currents are possible if an idle loop that crosses cache lines is run from cache. Design and write code to avoid this condition.

<sup>9</sup> Preliminary. Final specification pending characterization.

<sup>10</sup> [Figure 3](#) shows an illustration of the  $IDD_{STBY}$  values interpolated for these temperature values.

<sup>11</sup> Power requirements for the  $V_{DD33}$  supply depend on the frequency of operation and load of all I/O pins, and the voltages on the I/O segments. Refer to [Table 11](#) for values to calculate power dissipation for specific operation.

<sup>12</sup> Power requirements for each I/O segment are dependent on the frequency of operation and load of the I/O pins on a particular I/O segment, and the voltage of the I/O segment. Refer to [Table 10](#) for values to calculate power dissipation for specific operation. The total power consumption of an I/O segment is the sum of the individual power consumptions for each pin on the segment.

<sup>13</sup> Absolute value of current, measured at  $V_{IL}$  and  $V_{IH}$ .

<sup>14</sup> Absolute value of current, measured at  $V_{IL}$  and  $V_{IH}$ .

<sup>15</sup> Weak pullup/down inactive. Measured at  $V_{DDE} = 3.6$  V and  $V_{DDEH} = 5.25$  V. Applies to pad types: pad\_fc, pad\_sh, and pad\_mh.

<sup>16</sup> Maximum leakage occurs at maximum operating temperature. Leakage current decreases by approximately one-half for each 8 °C to 12 °C, in the ambient temperature range of 50 °C to 125 °C. Applies to pad types: pad\_a and pad\_ae.

<sup>17</sup>  $V_{SSA}$  refers to both  $V_{SSA0}$  and  $V_{SSA1}$ .  $|V_{SSA0} - V_{SSA1}|$  must be < 0.1 V.

<sup>18</sup> Up to 0.6 V during power up and power down.



Figure 3 shows an approximate interpolation of the  $I_{STBY}$  worst-case specification to help estimate the values at different voltages and temperatures. The vertical lines inside the graph show the actual specifications listed in Table 9. Refer to the  $IDD_{STBY}$  specifications (27d) in Table 9 for more information.

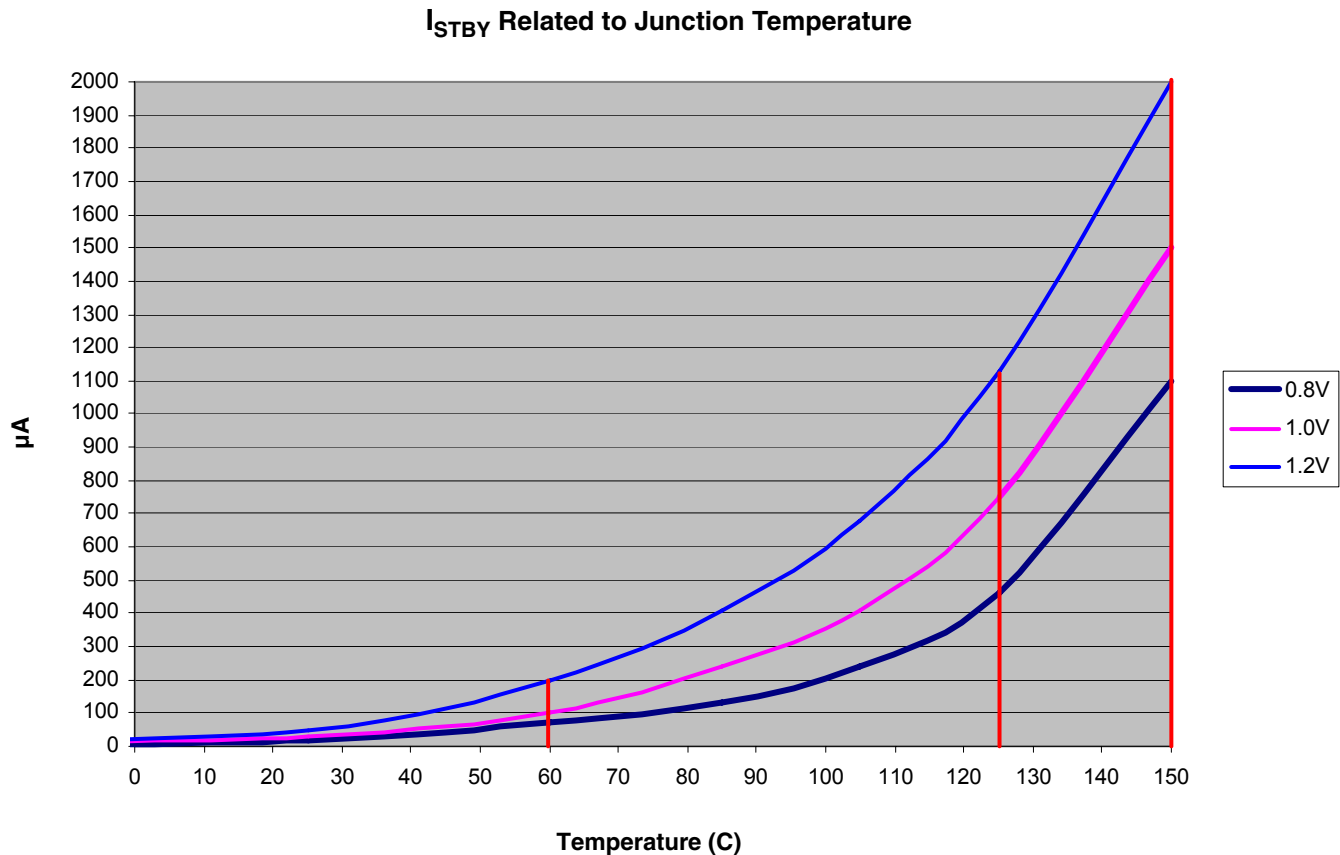


Figure 3.  $I_{STBY}$  Worst-case Specifications

### 3.8.1 I/O Pad Current Specifications

The power consumption of an I/O segment depends on the usage of the pins on a particular segment. The power consumption is the sum of all output pin currents for a particular segment. The output pin current can be calculated from Table 10 based on the voltage, frequency, and load on the pin. Use linear scaling to calculate pin currents for voltage, frequency, and load parameters that fall outside the values given in Table 10.

Table 10. I/O Pad Average DC Current<sup>1</sup>

Spec	Pad Type	Symbol	Frequency (MHz)	Load <sup>2</sup> (pF)	Voltage (V)	Drive Select / Slew Rate Control Setting	Current (mA)
1	Slow	IDRV_SH	25	50	5.25	11	8.0
2			10	50	5.25	01	3.2
3			2	50	5.25	00	0.7
4			2	200	5.25	00	2.4
5	Medium	IDRV_MH	50	50	5.25	11	17.3
6			20	50	5.25	01	6.5
7			3.33	50	5.25	00	1.1
8			3.33	200	5.25	00	3.9
9	Fast	IDRV_FC	66	10	3.6	00	2.8
10			66	20	3.6	01	5.2
11			66	30	3.6	10	8.5
12			66	50	3.6	11	11.0
13			66	10	1.98	00	1.6
14			66	20	1.98	01	2.9
15			66	30	1.98	10	4.2
16			66	50	1.98	11	6.7
17			56	10	3.6	00	2.4
18			56	20	3.6	01	4.4
19			56	30	3.6	10	7.2
20			56	50	3.6	11	9.3
21			56	10	1.98	00	1.3
22			56	20	1.98	01	2.5
23			56	30	1.98	10	3.5
24			56	50	1.98	11	5.7
25			40	10	3.6	00	1.7
26			40	20	3.6	01	3.1
27			40	30	3.6	10	5.1
28			40	50	3.6	11	6.6
29			40	10	1.98	00	1.0
30			40	20	1.98	01	1.8
31			40	30	1.98	10	2.5
32			40	50	1.98	11	4.0

<sup>1</sup> These values are estimates from simulation and are not tested. Currents apply to output pins only.

<sup>2</sup> All loads are lumped.

### 3.8.2 I/O Pad $V_{DD33}$ Current Specifications

The power consumption of the  $V_{DD33}$  supply depends on the usage of the pins on all I/O segments. The power consumption is the sum of all input and output pin  $V_{DD33}$  currents for all I/O segments. The output pin  $V_{DD33}$  current can be calculated from Table 11 based on the voltage, frequency, and load on all fast (pad\_fc) pins. The input pin  $V_{DD33}$  current can be calculated from Table 11 based on the voltage, frequency, and load on all pad\_sh and pad\_sh pins. Use linear scaling to calculate pin currents for voltage, frequency, and load parameters that fall outside the values given in Table 11.

Table 11.  $V_{DD33}$  Pad Average DC Current<sup>1</sup>

Spec	Pad Type	Symbol	Frequency (MHz)	Load <sup>2</sup> (pF)	$V_{DD33}$ (V)	$V_{DDE}$ (V)	Drive Select	Current (mA)
<b>Inputs</b>								
1	Slow	I33_SH	66	0.5	3.6	5.5	NA	0.003
2	Medium	I33_MH	66	0.5	3.6	5.5	NA	0.003
<b>Outputs</b>								
3	Fast	I33_FC	66	10	3.6	3.6	00	0.35
4			66	20	3.6	3.6	01	0.53
5			66	30	3.6	3.6	10	0.62
6			66	50	3.6	3.6	11	0.79
7			66	10	3.6	1.98	00	0.35
8			66	20	3.6	1.98	01	0.44
9			66	30	3.6	1.98	10	0.53
10			66	50	3.6	1.98	11	0.7
11			56	10	3.6	3.6	00	0.30
12			56	20	3.6	3.6	01	0.45
13			56	30	3.6	3.6	10	0.52
14			56	50	3.6	3.6	11	0.67
15			56	10	3.6	1.98	00	0.30
16			56	20	3.6	1.98	01	0.37
17			56	30	3.6	1.98	10	0.45
18			56	50	3.6	1.98	11	0.60
19			40	10	3.6	3.6	00	0.21
20			40	20	3.6	3.6	01	0.31
21			40	30	3.6	3.6	10	0.37
22			40	50	3.6	3.6	11	0.48
23			40	10	3.6	1.98	00	0.21
24			40	20	3.6	1.98	01	0.27
25			40	30	3.6	1.98	10	0.32
26			40	50	3.6	1.98	11	0.42

<sup>1</sup> These values are estimated from simulation and not tested. Currents apply to output pins for the fast pads only and to input pins for the slow and medium pads only.

<sup>2</sup> All loads are lumped.

### 3.9 Oscillator and FMPLL Electrical Characteristics

**Table 12. FMPLL Electrical Specifications**  
 ( $V_{DDSYN} = 3.0\text{--}3.6\text{ V}$ ;  $V_{SS} = V_{SSSYN} = 0.0\text{ V}$ ;  $T_A = T_L\text{ to }T_H$ )

Spec	Characteristic	Symbol	Minimum	Maximum	Unit
1	PLL reference frequency range: Crystal reference External reference Dual controller (1:1 mode)	$f_{ref\_crystal}$ $f_{ref\_ext}$ $f_{ref\_1:1}$	8 8 24	20 20 $f_{sys} \div 2$	MHz
2	System frequency <sup>1</sup>	$f_{sys}$	$f_{ICO(MIN)} \div 2^{RFD}$	$f_{MAX}$ <sup>2</sup>	MHz
3	System clock period	$t_{CYC}$	—	$1 \div f_{sys}$	ns
4	Loss of reference frequency <sup>3</sup>	$f_{LOR}$	100	1000	kHz
5	Self clocked mode (SCM) frequency <sup>4</sup>	$f_{SCM}$	7.4	17.5	MHz
6	EXTAL input high voltage crystal mode <sup>5</sup>	$V_{IHEXT}$	$V_{XTAL} + 0.4\text{ V}$	—	V
	All other modes (dual controller (1:1), bypass, external reference)	$V_{IHEXT}$	$[(V_{DDE5} \div 2) + 0.4\text{ V}]$	—	V
7	EXTAL input low voltage crystal mode <sup>6</sup>	$V_{ILEXT}$	—	$V_{XTAL} - 0.4\text{ V}$	V
	All other modes (dual controller (1:1), bypass, external reference)	$V_{ILEXT}$	—	$[(V_{DDE5} \div 2) - 0.4\text{ V}]$	V
8	XTAL current <sup>7</sup>	$I_{XTAL}$	0.8	3	mA
9	Total on-chip stray capacitance on XTAL	$C_{S\_XTAL}$	—	1.5	pF
10	Total on-chip stray capacitance on EXTAL	$C_{S\_EXTAL}$	—	1.5	pF
11	Crystal manufacturer's recommended capacitive load	$C_L$	Refer to crystal specification	Refer to crystal specification	pF
12	Discrete load capacitance to connect to EXTAL	$C_{L\_EXTAL}$	—	$(2 \times C_L) - C_{S\_EXTAL}$ $- C_{PCB\_EXTAL}$ <sup>8</sup>	pF
13	Discrete load capacitance to connect to XTAL	$C_{L\_XTAL}$	—	$(2 \times C_L) - C_{S\_XTAL}$ $- C_{PCB\_XTAL}$ <sup>8</sup>	pF
14	PLL lock time <sup>9</sup>	$t_{pLL}$	—	750	$\mu\text{s}$
15	Dual controller (1:1) clock skew (between CLKOUT and EXTAL) <sup>10, 11</sup>	$t_{skew}$	-2	2	ns
16	Duty cycle of reference	$t_{DC}$	40	60	%
17	Frequency un-LOCK range	$f_{UL}$	-4.0	4.0	% $f_{SYS}$
18	Frequency LOCK range	$f_{LCK}$	-2.0	2.0	% $f_{SYS}$

Table 12. FMPLL Electrical Specifications (continued)

 $(V_{DDSYN} = 3.0\text{--}3.6\text{ V}; V_{SS} = V_{SSSYN} = 0.0\text{ V}; T_A = T_L \text{ to } T_H)$ 

Spec	Characteristic	Symbol	Minimum	Maximum	Unit
19	CLKOUT period jitter, <sup>12, 13</sup> measured at $f_{SYS}$ maximum peak-to-peak jitter (clock edge to clock edge) Long term jitter (averaged over 2 ms interval)	$C_{JITTER}$	—	5.0	% $f_{CLKOUT}$
20	Frequency modulation range limit <sup>14</sup> (do not exceed $f_{SYS}$ maximum)	$C_{MOD}$	0.8	2.4	% $f_{SYS}$
21	ICO frequency $f_{ICO} = [f_{REF} \times (MFD + 4)] \div (PREDIV + 1)$ <sup>15</sup>	$f_{ICO}$	48	$f_{SYS}$	MHz
22	Predivider output frequency (to PLL)	$f_{PREDIV}$	4	$f_{MAX}$	MHz

<sup>1</sup> All internal registers retain data at 0 Hz.

<sup>2</sup> Up to the maximum frequency rating of the device (refer to Table 1).

<sup>3</sup> Loss of reference frequency is defined as the reference frequency detected internally, which transitions the PLL into self-clocked mode.

<sup>4</sup> The PLL operates at self-clocked mode (SCM) frequency when the reference frequency falls below  $f_{LOR}$ . SCM frequency is measured on the CLKOUT ball with the divider set to divide-by-two of the system clock.

NOTE: In SCM, the MFD and PREDIV have no effect and the RFD is bypassed.

<sup>5</sup> Use the EXTAL input high voltage parameter when using the FlexCAN oscillator in crystal mode (no quartz crystals or resonators).  $(V_{EXTAL} - V_{XTAL})$  must be  $\geq 400$  mV for the oscillator's comparator to produce the output clock.

<sup>6</sup> Use the EXTAL input low voltage parameter when using the FlexCAN oscillator in crystal mode (no quartz crystals or resonators).  $(V_{XTAL} - V_{EXTAL})$  must be  $\geq 400$  mV for the oscillator's comparator to produce the output clock.

<sup>7</sup>  $I_{XTAL}$  is the oscillator bias current out of the XTAL pin with both EXTAL and XTAL pins grounded.

<sup>8</sup>  $C_{PCB\_EXTAL}$  and  $C_{PCB\_XTAL}$  are the measured PCB stray capacitances on EXTAL and XTAL, respectively.

<sup>9</sup> This specification applies to the period required for the PLL to relock after changing the MFD frequency control bits in the synthesizer control register (SYNCR). From power up with crystal oscillator reference, the lock time also includes the crystal startup time.

<sup>10</sup> PLL is operating in 1:1 PLL mode.

<sup>11</sup>  $V_{DDE} = 3.0\text{--}3.6\text{ V}$

<sup>12</sup> Jitter is the average deviation from the programmed frequency measured over the specified interval at maximum  $f_{SYS}$ . Measurements are made with the device powered by filtered supplies and clocked by a stable external clock signal. Noise injected into the PLL circuitry via  $V_{DDSYN}$  and  $V_{SSSYN}$  and variation in crystal oscillator frequency increase the jitter percentage for a given interval. CLKOUT divider is set to divide-by-two.

<sup>13</sup> Values are with frequency modulation disabled. If frequency modulation is enabled, jitter is the sum of (jitter + Cmod).

<sup>14</sup> Modulation depth selected must not result in  $f_{SYS}$  value greater than the  $f_{SYS}$  maximum specified value.

<sup>15</sup>  $f_{SYS} = f_{ICO} \div (2^{RFD})$ .

### 3.10 eQADC Electrical Characteristics

Table 13. eQADC Conversion Specifications (Operating)

Spec	Characteristic	Symbol	Minimum	Maximum	Unit
1	ADC clock (ADCLK) frequency <sup>1</sup>	F <sub>ADCLK</sub>	1	12	MHz
2	Conversion cycles Differential Single ended	CC	13 + 2 (15) 14 + 2 (16)	13 + 128 (141) 14 + 128 (142)	ADCLK cycles
3	Stop mode recovery time <sup>2</sup>	T <sub>SR</sub>	10	—	μs
4	Resolution <sup>3</sup>	—	1.25	—	mV
5	INL: 6 MHz ADC clock	INL6	-4	4	Counts <sup>3</sup>
6	INL: 12 MHz ADC clock	INL12	-8	8	Counts
7	DNL: 6 MHz ADC clock	DNL6	-3 <sup>4</sup>	3 <sup>4</sup>	Counts
8	DNL: 12 MHz ADC clock	DNL12	-6 <sup>4</sup>	6 <sup>4</sup>	Counts
9	Offset error with calibration	OFFWC	-4 <sup>5</sup>	4 <sup>5</sup>	Counts
10	Full-scale gain error with calibration	GAINWC	-8 <sup>6</sup>	8 <sup>6</sup>	Counts
11	Disruptive input injection current <sup>7, 8, 9, 10</sup>	I <sub>INJ</sub>	-1	1	mA
12	Incremental error due to injection current. All channels have same 10 kΩ < R <sub>s</sub> < 100 kΩ Channel under test has R <sub>s</sub> = 10 kΩ, I <sub>INJ</sub> = I <sub>INJMAX</sub> , I <sub>INJMIN</sub>	E <sub>INJ</sub>	-4	4	Counts
13	Total Unadjusted Error for single ended conversions with calibration <sup>11, 12, 13, 14, 15</sup>	TUE	-4	4	Counts

<sup>1</sup> Conversion characteristics vary with F<sub>ADCLK</sub> rate. Reduced conversion accuracy occurs at maximum F<sub>ADCLK</sub> rate. The maximum value is based on 800 KS/s and the minimum value is based on 20 MHz oscillator clock frequency divided by a maximum 16 factor.

<sup>2</sup> Stop mode recovery time begins when the ADC control register enable bits are set until the ADC is ready to perform conversions.

<sup>3</sup> At V<sub>RH</sub> - V<sub>RL</sub> = 5.12 V, one least significant bit (LSB) = 1.25, mV = one count.

<sup>4</sup> Guaranteed 10-bit monotonicity.

<sup>5</sup> The absolute value of the offset error without calibration ≤ 100 counts.

<sup>6</sup> The absolute value of the full scale gain error without calibration ≤ 120 counts.

<sup>7</sup> Below disruptive current conditions, the channel being stressed has conversion values of: 0x3FF for analog inputs greater than V<sub>RH</sub>, and 0x000 for values less than V<sub>RL</sub>. This assumes that V<sub>RH</sub> ≤ V<sub>DDA</sub> and V<sub>RL</sub> ≥ V<sub>SSA</sub> due to the presence of the sample amplifier. Other channels are not affected by non-disruptive conditions.

<sup>8</sup> Exceeding the limit can cause a conversion error on both stressed and unstressed channels. Transitions within the limit do not affect device reliability or cause permanent damage.

<sup>9</sup> Input must be current limited to the value specified. To determine the value of the required current-limiting resistor, calculate resistance values using V<sub>POSClamp</sub> = V<sub>DDA</sub> + 0.5 V and V<sub>NEGClamp</sub> = -0.3 V, then use the larger of the calculated values.

<sup>10</sup> Condition applies to two adjacent pads on the internal pad.

<sup>11</sup> The TUE specification is always less than the sum of the INL, DNL, offset, and gain errors due to canceling errors.

<sup>12</sup> TUE does not apply to differential conversions.

<sup>13</sup> Measured at 6 MHz ADC clock. TUE with a 12 MHz ADC clock is: -16 counts < TUE < 16 counts.

<sup>14</sup> TUE includes all internal device errors such as internal reference variation (75% Ref, 25% Ref).

<sup>15</sup> Depending on the input impedance, the analog input leakage current (DC Electrical specification 35a) can affect the actual TUE measured on analog channels AN[12], AN[13], AN[14], AN[15].

### 3.11 H7Fa Flash Memory Electrical Characteristics

Table 14. Flash Program and Erase Specifications<sup>1</sup>

Spec	Flash Program Characteristic	Symbol	Min	Typical	Initial Max <sup>2</sup>	Max <sup>3</sup>	Unit
3	Doubleword (64 bits) program time <sup>4</sup>	$T_{dwprogram}$	—	10	—	500	$\mu$ s
4	Page program time <sup>4</sup>	$T_{pprogram}$	—	22	44 <sup>5</sup>	500	$\mu$ s
7	16 Kbyte block pre-program and erase time	$T_{16kpperase}$	—	325	525	5000	ms
9	48 Kbyte block pre-program and erase time	$T_{48kpperase}$	—	435	525	5000	ms
10	64 Kbyte block pre-program and erase time	$T_{64kpperase}$	—	525	675	5000	ms
8	128 Kbyte block pre-program and erase time	$T_{128kpperase}$	—	675	1800	15,000	ms
11	Minimum operating frequency for program and erase operations <sup>6</sup>	—	25	—	—	—	MHz

<sup>1</sup> Typical program and erase times assume nominal supply values and operation at 25 °C.

<sup>2</sup> Initial factory condition:  $\leq 100$  program/erase cycles, 25 °C, typical supply voltage, 80 MHz minimum system frequency.

<sup>3</sup> The maximum erase time occurs after the specified number of program/erase cycles. This maximum value is characterized but not guaranteed.

<sup>4</sup> Actual hardware programming times. This does not include software overhead.

<sup>5</sup> Page size is 256 bits (8 words).

<sup>6</sup> Read frequency of the flash can be up to the maximum operating frequency of the device. There is no minimum read frequency condition.

Table 15. Flash EEPROM Module Life (Full Temperature Range)

Spec	Characteristic	Symbol	Min	Typical <sup>1</sup>	Unit
1a	Number of program/erase cycles per block for 16 Kbyte, 48 Kbyte, and 64 Kbyte blocks over the operating temperature range ( $T_J$ )	P/E	100,000	—	cycles
1b	Number of program/erase cycles per block for 128 Kbyte blocks over the operating temperature range ( $T_J$ )	P/E	10,000	100,000	cycles
2	Data retention Blocks with 0–1,000 P/E cycles Blocks with 1,001–100,000 P/E cycles	Retention	20 5	—	years

<sup>1</sup> Typical endurance is evaluated at 25° C. Product qualification is performed to the minimum specification. For additional information on the Freescale definition of typical endurance, refer to engineering bulletin EB619 *Typical Endurance for Nonvolatile Memory*.

## Electrical Characteristics

Table 16 shows the FLASH\_BIU settings versus frequency of operation. Refer to the device reference manual for definitions of these bit fields.

**Table 16. FLASH\_BIU Settings vs. Frequency of Operation**

Maximum Frequency (MHz)	APC	RWSC	WWSC	DPFEN	IPFEN	PFLIM	BFEN
Up to and including 82 MHz <sup>1</sup>	0b001	0b001	0b01	0b00, 0b01, or 0b11 <sup>2</sup>	0b00, 0b01, or 0b11 <sup>2</sup>	0b000 to 0b110 <sup>3</sup>	0b0, 0b1 <sup>4</sup>
Up to and including 102 MHz <sup>5</sup>	0b001	0b010	0b01	0b00, 0b01, or 0b11 <sup>2</sup>	0b00, 0b01, or 0b11 <sup>2</sup>	0b000 to 0b110 <sup>3</sup>	0b0, 0b1 <sup>4</sup>
Up to and including 132 MHz <sup>6</sup>	0b010	0b011	0b01	0b00, 0b01, or 0b11 <sup>2</sup>	0b00, 0b01, or 0b11 <sup>2</sup>	0b000 to 0b110 <sup>3</sup>	0b0, 0b1 <sup>4</sup>
Default setting after reset	0b111	0b111	0b11	0b00	0b00	0b000	0b0

<sup>1</sup> Allows for 80 MHz system clock with 2% frequency modulation.

<sup>2</sup> For maximum flash performance, set to 0b11.

<sup>3</sup> For maximum flash performance, set to 0b110.

<sup>4</sup> For maximum flash performance, set to 0b1.

<sup>5</sup> Allows for 100 MHz system clock with 2% frequency modulation.

<sup>6</sup> Allows for 128 MHz system clock with 2% frequency modulation.

## 3.12 AC Specifications

### 3.12.1 Pad AC Specifications

**Table 17. Pad AC Specifications ( $V_{DDEH} = 5.0\text{ V}$ ,  $V_{DDE} = 1.8\text{ V}$ )<sup>1</sup>**

Spec	Pad	SRC/DSC (binary)	Out Delay <sup>2, 3, 4</sup> (ns)	Rise / Fall <sup>4, 5</sup> (ns)	Load Drive (pF)
1	Slow high voltage (SH)	11	26	15	50
			82	60	200
		01	75	40	50
			137	80	200
		00	377	200	50
			476	260	200
2	Medium high voltage (MH)	11	16	8	50
			43	30	200
		01	34	15	50
			61	35	200
		00	192	100	50
			239	125	200



Table 17. Pad AC Specifications ( $V_{DDEH} = 5.0\text{ V}$ ,  $V_{DDE} = 1.8\text{ V}$ )<sup>1</sup> (continued)

Spec	Pad	SRC/DSC (binary)	Out Delay <sup>2, 3, 4</sup> (ns)	Rise / Fall <sup>4, 5</sup> (ns)	Load Drive (pF)
3	Fast	00	3.1	2.7	10
		01		2.5	20
		10		2.4	30
		11		2.3	50
4	Pullup/down (3.6 V max)	—	—	7500	50
5	Pullup/down (5.5 V max)	—	—	9000	50

<sup>1</sup> These are worst case values that are estimated from simulation and not tested. The values in the table are simulated at  $F_{SYS} = 132\text{ MHz}$ ,  $V_{DD} = 1.35\text{--}1.65\text{ V}$ ,  $V_{DDE} = 1.62\text{--}1.98\text{ V}$ ,  $V_{DDEH} = 4.5\text{--}5.5\text{ V}$ ,  $V_{DD33}$  and  $V_{DDSYN} = 3.0\text{--}3.6\text{ V}$ ,  $T_A = T_L$  to  $T_H$ .

<sup>2</sup> This parameter is supplied for reference and is guaranteed by design and tested.

<sup>3</sup> Out delay is shown in Figure 4. Add a maximum of one system clock to the output delay for delay with respect to system clock.

<sup>4</sup> Delay and rise and fall are measured to 20% or 80% of the respective signal.

<sup>5</sup> This parameter is guaranteed by characterization before qualification rather than 100% tested.

Table 18. De-rated Pad AC Specifications ( $V_{DDEH} = 3.3\text{ V}$ ,  $V_{DDE} = 3.3\text{ V}$ )<sup>1</sup>

Spec	Pad	SRC/DSC (binary)	Out Delay <sup>2, 3, 4</sup> (ns)	Rise/Fall <sup>3, 5</sup> (ns)	Load Drive (pF)
1	Slow high voltage (SH)	11	39	23	50
			120	87	200
		01	101	52	50
			188	111	200
		00	507	248	50
			597	312	200
2	Medium high voltage (MH)	11	23	12	50
			64	44	200
		01	50	22	50
			90	50	200
		00	261	123	50
			305	156	200
3	Fast	3.2	3.2	2.4	10
				2.2	20
				2.1	30
				2.1	50
4	Pullup/down (3.6 V max)	—	—	7500	50
5	Pullup/down (5.5 V max)	—	—	9500	50

<sup>1</sup> These are worst-case values that are estimated from simulation and not tested. The values in the table are simulated at:  $F_{SYS} = 132\text{ MHz}$ ;  $V_{DD} = 1.35\text{--}1.65\text{ V}$ ;  $V_{DDE} = 3.0\text{--}3.6\text{ V}$ ;  $V_{DDEH} = 3.0\text{--}3.6\text{ V}$ ;  $V_{DD33}$  and  $V_{DDSYN} = 3.0\text{--}3.6\text{ V}$ ; and  $T_A = T_L$  to  $T_H$ .

## Electrical Characteristics

- <sup>2</sup> This parameter is supplied for reference and is guaranteed by design and tested.
- <sup>3</sup> The delay, and the rise and fall, are measured to 20% or 80% of the respective signal.
- <sup>4</sup> Out delay is shown in Figure 4. Add a maximum of one system clock to the output delay for delay with respect to system clock.
- <sup>5</sup> This parameter is guaranteed by characterization before qualification rather than 100% tested.

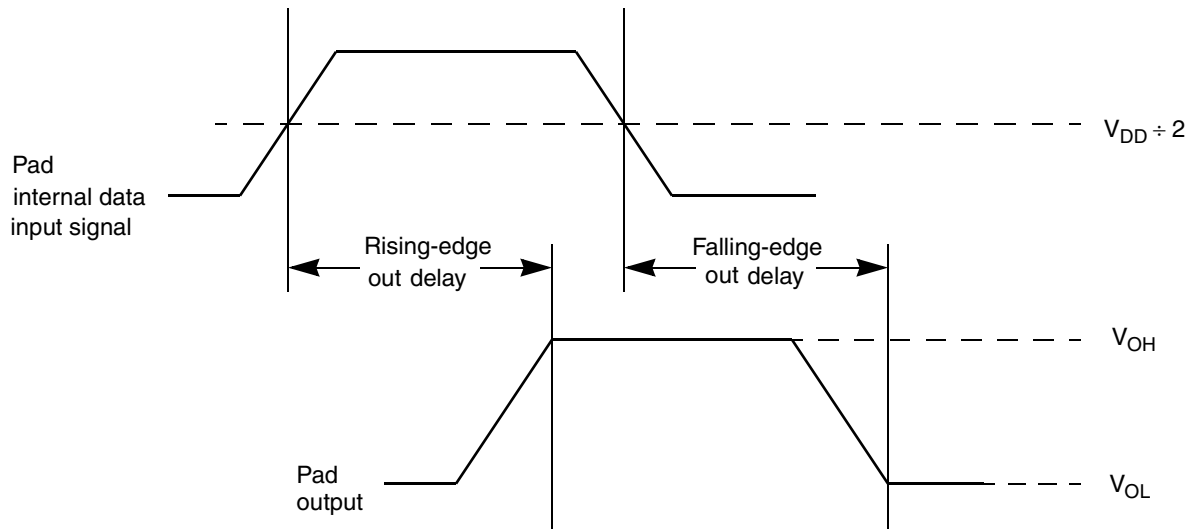


Figure 4. Pad Output Delay

## 3.13 AC Timing

### 3.13.1 Reset and Configuration Pin Timing

Table 19. Reset and Configuration Pin Timing<sup>1</sup>

Spec	Characteristic	Symbol	Min	Max	Unit
1	$\overline{\text{RESET}}$ pulse width	$t_{RPW}$	10	—	$t_{CYC}$
2	$\overline{\text{RESET}}$ glitch detect pulse width	$t_{GPW}$	2	—	$t_{CYC}$
3	PLLCFG, BOOTCFG, WKPCFG, $\overline{\text{RSTCFG}}$ setup time to $\overline{\text{RSTOUT}}$ valid	$t_{RCSU}$	10	—	$t_{CYC}$
4	PLLCFG, BOOTCFG, WKPCFG, $\overline{\text{RSTCFG}}$ hold time from $\overline{\text{RSTOUT}}$ valid	$t_{RCH}$	0	—	$t_{CYC}$

<sup>1</sup> Reset timing specified at:  $F_{SYS} = 132 \text{ MHz}$ ;  $V_{DDEH} = 3.0\text{--}5.25 \text{ V}$ ;  $V_{DD} = 1.35\text{--}1.65 \text{ V}$ ; and  $T_A = T_L \text{ to } T_H$ .

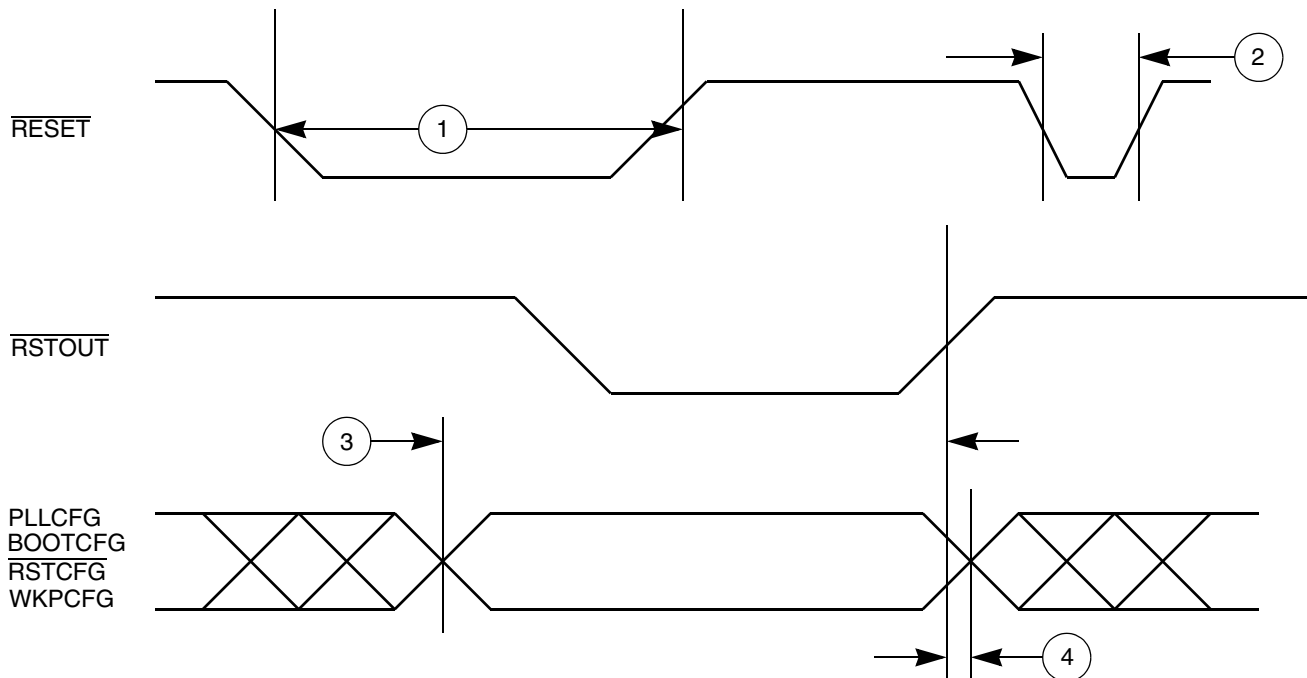


Figure 5. Reset and Configuration Pin Timing

### 3.13.2 IEEE 1149.1 Interface Timing

Table 20. JTAG Pin AC Electrical Characteristics<sup>1</sup>

Spec	Characteristic	Symbol	Min	Max	Unit
1	TCK cycle time	$t_{JCYC}$	100	—	ns
2	TCK clock pulse width (measured at $V_{DDE} \div 2$ )	$t_{JDC}$	40	60	ns
3	TCK rise and fall times (40% to 70%)	$t_{TCKRISE}$	—	3	ns
4	TMS, TDI data setup time	$t_{TMSS}, t_{TDIS}$	5	—	ns
5	TMS, TDI data hold time	$t_{TMSH}, t_{TDIH}$	25	—	ns
6	TCK low to TDO data valid	$t_{TDOV}$	—	20	ns
7	TCK low to TDO data invalid	$t_{TDOI}$	0	—	ns
8	TCK low to TDO high impedance	$t_{TDOHZ}$	—	20	ns
9	JCOMP assertion time	$t_{JCOMPW}$	100	—	ns
10	JCOMP setup time to TCK low	$t_{JCMPS}$	40	—	ns
11	TCK falling-edge to output valid	$t_{BSDV}$	—	50	ns
12	TCK falling-edge to output valid out of high impedance	$t_{BSDVZ}$	—	50	ns
13	TCK falling-edge to output high impedance (Hi-Z)	$t_{BSDHZ}$	—	50	ns
14	Boundary scan input valid to TCK rising-edge	$t_{BSDST}$	50	—	ns
15	TCK rising-edge to boundary scan input invalid	$t_{BSDHT}$	50	—	ns

<sup>1</sup> These specifications apply to JTAG boundary scan only. JTAG timing specified at  $V_{DD} = 1.35\text{--}1.65\text{ V}$ ,  $V_{DDE} = 3.0\text{--}3.6\text{ V}$ ,  $V_{DD33}$  and  $V_{DDSYN} = 3.0\text{--}3.6\text{ V}$ ,  $T_A = T_L$  to  $T_H$ , and  $CL = 30\text{ pF}$  with  $DSC = 0b10$ ,  $SRC = 0b11$ . Refer to Table 21 for functional specifications.

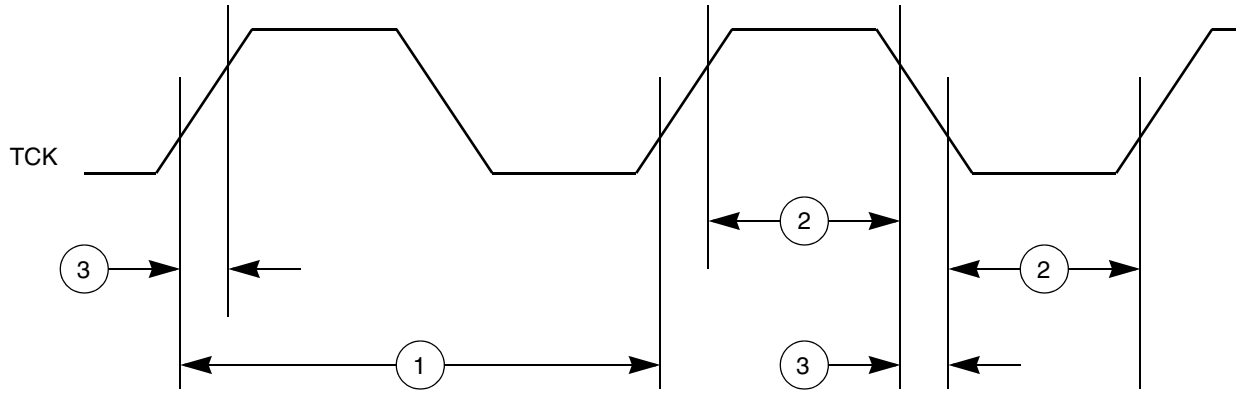


Figure 6. JTAG Test Clock Input Timing

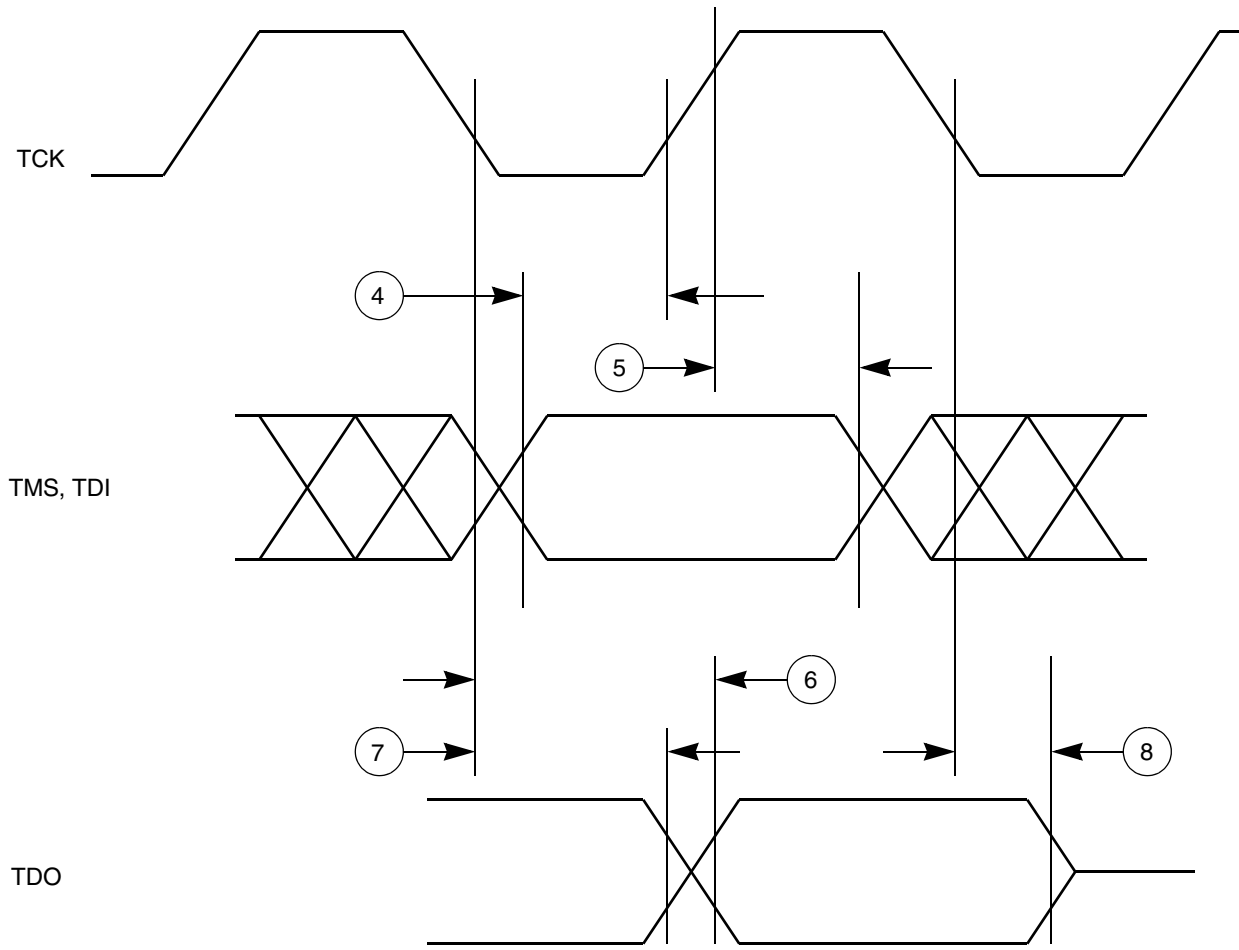


Figure 7. JTAG Test Access Port Timing

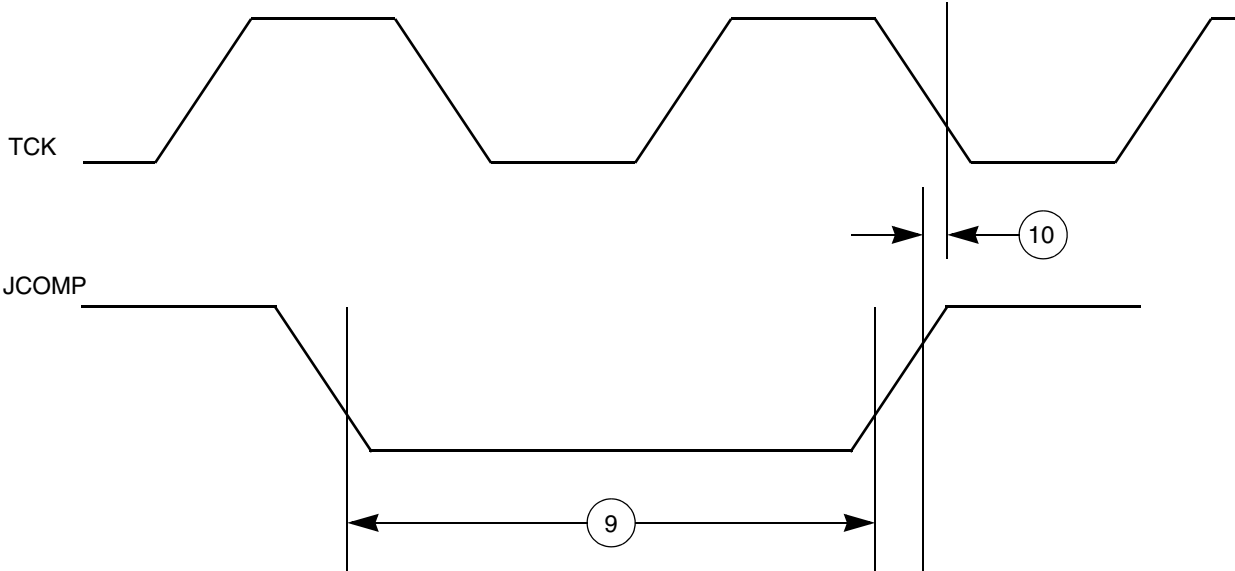


Figure 8. JTAG JCOMP Timing

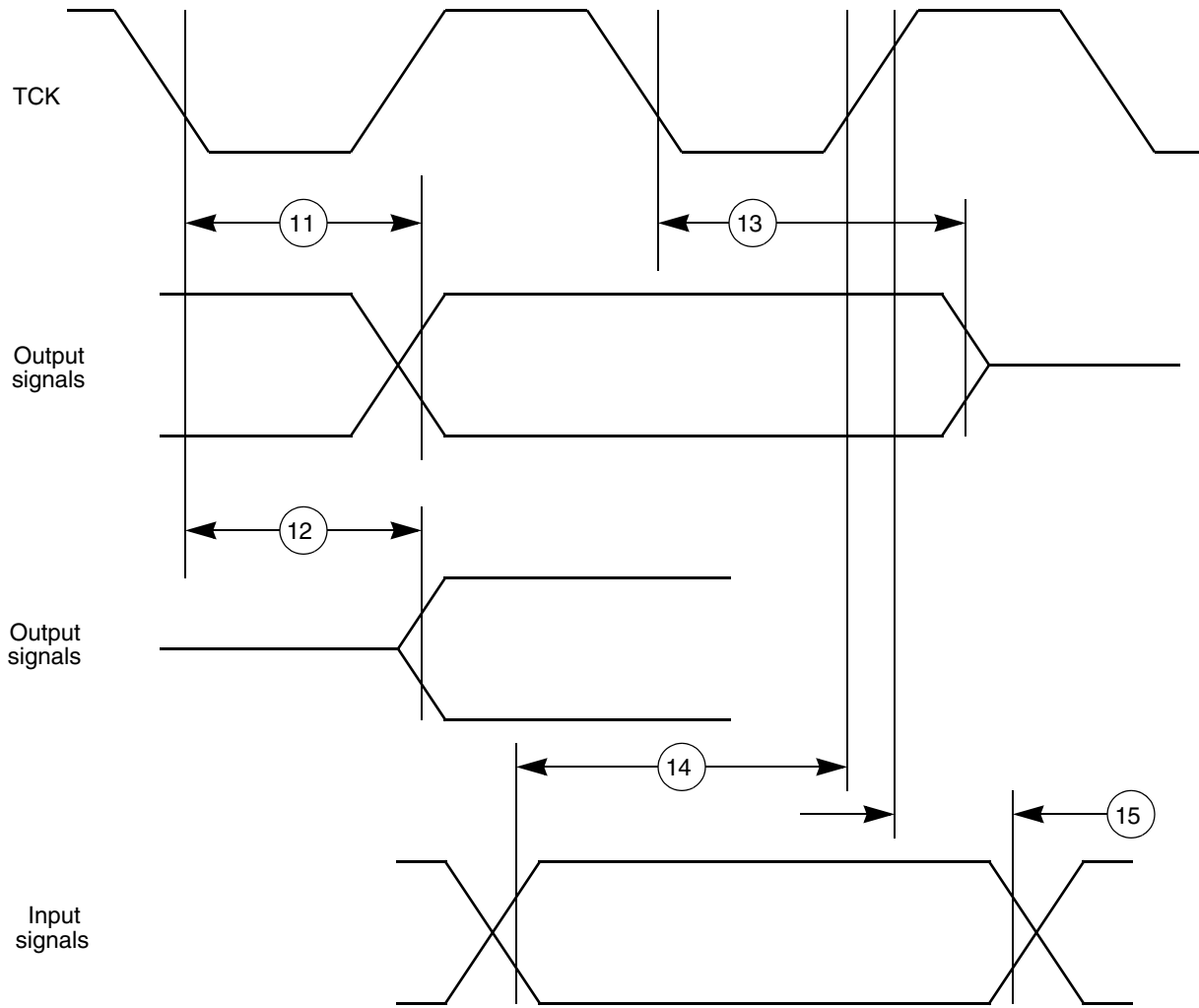


Figure 9. JTAG Boundary Scan Timing

### 3.13.3 Nexus Timing

Table 21. Nexus Debug Port Timing<sup>1</sup>

Spec	Characteristic	Symbol	Min	Max	Unit
1	MCKO cycle time	$t_{MICYC}$	1 <sup>2</sup>	8	$t_{CYC}$
2	MCKO duty cycle	$t_{MDC}$	40	60	%
3	MCKO low to MDO data valid <sup>3</sup>	$t_{MDOV}$	-1.5	3.0	ns
4	MCKO low to $\overline{MSEO}$ data valid <sup>3</sup>	$t_{MSEOV}$	-1.5	3.0	ns
5	MCKO low to $\overline{EVTO}$ data valid <sup>3</sup>	$t_{EVTOV}$	-1.5	3.0	ns
6	$\overline{EVTI}$ pulse width	$t_{EVTIPW}$	4.0	—	$t_{TCYC}$
7	$\overline{EVTO}$ pulse width	$t_{EVTOPW}$	1	—	$t_{MICYC}$
8	TCK cycle time	$t_{TCYC}$	4 <sup>4</sup>	—	$t_{CYC}$
9	TCK duty cycle	$t_{TDC}$	40	60	%
10	TDI, TMS data setup time	$t_{NTDIS}, t_{NTMSS}$	8	—	ns
11	TDI, TMS data hold time	$t_{NTDIH}, t_{NTMSH}$	5	—	ns
12	TCK low to TDO data valid	$t_{JOV}$	0	12	ns
	$V_{DDE} = 2.25\text{--}3.0\text{ V}$ $V_{DDE} = 3.0\text{--}3.6\text{ V}$		0	9	ns
13	$\overline{RDY}$ valid to MCKO <sup>5</sup>	—	—	—	—

<sup>1</sup> JTAG specifications apply when used for debug functionality. All Nexus timing relative to MCKO is measured from 50% of MCKO and 50% of the respective signal. Nexus timing specified at  $V_{DD} = 1.35\text{--}1.65\text{ V}$ ,  $V_{DDE} = 2.25\text{--}3.6\text{ V}$ ,  $V_{DD33}$  and  $V_{DDSYN} = 3.0\text{--}3.6\text{ V}$ ,  $T_A = T_L$  to  $T_H$ , and  $CL = 30\text{ pF}$  with  $DSC = 0b10$ .

<sup>2</sup> The Nexus AUX port runs up to 82 MHz. Set  $NPC\_PCR[MCKO\_DIV]$  to divide-by-two if the system frequency is greater than 82 MHz.

<sup>3</sup> MDO,  $\overline{MSEO}$ , and  $\overline{EVTO}$  data is held valid until the next MCKO low cycle occurs.

<sup>4</sup> Limit the maximum frequency to approximately 16 MHz ( $V_{DDE} = 2.25\text{--}3.0\text{ V}$ ) or 22 MHz ( $V_{DDE} = 3.0\text{--}3.6\text{ V}$ ) to meet the timing specification for  $t_{JOV}$  of  $[0.2 \times t_{JCYC}]$  as outlined in the IEEE-ISTO 5001-2003 specification.

<sup>5</sup> The  $\overline{RDY}$  pin timing is asynchronous to MCKO. The timing is guaranteed by design to function correctly.

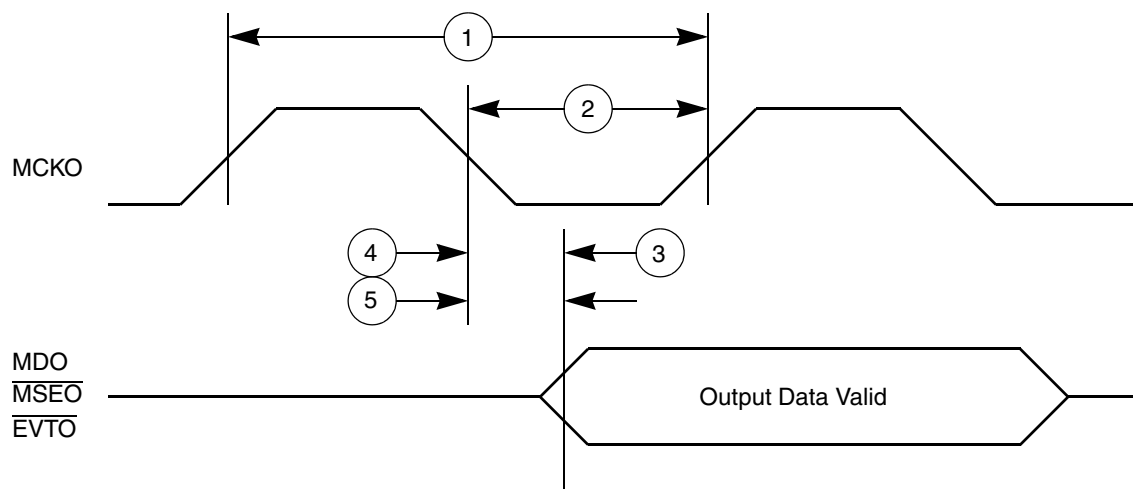


Figure 10. Nexus Output Timing

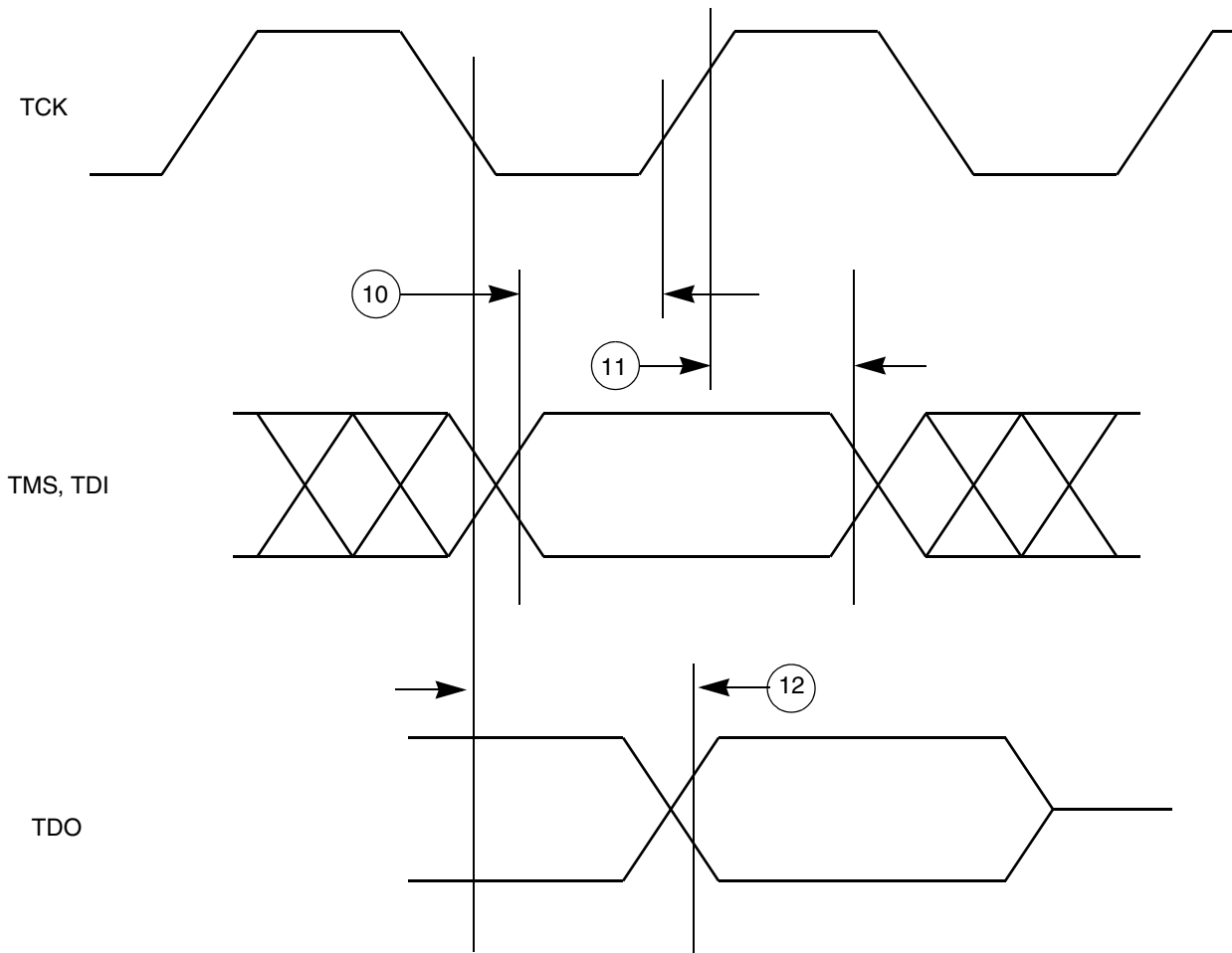


Figure 11. Nexus TDI, TMS, TDO Timing

### 3.13.4 External Bus Interface (EBI) Timing

Table 22. Bus Operation Timing<sup>1</sup>

Spec	Characteristic and Description	Symbol	40 MHz (ext. bus) <sup>2</sup>		56 MHz (ext. bus) <sup>2</sup>		66 MHz (ext. bus) <sup>2</sup>		Unit	Notes
			Min	Max	Min	Max	Min	Max		
1	CLKOUT period	$T_C$	25.0	—	17.9	—	15.2	—	ns	Signals are measured at 50% $V_{DDE}$ .
2	CLKOUT duty cycle	$t_{CDC}$	45%	55%	45%	55%	45%	55%	$T_C$	
3	CLKOUT rise time	$t_{CRT}$	—	— <sup>3</sup>	—	— <sup>3</sup>	—	— <sup>3</sup>	ns	
4	CLKOUT fall time	$t_{CFT}$	—	— <sup>3</sup>	—	— <sup>3</sup>	—	— <sup>3</sup>	ns	



Table 22. Bus Operation Timing<sup>1</sup> (continued)

Spec	Characteristic and Description	Symbol	40 MHz (ext. bus) <sup>2</sup>		56 MHz (ext. bus) <sup>2</sup>		66 MHz (ext. bus) <sup>2</sup>		Unit	Notes
			Min	Max	Min	Max	Min	Max		
5	CLKOUT positive edge to output signal invalid or Hi-Z (hold time)	$t_{COH}$	1.0 <sup>4</sup> 1.5	—	1.0 <sup>4</sup> 1.5	—	1.0 <sup>4</sup> 1.5	—	ns	EBTS=0 EBTS=1  Hold time selectable via SIU_ECCR[EBTS] bit.
	External bus interface ADDR[8:31] $\overline{BDIP}$ $\overline{CS}[0:3]$ DATA[0:31] $\overline{OE}$ RD_ $\overline{WR}$ $\overline{TA}$ $\overline{TEA}$ $\overline{TS}$ WE/ $\overline{BE}[0:3]$									
	Calibration bus interface CAL_ADDR[10:11, 27:30] CAL_ $\overline{CS}[0, 2:3]$ CAL_DATA[0:15] CAL_ $\overline{WE}/\overline{BE}[0:1]$	$t_{CCOH}$	1.0 <sup>5</sup> 1.5	—	1.0 <sup>4</sup> 1.5	—	1.0 <sup>4</sup> 1.5	—	ns	EBTS=0 EBTS=1  Hold time selectable via SIU_ECCR[EBTS] bit.
6	CLKOUT positive edge to output signal valid (output delay)	$t_{COV}$	—	10.0 <sup>4</sup> 11.0	—	7.5 <sup>4</sup> 8.5	—	6.0 <sup>4</sup> 7.0	ns	EBTS=0 EBTS=1  Output valid time selectable via SIU_ECCR[EBTS] bit.
	External bus interface ADDR[8:31] $\overline{BDIP}$ $\overline{CS}[0:3]$ DATA[0:31] $\overline{OE}$ RD_ $\overline{WR}$ $\overline{TA}$ $\overline{TEA}$ $\overline{TS}$ WE/ $\overline{BE}[0:3]$									
6a	CLKOUT positive edge to output signal valid (output delay)	$t_{CCOV}$	—	11.0 <sup>4</sup> 12.0	—	8.5 <sup>4</sup> 9.5	—	7.0 <sup>4</sup> 8.0	ns	EBTS=0 EBTS=1  Output valid time selectable via SIU_ECCR[EBTS] bit.
	Calibration bus interface CAL_ADDR[10:11, 27:30] CAL_ $\overline{CS}[0, 2:3]$ CAL_DATA[0:15] CAL_ $\overline{WE}/\overline{BE}[0:1]$									

Table 22. Bus Operation Timing<sup>1</sup> (continued)

Spec	Characteristic and Description	Symbol	40 MHz (ext. bus) <sup>2</sup>		56 MHz (ext. bus) <sup>2</sup>		66 MHz (ext. bus) <sup>2</sup>		Unit	Notes
			Min	Max	Min	Max	Min	Max		
7	Input signal valid to CLKOUT positive edge (setup time)	$t_{CIS}$	10.0	—	7.0	—	5.0	—	ns	
	External bus interface ADDR[8:31] BDIP DATA[0:31] $\overline{OE}$ $\overline{RD\_WR}$ $\overline{TA}$ $\overline{TEA}$ $\overline{TS}$ $\overline{WE/BE}[0:3]$									
7a	Input signal valid to CLKOUT positive edge (setup time)	$t_{CCIS}$	11.0	—	8.0	—	6.0	—	ns	
	Calibration bus interface CAL_ADDR[10:11, 27:30] CAL_CS[0, 2:3] CAL_DATA[0:15] CAL_WE/BE[0:1]									
8	CLKOUT positive edge to input signal invalid (hold time)	$t_{CIH}$	1.0	—	1.0	—	1.0	—	ns	
	External bus interface ADDR[8:31] BDIP DATA[0:31] $\overline{OE}$ $\overline{RD\_WR}$ $\overline{TA}$ $\overline{TEA}$ $\overline{TS}$ $\overline{WE/BE}[0:3]$									
	Calibration bus interface CAL_ADDR[10:11, 27:30] CAL_CS[0, 2:3] CAL_DATA[0:15] CAL_WE/BE[0:1]	$t_{CIIH}$	1.0	—	1.0	—	1.0	—	ns	

<sup>1</sup> EBI timing specified at  $V_{DD} = 1.35\text{--}1.65\text{ V}$ ,  $V_{DDE} = 1.6\text{--}3.6\text{ V}$  (unless stated otherwise),  $V_{DD33}$  and  $V_{DDSYN} = 3.0\text{--}3.6\text{ V}$ ,  $T_A = T_L$  to  $T_H$ , and  $CL = 30\text{ pF}$  with  $DSC = 0b10$ .

<sup>2</sup> The external bus is limited to half the speed of the internal bus.

<sup>3</sup> Refer to fast pad timing in Table 17 and Table 18 (different values for 1.8 V and 3.3 V).

<sup>4</sup> The EBTS = 0 timings are tested and valid at  $V_{DDE} = 2.25\text{--}3.6\text{ V}$  only, whereas EBTS = 1 timings are tested and valid at  $V_{DDE} = 1.6\text{--}3.6\text{ V}$ .

<sup>5</sup> The EBTS = 0 timings are tested and valid at  $V_{DDE} = 2.25\text{--}3.6\text{ V}$  only, whereas EBTS = 1 timings are tested and valid at  $V_{DDE} = 1.6\text{--}3.6\text{ V}$ .

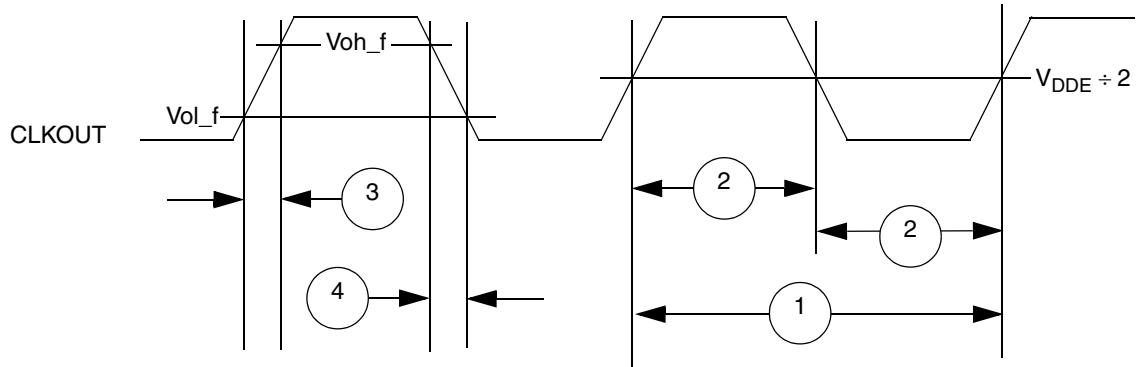


Figure 12. CLKOUT Timing

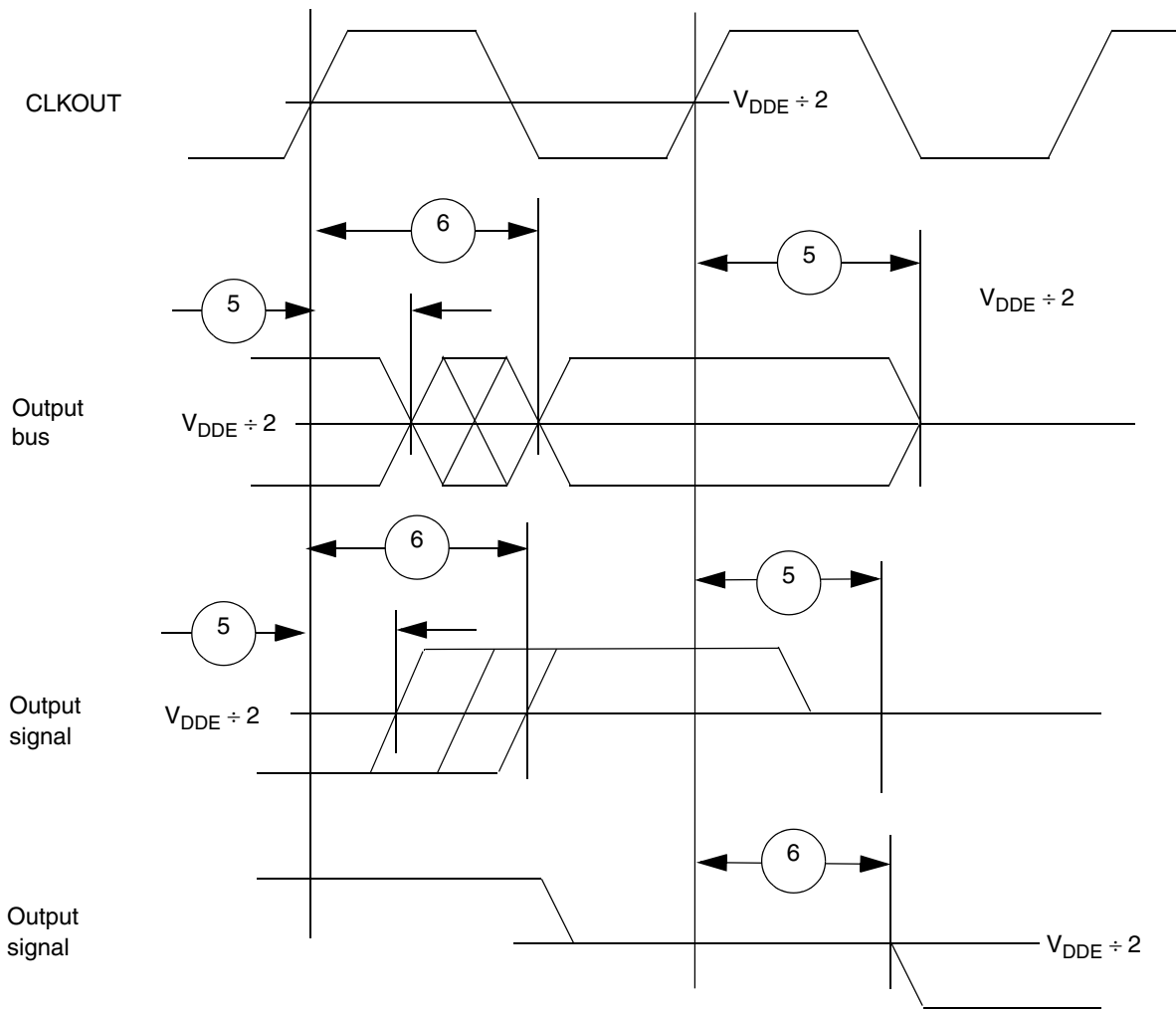


Figure 13. Synchronous Output Timing

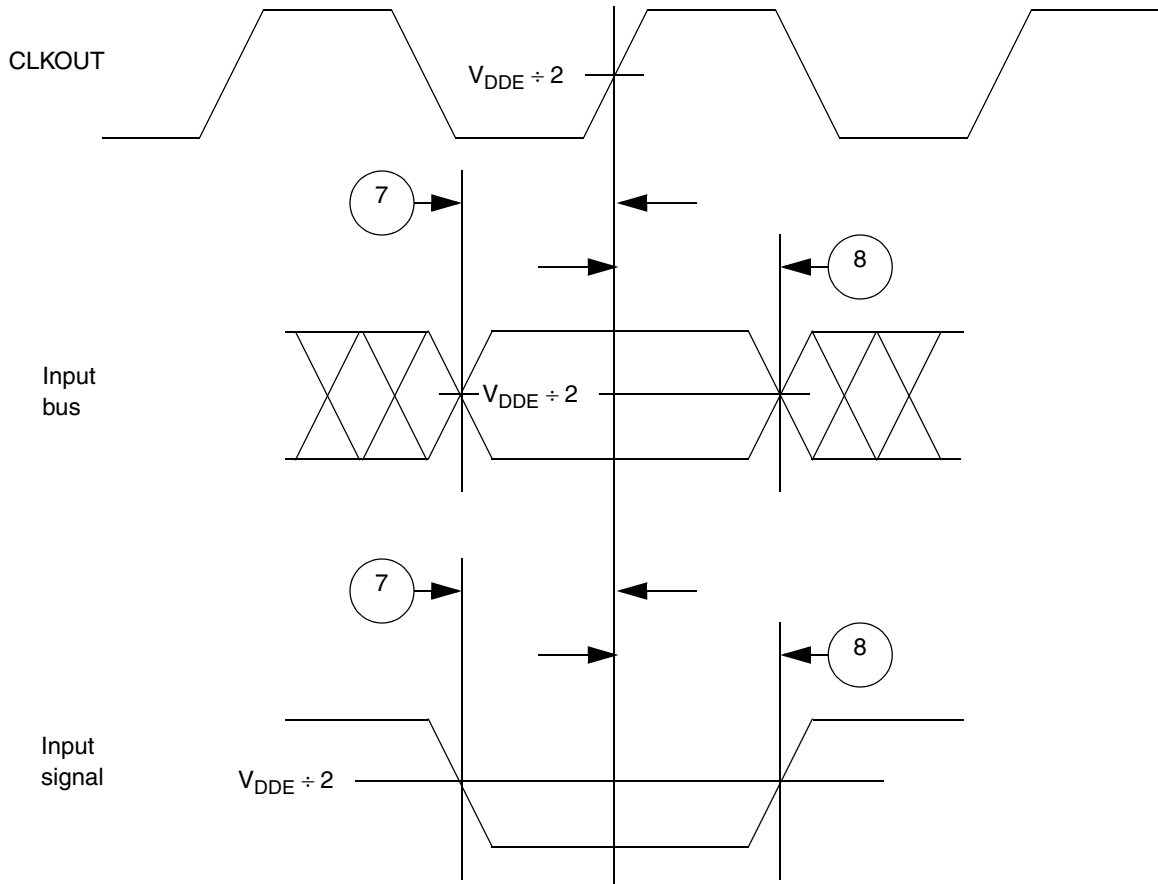


Figure 14. Synchronous Input Timing

### 3.13.5 External Interrupt Timing (IRQ Signals)

Table 23. External Interrupt Timing<sup>1</sup>

Spec	Characteristic	Symbol	Min	Max	Unit
1	IRQ pulse-width low	$t_{IPWL}$	3	—	$t_{CYC}$
2	IRQ pulse-width high	$T_{IPWH}$	3	—	$t_{CYC}$
3	IRQ edge-to-edge time <sup>2</sup>	$t_{ICYC}$	6	—	$t_{CYC}$

<sup>1</sup> IRQ timing specified at  $F_{SYS} = 132 \text{ MHz}$ ,  $V_{DD} = 1.35\text{--}1.65 \text{ V}$ ,  $V_{DDEH} = 3.0\text{--}5.5 \text{ V}$ ,  $V_{DD33}$  and  $V_{DDSYN} = 3.0\text{--}3.6 \text{ V}$ ,  $T_A = T_L$  to  $T_H$ , and  $CL = 200\text{pF}$  with  $SRC = 0b11$ .

<sup>2</sup> Applies when IRQ signals are configured for rising-edge or falling-edge events, but not both.

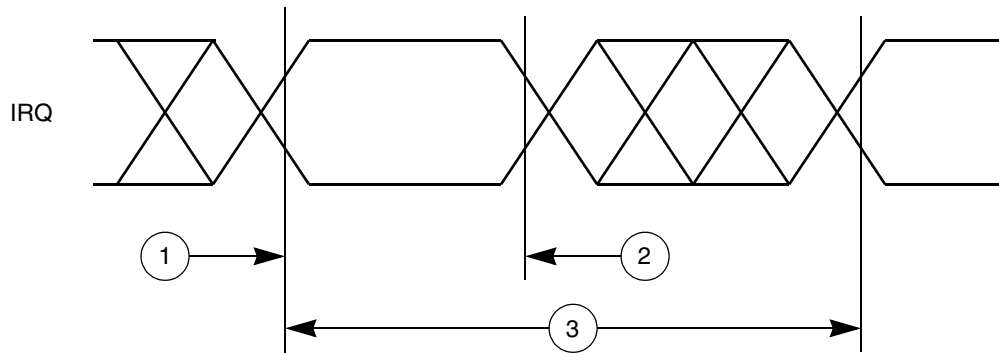


Figure 15. External Interrupt Timing

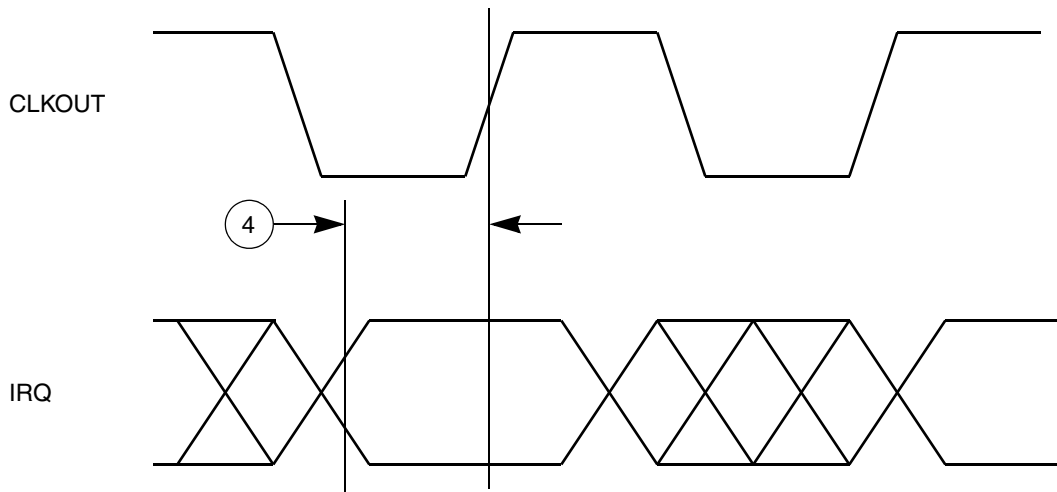


Figure 16. External Interrupt Setup Timing

### 3.13.6 eTPU Timing

Table 24. eTPU Timing<sup>1</sup>

Spec	Characteristic	Symbol	Min	Max	Unit
1	eTPU input channel pulse width	$t_{ICPW}$	4	—	$t_{CYC}$
2	eTPU output channel pulse width	$t_{OCPW}$	2	—	$t_{CYC}$

<sup>1</sup> eTPU timing specified at  $F_{SYS} = 132 \text{ MHz}$ ,  $V_{DD} = 1.35\text{--}1.65 \text{ V}$ ,  $V_{DDEH} = 3.0\text{--}5.5 \text{ V}$ ,  $V_{DD33}$  and  $V_{DDSYN} = 3.0\text{--}3.6 \text{ V}$ ,  $T_A = T_L$  to  $T_H$ , and  $CL = 200 \text{ pF}$  with  $SRC = 0b11$ .

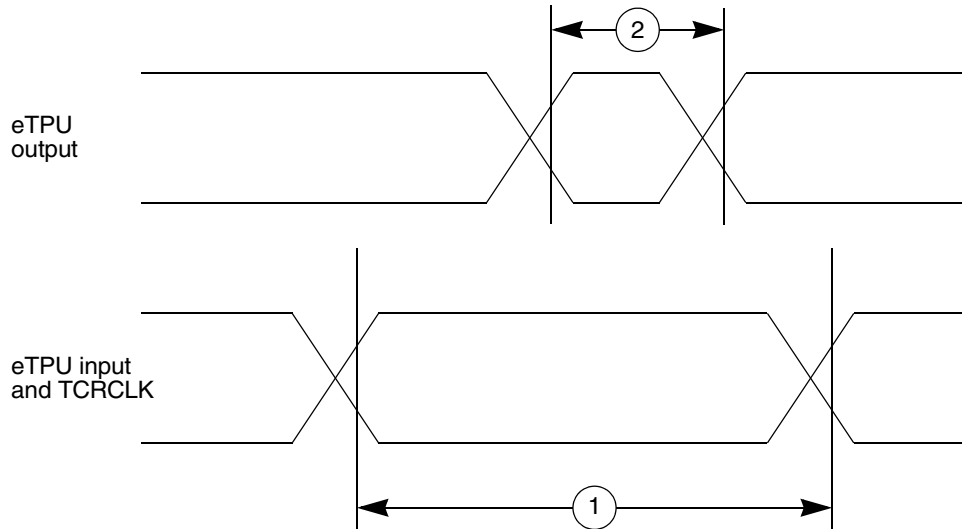


Figure 17. eTPU Timing

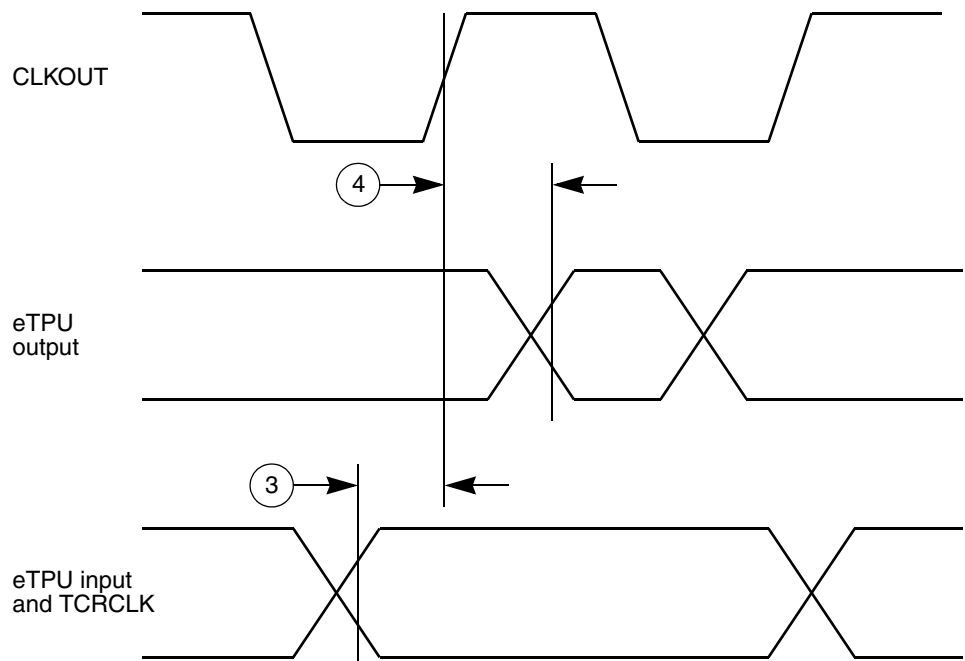


Figure 18. eTPU Input/Output Timing

### 3.13.7 eMIOS (MTS) Timing

Table 25. MTS Timing<sup>1</sup>

Spec	Characteristic	Symbol	Min	Max	Unit
1	eMIOS (MTS) input pulse width	$t_{MIPW}$	4	—	$t_{CYC}$
2	eMIOS (MTS) output pulse width	$t_{MOPW}$	1	—	$t_{CYC}$

<sup>1</sup> MTS timing specified at  $F_{SYS} = 132$  MHz,  $V_{DD} = 1.35$ – $1.65$  V,  $V_{DDEH} = 3.0$ – $5.5$  V,  $V_{DD33}$  and  $V_{DDSYN} = 3.0$ – $3.6$  V,  $T_A = T_L$  to  $T_H$ , and  $CL = 50$  pF with  $SRC = 0b11$ .

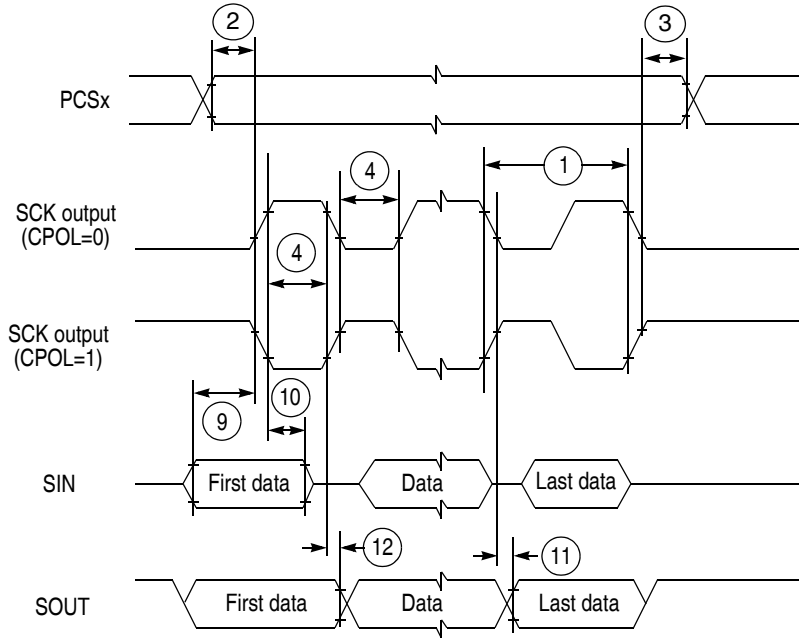
### 3.13.8 DSPI Timing

Table 26. DSPI Timing<sup>1</sup>

Spec	Characteristic	Symbol	80 MHz		112 MHz		132 MHz		Unit
			Min	Max	Min	Max	Min	Max	
1	SCK cycle time <sup>2,3</sup>	$t_{SCK}$	25 ns	2.9 ms	17.9 ns	2.0 ms	15.2 ns	1.7 ms	—
2	PCS to SCK delay <sup>4</sup>	$t_{CSC}$	23	—	15	—	13	—	ns
3	After SCK delay <sup>5</sup>	$t_{ASC}$	22	—	14	—	12	—	ns
4	SCK duty cycle	$t_{SDC}$	$(t_{SCK} \div 2) - 2$ ns	$(t_{SCK} \div 2) + 2$ ns	—	—	—	—	ns
5	Slave access time ( $\overline{SS}$ active to SOUT driven)	$t_A$	—	25	—	25	—	25	ns
6	Slave SOUT disable time ( $\overline{SS}$ inactive to SOUT Hi-Z, or invalid)	$t_{DIS}$	—	25	—	25	—	25	ns
7	PCSx to $\overline{PCSS}$ time	$t_{PCSC}$	4	—	4	—	4	—	ns
8	$\overline{PCSS}$ to PCSx time	$t_{PASC}$	5	—	5	—	5	—	ns
9	Data setup time for inputs Master (MTFE = 0)	$t_{SUI}$	20	—	20	—	20	—	ns
	Slave		2	—	2	—	2	—	ns
	Master (MTFE = 1, CPHA = 0) <sup>6</sup>		–4	—	3	—	6	—	ns
	Master (MTFE = 1, CPHA = 1)		20	—	20	—	20	—	ns
10	Data hold time for inputs Master (MTFE = 0)	$t_{HI}$	–4	—	–4	—	–4	—	ns
	Slave		7	—	7	—	7	—	ns
	Master (MTFE = 1, CPHA = 0) <sup>6</sup>		21	—	14	—	12	—	ns
	Master (MTFE = 1, CPHA = 1)		–4	—	–4	—	–4	—	ns
11	Data valid (after SCK edge) Master (MTFE = 0)	$t_{SUO}$	—	5	—	5	—	5	ns
	Slave		—	25	—	25	—	25	ns
	Master (MTFE = 1, CPHA = 0)		—	18	—	14	—	13	ns
	Master (MTFE = 1, CPHA = 1)		—	5	—	5	—	5	ns
12	Data hold time for outputs Master (MTFE = 0)	$t_{HO}$	–5	—	–5	—	–5	—	ns
	Slave		5.5	—	5.5	—	5.5	—	ns
	Master (MTFE = 1, CPHA = 0)		8	—	4	—	3	—	ns
	Master (MTFE = 1, CPHA = 1)		–5	—	–5	—	–5	—	ns

## Electrical Characteristics

- <sup>1</sup> All DSPI timing specifications use the fastest slew rate (SRC = 0b11) on pad type M or MH. DSPI signals using pad types of S or SH have an additional delay based on the slew rate. DSPI timing is specified at  $V_{DD} = 1.35\text{--}1.65\text{ V}$ ,  $V_{DDEH} = 3.0\text{--}5.5\text{ V}$ ,  $V_{DD33}$  and  $V_{DDSYN} = 3.0\text{--}3.6\text{ V}$ ,  $T_A = T_L$  to  $T_H$ , and  $CL = 50\text{ pF}$  with SRC = 0b11.
- <sup>2</sup> The minimum SCK cycle time restricts the baud rate selection for the given system clock rate. These numbers are calculated based on two MPC55xx devices communicating over a DSPI link.
- <sup>3</sup> The actual minimum SCK cycle time is limited by pad performance.
- <sup>4</sup> The maximum value is programmable in DSPI\_CTARx[PSSCK] and DSPI\_CTARx[CSSCK].
- <sup>5</sup> The maximum value is programmable in DSPI\_CTARx[PASC] and DSPI\_CTARx[ASC].
- <sup>6</sup> This number is calculated using the SMPL\_PT bit field in DSPI\_MCR set to 0b10.



**Figure 19. DSPI Classic SPI Timing—Master, CPHA = 0**



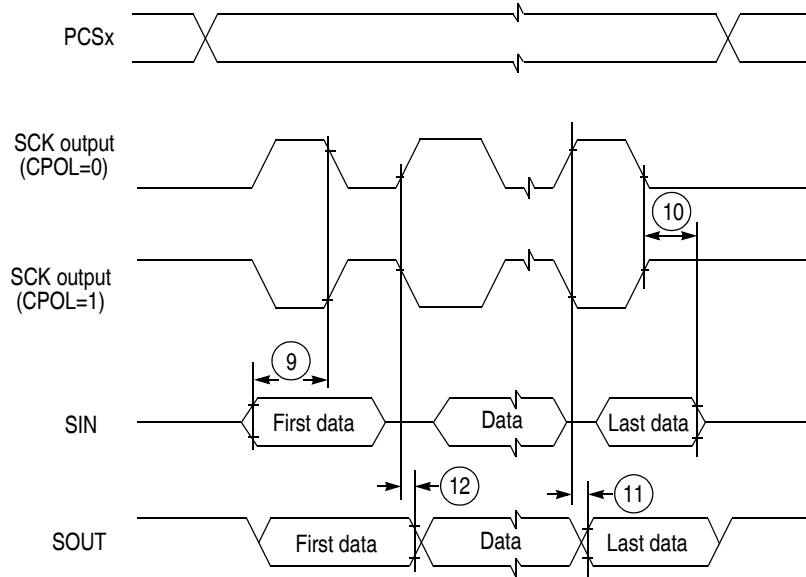


Figure 20. DSPI Classic SPI Timing—Master, CPHA = 1

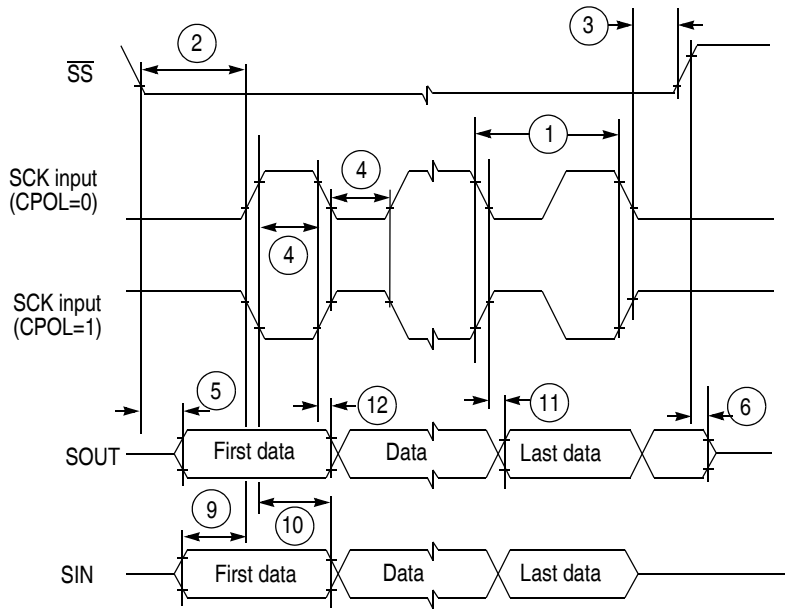


Figure 21. DSPI Classic SPI Timing—Slave, CPHA = 0

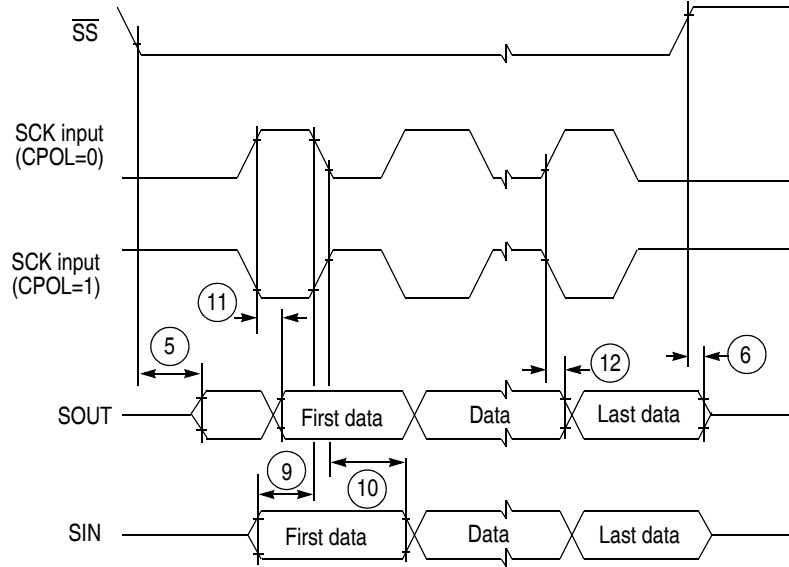


Figure 22. DSPI Classic SPI Timing—Slave, CPHA = 1

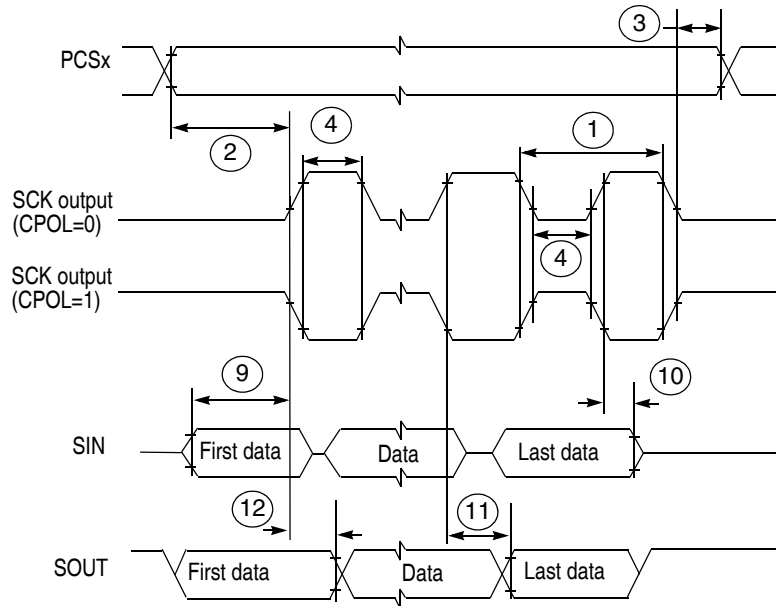


Figure 23. DSPI Modified Transfer Format Timing—Master, CPHA = 0

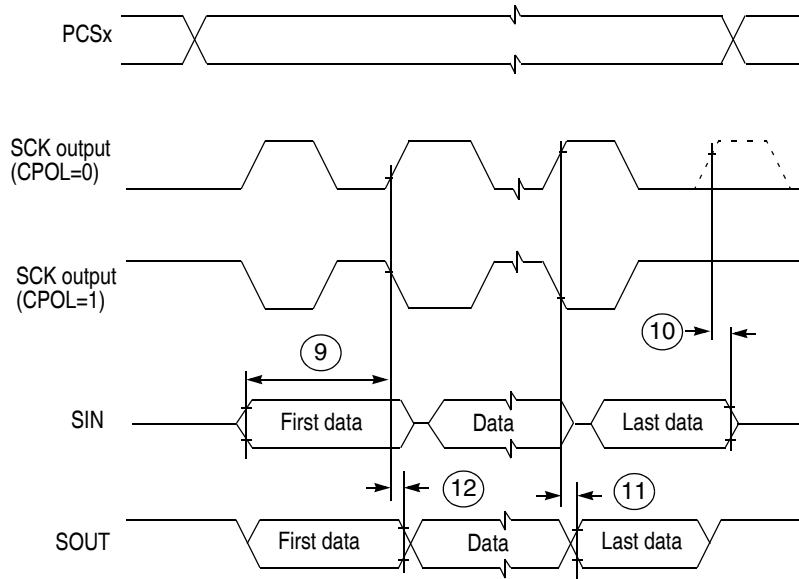


Figure 24. DSPI Modified Transfer Format Timing—Master, CPHA = 1

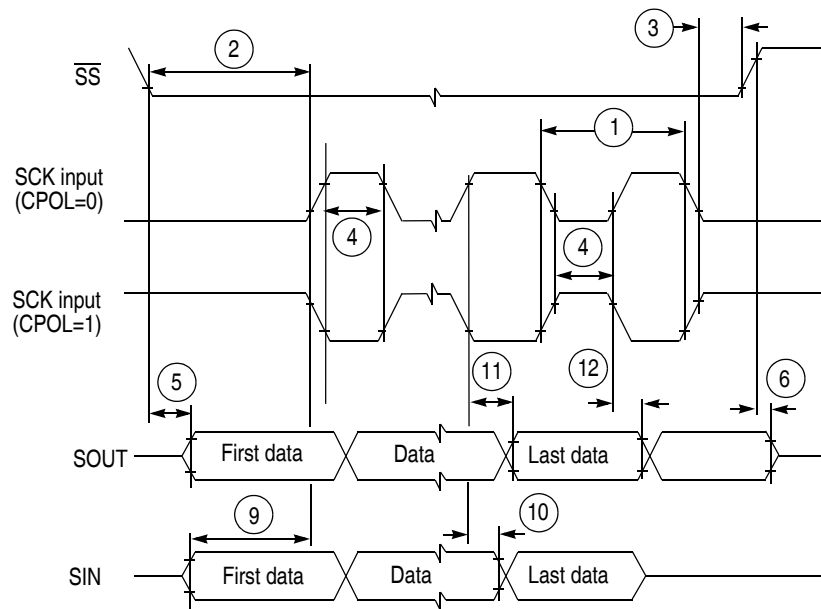


Figure 25. DSPI Modified Transfer Format Timing—Slave, CPHA = 0

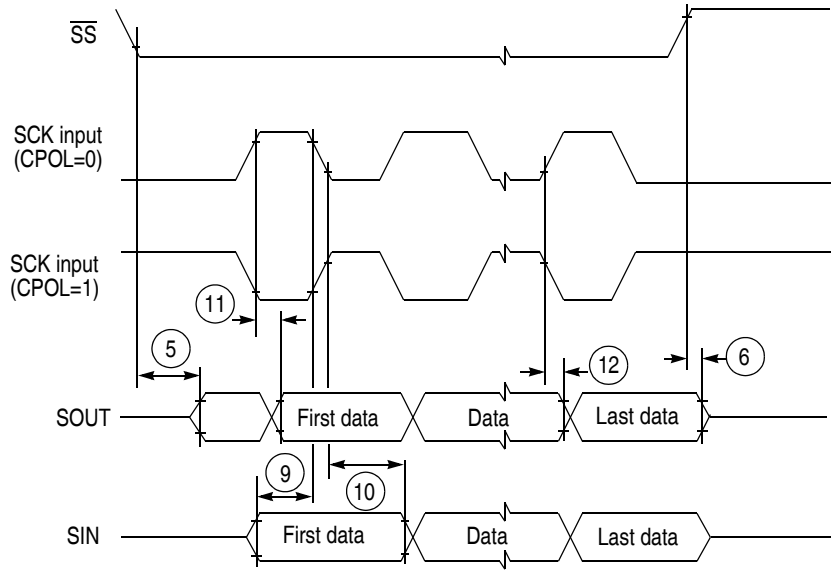


Figure 26. DSPI Modified Transfer Format Timing—Slave, CPHA =1

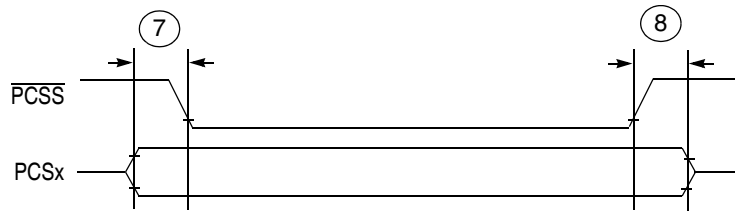


Figure 27. DSPI PCS Strobe ( $\overline{PCSS}$ ) Timing

### 3.13.9 eQADC SSI Timing

Table 27. EQADC SSI Timing Characteristics (Pads at 3.3 V or 5.0 V)

CLOAD = 25 pF on all outputs. Pad drive strength set to maximum.

Spec	Rating	Symbol	Minimum	Typical	Maximum	Unit
2	FCK period ( $t_{FCK} = 1 \div f_{FCK}$ ) <sup>1, 2</sup>	$t_{FCK}$	2	—	17	$t_{SYS\_CLK}$
3	Clock (FCK) high time	$t_{FCKHT}$	$t_{SYS\_CLK} - 6.5$	—	$9 \times (t_{SYS\_CLK} + 6.5)$	ns
4	Clock (FCK) low time	$t_{FCKLT}$	$t_{SYS\_CLK} - 6.5$	—	$8 \times (t_{SYS\_CLK} + 6.5)$	ns
5	SDS lead / lag time	$t_{SDS\_LL}$	-7.5	—	+7.5	ns
6	SDO lead / lag time	$t_{SDO\_LL}$	-7.5	—	+7.5	ns
7	EQADC data setup time (inputs)	$t_{EQ\_SU}$	22	—	—	ns
8	EQADC data hold time (inputs)	$t_{EQ\_HO}$	1	—	—	ns

<sup>1</sup> SS timing specified at  $F_{SYS} = 132$  MHz,  $V_{DD} = 1.35-1.65$  V,  $V_{DDEH} = 3.0-5.5$  V,  $V_{DD33}$  and  $V_{DDSYN} = 3.0-3.6$  V,  $T_A = T_L$  to  $T_H$ , and  $CL = 50$  pF with  $SRC = 0b11$ . Maximum operating frequency varies depending on track delays, master pad delays, and slave pad delays.

<sup>2</sup> FCK duty is not 50% when it is generated through the division of the system clock by an odd number.

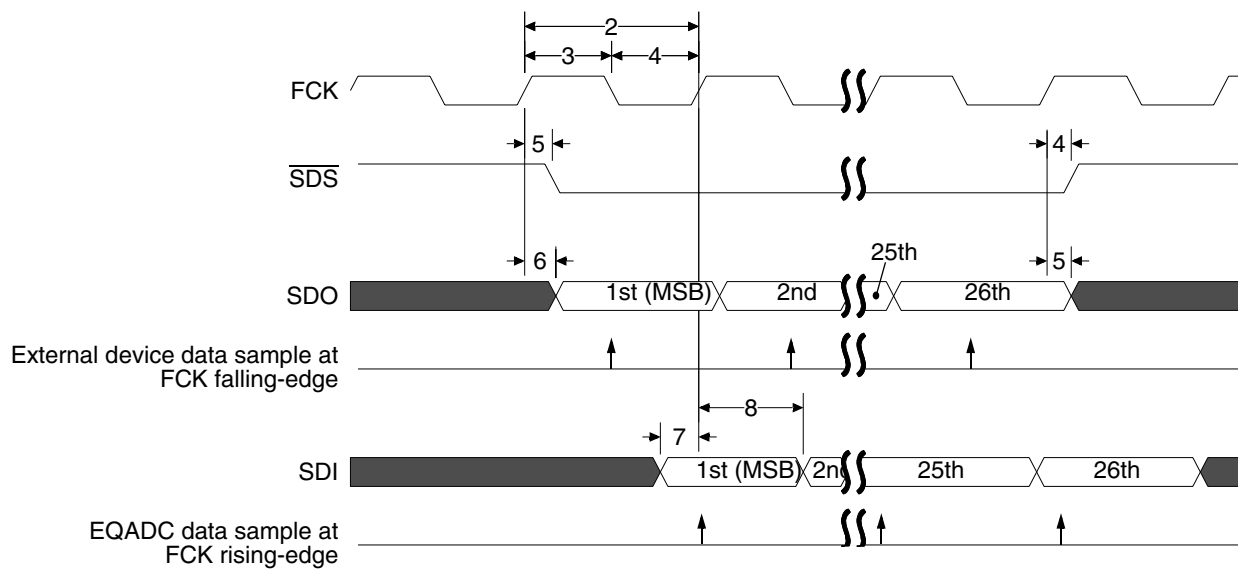


Figure 28. EQADC SSI Timing

### 3.14 Fast Ethernet AC Timing Specifications

Media Independent Interface (MII) Fast Ethernet Controller (FEC) signals use transistor-to-transistor logic (TTL) signal levels compatible with devices operating at 3.3 V. The timing specifications for the MII FEC signals are independent of the system clock frequency (part speed designation).

#### 3.14.1 MII FEC Receive Signal Timing FEC\_RXD[3:0], FEC\_RX\_DV, FEC\_RX\_ER, and FEC\_RX\_CLK

The receive functions correctly up to an FEC\_RX\_CLK maximum frequency of 25 MHz plus one percent. There is no minimum frequency requirement. The processor clock frequency must exceed four times the FEC\_RX\_CLK frequency.

Table 28 lists MII FEC receive channel timings.

Table 28. MII FEC Receive Signal Timing

Spec	Characteristic	Min	Max	Unit
1	FEC_RXD[3:0], FEC_RX_DV, FEC_RX_ER to FEC_RX_CLK setup	5	—	ns
2	FEC_RX_CLK to FEC_RXD[3:0], FEC_RX_DV, FEC_RX_ER hold	5	—	ns
3	FEC_RX_CLK pulse-width high	35%	65%	FEC_RX_CLK period
4	FEC_RX_CLK pulse-width low	35%	65%	FEC_RX_CLK period

Figure 29 shows MII FEC receive signal timings listed in Table 28.

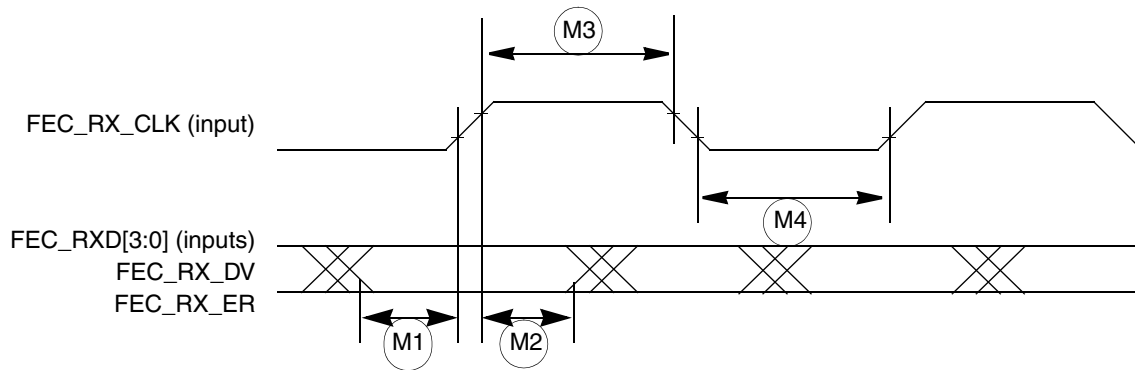


Figure 29. MII FEC Receive Signal Timing Diagram

### 3.14.2 MII FEC Transmit Signal Timing FEC\_TXD[3:0], FEC\_TX\_EN, FEC\_TX\_ER, FEC\_TX\_CLK

The transmitter functions correctly up to the FEC\_TX\_CLK maximum frequency of 25 MHz plus one percent. There is no minimum frequency requirement. In addition, the processor clock frequency must exceed twice the FEC\_TX\_CLK frequency.

The transmit outputs (FEC\_TXD[3:0], FEC\_TX\_EN, FEC\_TX\_ER) can be programmed to transition from either the rising- or falling-edge of TX\_CLK, and the timing is the same in either case. These options allow the use of non-compliant MII PHYs.

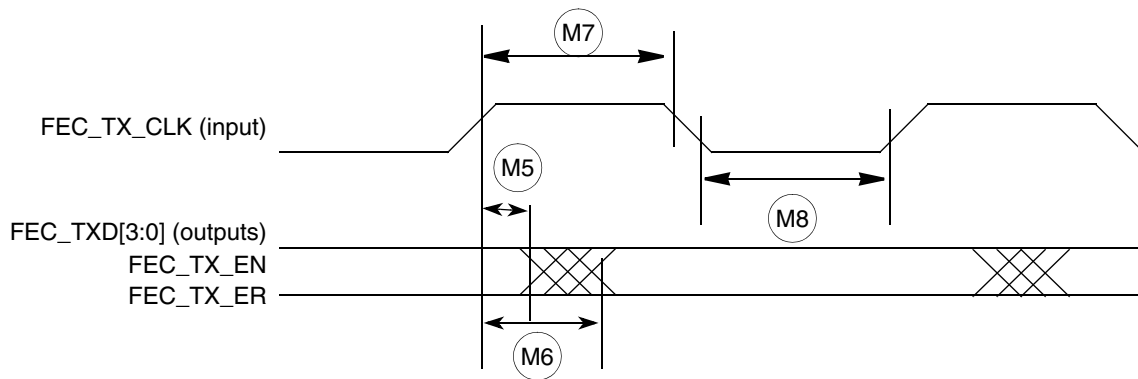
Refer to the Fast Ethernet Controller (FEC) chapter of the device reference manual for details of this option and how to enable it.

Table 29 lists MII FEC transmit channel timings.

**Table 29. MII FEC Transmit Signal Timing**

Spec	Characteristic	Min	Max	Unit
5	FEC_TX_CLK to FEC_TXD[3:0], FEC_TX_EN, FEC_TX_ER invalid	5	—	ns
6	FEC_TX_CLK to FEC_TXD[3:0], FEC_TX_EN, FEC_TX_ER valid	—	25	ns
7	FEC_TX_CLK pulse-width high	35%	65%	FEC_TX_CLK period
8	FEC_TX_CLK pulse-width low	35%	65%	FEC_TX_CLK period

Figure 30 shows MII FEC transmit signal timings listed in Table 29.



**Figure 30. MII FEC Transmit Signal Timing Diagram**

### 3.14.3 MII FEC Asynchronous Inputs Signal Timing FEC\_CRIS and FEC\_COL

Table 30 lists MII FEC asynchronous input signal timing.

Table 30. MII FEC Asynchronous Inputs Signal Timing

Spec	Characteristic	Min	Max	Unit
9	FEC_CRIS, FEC_COL minimum pulse width	1.5	—	FEC_TX_CLK period

Figure 31 shows MII FEC asynchronous input timing listed in Table 30.

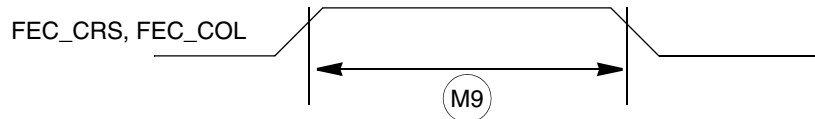


Figure 31. MII FEC Asynchronous Inputs Timing Diagram

### 3.14.4 MII FEC Serial Management Channel Timing FEC\_MDIO and FEC\_MDC

Table 31 lists MII FEC serial management channel timings. The FEC functions correctly with a maximum FEC\_MDC frequency of 2.5 MHz.

Table 31. MII FEC Serial Management Channel Timing

Spec	Characteristic	Min	Max	Unit
10	FEC_MDC falling-edge to FEC_MDIO output invalid (minimum propagation delay)	0	—	ns
11	FEC_MDC falling-edge to FEC_MDIO output valid (maximum propagation delay)	—	25	ns
12	FEC_MDIO (input) to FEC_MDC rising-edge setup	10	—	ns
13	FEC_MDIO (input) to FEC_MDC rising-edge hold	0	—	ns
14	FEC_MDC pulse-width high	40%	60%	FEC_MDC period
15	FEC_MDC pulse-width low	40%	60%	FEC_MDC period

Figure 32 shows MII FEC serial management channel timings listed in Table 31.



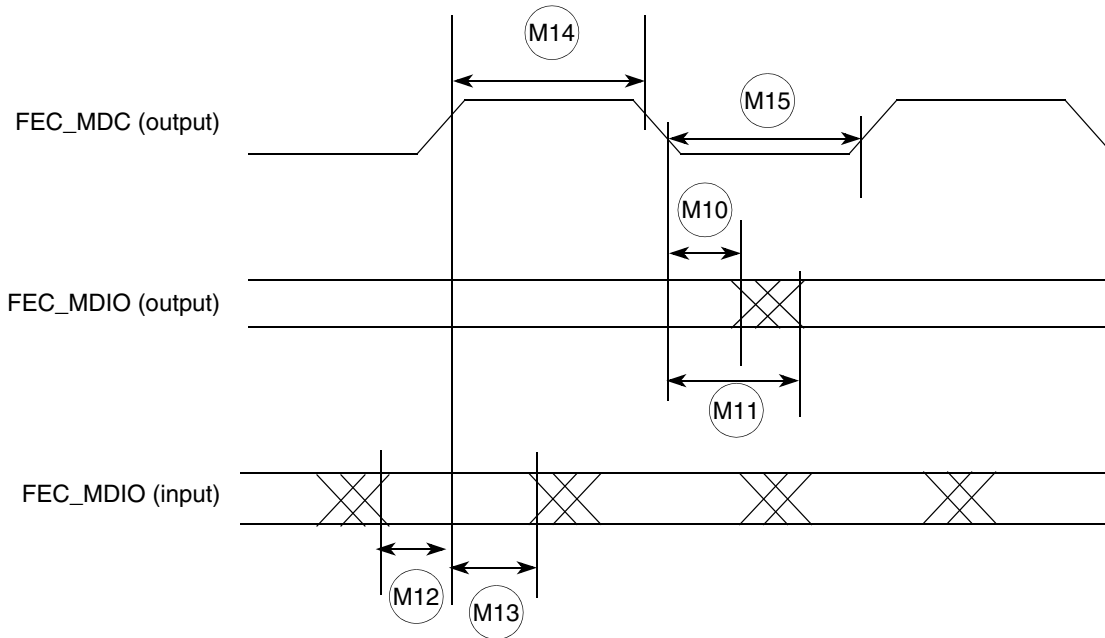


Figure 32. MII FEC Serial Management Channel Timing Diagram

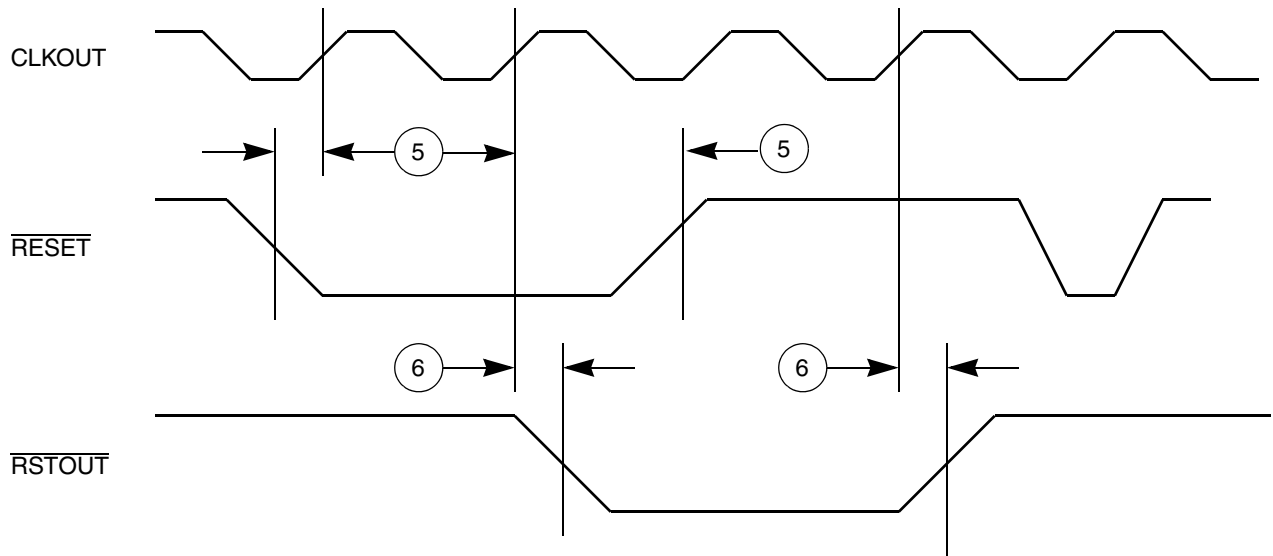


Figure 33. Reset and Configuration Pin Timing

## 4 Mechanicals

### 4.1 Pinouts

#### 4.1.1 MPC5553 416 PBGA Pinout

Figure 34, Figure 35, and Figure 36 show the pinout for the MPC5553 416 PBGA package. While the MPC5553 and the MPC5554/MPC5565/MPC5566 are pin-compatible, the MPC5553 BGA is shown to highlight the balls that are not connected to any signal on the MPC5553 (the eTPUB[0:31] and TSIZ[0:1]). The alternate Fast Ethernet Controller (FEC) signals that are multiplexed with the data bus are not shown for the MPC5553.

#### NOTE

Some pins have names that include functions that are not available on all MPC55xx devices. For example, ball R25 of the 416 BGA package is named ‘SINA,’ but the MPC5553 does not have a DSPI A module. In this case, the SINA pin can only be used for its alternate functions of GPIO[94] or PCSC[2]. Refer to the specific device Reference Manual for functions available on each device.

	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18	19	20	21	22	23	24	25	26		
A	VSS	VSTBY	AN37	AN11	VDDA1	AN16	AN1	AN5	VRH	AN23	AN27	AN28	AN35	VSSA0	AN15	ETRIG1	NC_1	NC_2	NC_3	NC_4	GPIO205	MDO11	MD08	VDD	VDD33	VSS	A	
B	VDD	VSS	AN36	AN39	AN19	AN20	AN0	AN4	REFBYPC	AN22	AN26	AN31	AN32	VSSA0	AN14	ETRIG0	NC_5	NC_6	NC_7	NC_8	MDO10	MDO7	MD04	MD00	VSS	VDDE7	B	
C	VDD33	VDD	VSS	AN8	AN17	VSSA1	AN21	AN3	AN7	VRL	AN25	AN30	AN33	VDDA0	AN13	NC_9	NC_10	NC_11	NC_12	MDO9	MDO6	MDO3	MD01	VSS	VDDE7	VDD	C	
D	ETPUA30	ETPUA31	VDD	VSS	AN38	AN9	AN10	AN18	AN2	AN6	AN24	AN29	AN34	VDDEH9	AN12	NC_13	NC_14	NC_15	NC_16	MDO5	MDO2	VDDEH8	VSS	VDDE7	TCK	TDI	D	
E	ETPUA28	ETPUA29	VDDEH1	VDD																			VDDE7	TMS	TDO	TEST	E	
F	ETPUA24	ETPUA27	ETPUA26	VDDEH1																			MSE00	JCOMP	EVTI	EVTO	F	
G	ETPUA23	ETPUA22	ETPUA25	ETPUA21																			MSE01	MCKO	GPIO204	NC_17	G	
H	ETPUA20	ETPUA19	ETPUA18	ETPUA17																			RDY	GPIO203	NC_18	NC_19	H	
J	ETPUA16	ETPUA15	ETPUA14	ETPUA13																			VDDEH6	NC_20	NC_21	NC_22	J	
K	ETPUA12	ETPUA11	ETPUA10	ETPUA9																			NC_23	NC_24	NC_25	NC_26	K	
L	ETPUA8	ETPUA7	ETPUA6	ETPUA5																			NC_27	NC_28	NC_29	NC_30	L	
M	ETPUA4	ETPUA3	ETPUA2	ETPUA1																			NC_31	NC_32	NC_33	SINB	M	
N	BDIP	TEA	ETPUA0	TCRCLKA																			SOUTB	PCSB3	PCSB0	PCSB1	N	
P	CS3	CS2	CS1	CS0																			PCSA3	PCSB4	SCKB	PCSB2	P	
R	WE3	WE2	WE1	WE0																			PCSB5	SOUTA	SINA	SCKA	R	
T	VDDE2	NC_34	RD_WR	VDDE2																			PCSA1	PCSA0	PCSA2	VPP	T	
U	ADDR16	NC_35	TA	VDD33																			PCSA4	TXDA	PCSA5	VFLASH	U	
V	ADDR18	ADDR17	TS	ADDR8																			CNTXC	RXDA	RSTOUT	RSTCFG	V	
W	ADDR20	ADDR19	ADDR9	ADDR10																				RXDB	CNRXC	TXDB	RESET	W
Y	ADDR22	ADDR21	ADDR11	VDDE2																				WKP_CFG	BOOT_CFG1	VRC_VSS	VSS_SYN	Y
AA	ADDR24	ADDR23	ADDR13	ADDR12																				VDDEH6	PLL_CFG1	BOOT_CFG0	EXTAL	AA
AB	VDDE2	ADDR25	ADDR15	ADDR14																				VDD	VRC_CTL	PLL_CFG0	XTAL	AB
AC	ADDR26	ADDR27	ADDR31	VSS	VDD	DATA26	DATA28	VDDE2	DATA30	DATA31	DATA8	DATA10	VDDE2	DATA12	DATA14	EMIOS2	EMIOS8	EMIOS12	EMIOS21	EMIOS22	VDDEH4	VDDE5	NC_36	VSS	VDD	VRC33	VDD_SYN	AC
AD	ADDR28	ADDR30	VSS	VDD	DATA24	DATA25	DATA27	DATA29	VDD33	GPIO207	DATA9	DATA11	DATA13	DATA15	EMIOS3	EMIOS6	EMIOS10	EMIOS15	EMIOS17	EMIOS23	CNTXA	VDDE5	NC_37	VSS	VDD	VDD33	AD	
AE	ADDR29	VSS	VDD	DATA17	DATA19	DATA21	DATA23	DATA0	DATA2	DATA4	DATA6	OE	BR	BG	EMIOS1	EMIOS5	EMIOS9	EMIOS13	EMIOS16	EMIOS19	EMIOS20	CNRXA	VDDE5	CLKOUT	VSS	VDD	AE	
AF	VSS	VDD	DATA16	DATA18	VDDE2	DATA20	DATA22	GPIO206	DATA1	DATA3	VDDE2	DATA5	DATA7	NC_38	EMIOS0	EMIOS4	EMIOS7	EMIOS11	EMIOS14	EMIOS18	EMIOS20	CNTXB	CNRXB	VDDE5	ENG_CLK	VSS	AF	

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VSS	VSS	VSS	VSS	VDDE7	VDDE7	VDDE7	VDDE7
VSS	VSS	VSS	VSS	VSS	VSS	VSS	VDDE7
VDDE2	VDDE2	VSS	VSS	VSS	VSS	VSS	VDDE7
VDDE2	VDDE2	VSS	VSS	VSS	VSS	VSS	VDDE7
VDDE2	VDDE2	VSS	VSS	VSS	VSS	VSS	VSS
VDDE2	VSS	VDDE2	VDDE2	VDDE2	VDDE2	VSS	VSS
VSS	VDDE2	VDDE2	VDDE2	VDDE2	VDDE2	VSS	VSS

Note: NC\_x No connects (x = 1 to 38)  
 NC\_36 NC\_37 No connect. AC22 & AD23 reserved

Figure 34. MPC5553 416 Package

**Mechanicals**

	1	2	3	4	5	6	7	8	9	10	11	12	13
A	VSS	VSTBY	AN37	AN11	VDDA1	AN16	AN1	AN5	VRH	AN23	AN27	AN28	AN35
B	VDD	VSS	AN36	AN39	AN19	AN20	AN0	AN4	REF BYPC	AN22	AN26	AN31	AN32
C	VDD33	VDD	VSS	AN8	AN17	VSSA1	AN21	AN3	AN7	VRL	AN25	AN30	AN33
D	ETPUA 30	ETPUA 31	VDD	VSS	AN38	AN9	AN10	AN18	AN2	AN6	AN24	AN29	AN34
E	ETPUA 28	ETPUA 29	VDDEH 1	VDD									
F	ETPUA 24	ETPUA 27	ETPUA 26	VDDEH 1									
G	ETPUA 23	ETPUA 22	ETPUA 25	ETPUA 21									
H	ETPUA 20	ETPUA 19	ETPUA 18	ETPUA 17									
J	ETPUA 16	ETPUA 15	ETPUA 14	ETPUA 13									
K	ETPUA 12	ETPUA 11	ETPUA 10	ETPUA 9						VSS	VSS	VSS	VSS
L	ETPUA 8	ETPUA 7	ETPUA 6	ETPUA 5						VSS	VSS	VSS	VSS
M	ETPUA 4	ETPUA 3	ETPUA 2	ETPUA 1						VDDE2	VDDE2	VSS	VSS
N	BDIP	TEA	ETPUA 0	TCRCLK A						VDDE2	VDDE2	VSS	VSS
P	CS3	CS2	CS1	CS0						VDDE2	VDDE2	VSS	VSS
R	WE3	WE2	WE1	WE0						VDDE2	VDDE2	VSS	VSS
T	VDDE2	NC_34	RD_WR	VDDE2						VDDE2	VSS	VDDE2	VDDE2
U	ADDR 16	NC_35	TA	VDD33						VSS	VDDE2	VDDE2	VDDE2
V	ADDR 18	ADDR 17	TS	ADDR 8									
W	ADDR 20	ADDR 19	ADDR 9	ADDR 10									
Y	ADDR 22	ADDR 21	ADDR 11	VDDE2									
AA	ADDR 24	ADDR 23	ADDR 13	ADDR 12									
AB	VDDE2	ADDR 25	ADDR 15	ADDR 14									
AC	ADDR 26	ADDR 27	ADDR 31	VSS	VDD	DATA 26	DATA 28	VDDE2	DATA 30	DATA 31	DATA 8	DATA 10	VDDE2
AD	ADDR 28	ADDR 30	VSS	VDD	DATA 24	DATA 25	DATA 27	DATA 29	VDD33	GPIO 207	DATA 9	DATA 11	DATA 13
AE	ADDR 29	VSS	VDD	DATA 17	DATA 19	DATA 21	DATA 23	DATA 0	DATA 2	DATA 4	DATA 6	OE	BR
AF	VSS	VDD	DATA 16	DATA 18	VDDE2	DATA 20	DATA 22	GPIO 206	DATA 1	DATA 3	VDDE2	DATA 5	DATA 7

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Note: NC\_X No connects (x = 1 to 38)  
NC\_36 NC\_37 No connect. AC22 & AD23 reserved

**Figure 35. MPC5553 416 Package, Left Side**

14	15	16	17	18	19	20	21	22	23	24	25	26	
VSSA0	AN15	ETRIG 1	NC_1	NC_2	NC_3	NC_4	GPIO 205	MDO11	MDO8	VDD	VDD33	VSS	A
VSSA0	AN14	ETRIG 0	NC_5	NC_6	NC_7	NC_8	MDO10	MDO7	MDO4	MDO0	VSS	VDDE7	B
VDDA0	AN13	NC_9	NC_10	NC_11	NC_12	MDO9	MDO6	MDO3	MDO1	VSS	VDDE7	VDD	C
VDDEH 9	AN12	NC_13	NC_14	NC_15	NC_16	MDO5	MDO2	VDDEH 8	VSS	VDDE7	TCK	TDI	D
									VDDE7	TMS	TDO	TEST	E
									MSE00	JCOMP	EVTI	EVTO	F
									MSE01	MCKO	GPIO 204	NC_17	G
									RDY	GPIO 203	NC_18	NC_19	H
									VDDEH 6	NC_20	NC_21	NC_22	J
VDDE7	VDDE7	VDDE7	VDDE7						NC_23	NC_24	NC_25	NC_26	K
VSS	VSS	VSS	VDDE7						NC_27	NC_28	NC_29	NC_30	L
VSS	VSS	VSS	VDDE7						NC_31	NC_32	NC_33	SINB	M
VSS	VSS	VSS	VDDE7						SOUTB	PCSB3	PCSB0	PCSB1	N
VSS	VSS	VSS	VSS						PCSA3	PCSB4	SCKB	PCSB2	P
VSS	VSS	VSS	VSS						PCSB5	SOUTA	SINA	SCKA	R
VDDE2	VDDE2	VSS	VSS						PCSA1	PCSA0	PCSA2	VPP	T
VDDE2	VDDE2	VSS	VSS						PCSA4	TXDA	PCSA5	VFLASH	U
									CNTXC	RXDA	RSTOUT	RST CFG	V
									RXDB	CNRXC	TXDB	RESET	W
									WKP CFG	BOOT CFG1	VRC VSS	VSS SYN	Y
									VDDEH 6	PLL CFG1	BOOT CFG0	EXTAL	AA
									VDD	VRC CTL	PLL CFG0	XTAL	AB
DATA 12	DATA 14	EMIOS 2	EMIOS 8	EMIOS 12	EMIOS 21	VDDEH 4	VDDE5	NC_36	VSS	VDD	VRC33	VDD SYN	AC
DATA 15	EMIOS 3	EMIOS 6	EMIOS 10	EMIOS 15	EMIOS 17	EMIOS 22	CNTXA	VDDE5	NC_37	VSS	VDD	VDD33	AD
BG	EMIOS 1	EMIOS 5	EMIOS 9	EMIOS 13	EMIOS 16	EMIOS 19	EMIOS 23	CNRXA	VDDE5	CLKOUT	VSS	VDD	AE
NC_38	EMIOS 0	EMIOS 4	EMIOS 7	EMIOS 11	EMIOS 14	EMIOS 18	EMIOS 20	CNTXB	CNRXB	VDDE5	ENG CLK	VSS	AF
14	15	16	17	18	19	20	21	22	23	24	25	26	

Figure 36. MPC5553 416 Package, Right Side

### 4.1.2 MPC5553 324 PBGA Pinout

Figure 37 is a pinout for the MPC5553 324 PBGA package.

	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18	19	20	21	22	
A	VSS	VDD	VSTBY	AN37	AN11	VDDA1	VSSA1	AN1	AN5	VRH	VRL	AN27	AN28	AN35	VSSA0	AN12	MDO11	MDO10	MDO8	VDD	VDD33	VSS	A
B	VDD33	VSS	VDD	AN36	AN39	AN19	AN16	AN0	AN4	REF BYPC	AN23	AN26	AN31	AN32	VSSA0	AN13	MDO9	MDO7	MDO4	MDO0	VSS	VDDE7	B
C	ETPUA 30	ETPUA 31	VSS	VDD	AN8	AN17	AN20	AN21	AN3	AN7	AN22	AN25	AN30	AN33	VDDA0	AN14	MDO5	MDO2	MDO1	VSS	VDDE7	VDD	C
D	ETPUA 28	ETPUA 29	ETPUA 26	VSS	VDD	AN38	AN9	AN10	AN18	AN2	AN6	AN24	AN29	AN34	VDDEH 9	AN15	MDO6	MDO3	VSS	VDDE7	TCK	TDI	D
E	ETPUA 24	ETPUA 27	ETPUA 25	ETPUA 21															VDDE7	TMS	TDO	TEST	E
F	ETPUA 23	ETPUA 22	ETPUA 17	ETPUA 18															VDDE7	JCOMP	EVTI	EVTO	F
G	ETPUA 20	ETPUA 19	ETPUA 14	ETPUA 13															RDY	MCKO	MSE00	MSE01	G
H	ETPUA 16	ETPUA 15	ETPUA 10	VDDEH 1															VDDEH 10	GPIO 203	GPIO 204	SINB	H
J	ETPUA 12	ETPUA 11	ETPUA 6	ETPUA 9															SOUTB	PCSB3	PCSB0	PCSB1	J
K	ETPUA 8	ETPUA 7	ETPUA 2	ETPUA 5															PCSA3	PCSB4	SCKB	PCSB2	K
L	ETPUA 4	ETPUA 3	ETPUA 0	ETPUA 1															PCSA5	SOUTA	SINA	SCKA	L
M	BDIP	TCRCLK A	CS1	CS0															PCSA1	PCSA0	PCSA2	VPP	M
N	CS3	CS2	WE1	WE0															PCSA4	TXDA	PCSA5	VFLASH	N
P	ADDR 16	ADDR 17	RD_WR	VDD33															CNTXC	RXDA	RSTOUT	RST CFG	P
R	ADDR 18	ADDR 19	VDDE2	TA															WKP CFG	CNRXC	TXDB	RESET	R
T	ADDR 20	ADDR 21	ADDR 12	TS															RXDB	BOOT CFG1	VRC VSS	VSS SYN	T
U	ADDR 22	ADDR 23	ADDR 13	ADDR 14															VDDEH 6	PLL CFG1	BOOT CFG0	EXTAL	U
V	ADDR 24	ADDR 25	ADDR 15	ADDR 31															VDD	VRC CTL	PLL CFG0	XTAL	V
W	ADDR 26	VDDE2	ADDR 30	VSS	VDD	VDDE2	VDD33	VDDE2	DATA 11	DATA 12	DATA 14	EMIOS 2	EMIOS 8	VDDEH 4	EMIOS 12	EMIOS 21	VDDE5	NC	VSS	VDD	VRC33	VDD SYN	W
Y	ADDR 28	ADDR 27	VSS	VDD	VDDE2	DATA 8	DATA 9	DATA 10	GPIO 207	DATA 13	DATA 15	EMIOS 6	EMIOS 10	EMIOS 15	EMIOS 17	EMIOS 22	CNTXA	VDDE5	NC	VSS	VDD	VDD33	Y
AA	ADDR 29	VSS	VDD	VDDE2	DATA 1	VDDE2	GPIO 206	DATA 5	DATA 7	VDDE2	EMIOS 3	EMIOS 5	EMIOS 9	EMIOS 13	EMIOS 16	EMIOS 19	EMIOS 23	CNRXA	VDDE5	CLKOUT	VSS	VDD	AA
AB	VSS	VDD	VDDE2	DATA 0	DATA 2	DATA 3	DATA 4	DATA 6	OE	EMIOS 0	EMIOS 1	EMIOS 4	EMIOS 7	EMIOS 11	EMIOS 14	EMIOS 18	EMIOS 20	CNTXB	CNRXB	VDDE5	ENG CLK	VSS	AB

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Note: NC No connect. Reserved (W18 & Y19 are shorted to each other)

Figure 37. MPC5553 324 Package

### 4.1.3 MPC5553 208 MAP BGA Pinout

Figure 38 is a pinout for the MPC55MPC5553 208 MAP BGA package.

#### NOTES

$V_{DDEH10}$  and  $V_{DDEH6}$  are connected internally on the 208-ball package and are listed as  $V_{DDEH6}$ .

	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	
A	VSS	AN9	AN11	VDDA1	VSSA1	AN1	AN5	VRH	VRL	AN27	VSSA0	AN12	MDO2	MDO0	VDD33	VSS	A
B	VDD	VSS	AN38	AN21	AN0	AN4	REF BYPC	AN22	AN25	AN28	VDDA0	AN13	MDO3	MDO1	VSS	VDD	B
C	VSTBY	VDD	VSS	AN17	AN34	AN16	AN3	AN7	AN23	AN32	AN33	AN14	AN15	VSS	MSE00	TCK	C
D	VDD33	AN39	VDD	VSS	AN18	AN2	AN6	AN24	AN30	AN31	AN35	VDDEH 9	VSS	TMS	EVTO	TEST	D
E	ETPUA 30	ETPUA 31	AN37	VDD									VDDE7	TDI	EVTI	MSE01	E
F	ETPUA 28	ETPUA 29	ETPUA 26	AN36									VDDEH 6	TDO	MCKO	JCOMP	F
8 June 2005p																	
G	ETPUA 24	ETPUA 27	ETPUA 25	ETPUA 21			VSS	VSS	VSS	VSS			SOUTB	PCSB3	SINB	PCSB0	G
H	ETPUA 23	ETPUA 22	ETPUA 17	ETPUA 18			VSS	VSS	VSS	VSS			PCSA3	PCSB4	PCSB2	PCSB1	H
J	ETPUA 20	ETPUA 19	ETPUA 14	ETPUA 13			VSS	VSS	VSS	VSS			PCSB5	TXDA	PCSA2	SCKB	J
K	ETPUA 16	ETPUA 15	ETPUA 7	VDDEH 1			VSS	VSS	VSS	VSS			CNTXC	RXDA	RSTOUT	VPP	K
L	ETPUA 12	ETPUA 11	ETPUA 6	TCRCLK A									TXDB	CNRXC	WKP CFG	RESET	L
M	ETPUA 10	ETPUA 9	ETPUA 1	ETPUA 5									RXDB	PLL CFG0	BOOT CFG1	VSS SYN	M
N	ETPUA 8	ETPUA 4	ETPUA 0	VSS	VDD	VDD33	EMIOS 2	EMIOS 10	VDDEH 4	EMIOS 12	EMIOS 21	VDD33	VSS	VRC CTL	PLL CFG1	EXTAL	N
P	ETPUA 3	ETPUA 2	VSS	VDD	GPIO 207	VDDE2	EMIOS 6	EMIOS 8	EMIOS 16	EMIOS 17	EMIOS 22	CNTXA	VDD	VSS	VRC33	XTAL	P
R	CS0	VSS	VDD	GPIO 206	EMIOS 4	EMIOS 3	EMIOS 9	EMIOS 11	EMIOS 14	EMIOS 19	EMIOS 23	CNRXA	CNRXB	VDD	VSS	VDD SYN	R
T	VSS	VDD	OE	EMIOS 0	EMIOS 1	EMIOS 5	EMIOS 7	EMIOS 13	EMIOS 15	EMIOS 18	EMIOS 20	CNTXB	VDDE5	ENG CLK	VDD	VSS	T

Figure 38. MPC5553 208 Package





NOTES:

1. ALL DIMENSIONS IN MILLIMETERS.
2. DIMENSIONING AND TOLERANCING PER ASME Y14.5M-1994.
3. MAXIMUM SOLDER BALL DIAMETER MEASURED PARALLEL TO DATUM A.
4. DATUM A, THE SEATING PLANE, IS DETERMINED BY THE SPHERICAL CROWNS OF THE SOLDER BALLS.

© FREESCALE SEMICONDUCTOR, INC. ALL RIGHTS RESERVED.		<b>MECHANICAL OUTLINE</b>		PRINT VERSION NOT TO SCALE	
TITLE: 416 I/O, PBGA 27 X 27 PKG, 1 MM PITCH (OMPAC)	DOCUMENT NO: 98ARE10523D		REV: A		
	CASE NUMBER: 1494-01		13 JUL 2005		
	STANDARD: JEDEC MS-034 AAL-1				

**Figure 39. MPC5553 416 TEPBGA Package (continued)**

### 4.2.2 MPC5553 324-Pin Package

The package drawings of the MPC5553 324-pin TEPBGA package are shown in Figure 40.

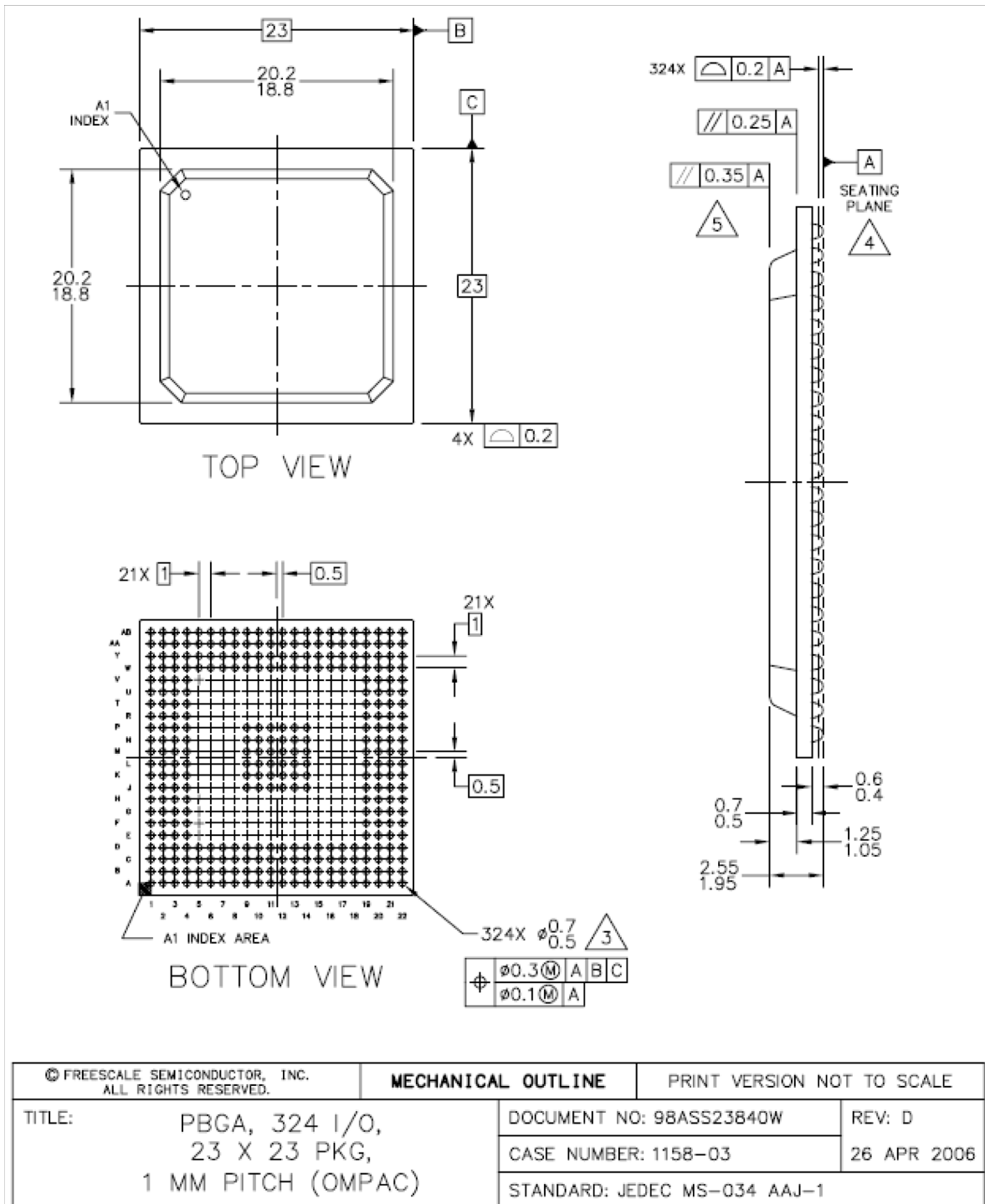


Figure 40. MPC5553 324 TEPBGA Package

## NOTES:

1. ALL DIMENSIONS IN MILLIMETERS.
2. DIMENSIONING AND TOLERANCING PER ASME Y14.5M-1994.
3. MAXIMUM SOLDER BALL DIAMETER MEASURED PARALLEL TO DATUM A.
4. DATUM A, THE SEATING PLANE, IS DETERMINED BY THE SPHERICAL CROWNS OF THE SOLDER BALLS.
5. PARALLELISM MEASUREMENT SHALL EXCLUDE ANY EFFECT OF MARK ON TOP SURFACE OF PACKAGE.

© FREESCALE SEMICONDUCTOR, INC. ALL RIGHTS RESERVED.	<b>MECHANICAL OUTLINE</b>	PRINT VERSION NOT TO SCALE	
TITLE: PBGA, 324 I/O, 23 X 23 PKG, 1 MM PITCH (OMPAC)	DOCUMENT NO: 98ASS23840W	REV: D	
	CASE NUMBER: 1158-03	26 APR 2006	
	STANDARD: JEDEC MS-034 AAJ-1		

Figure 40. MPC5553 324 TEPBGA Package (continued)

### 4.2.3 MPC5553 208-Pin Package

The package drawings of the MPC5553 208-pin MAP BGA package are shown in Figure 41.

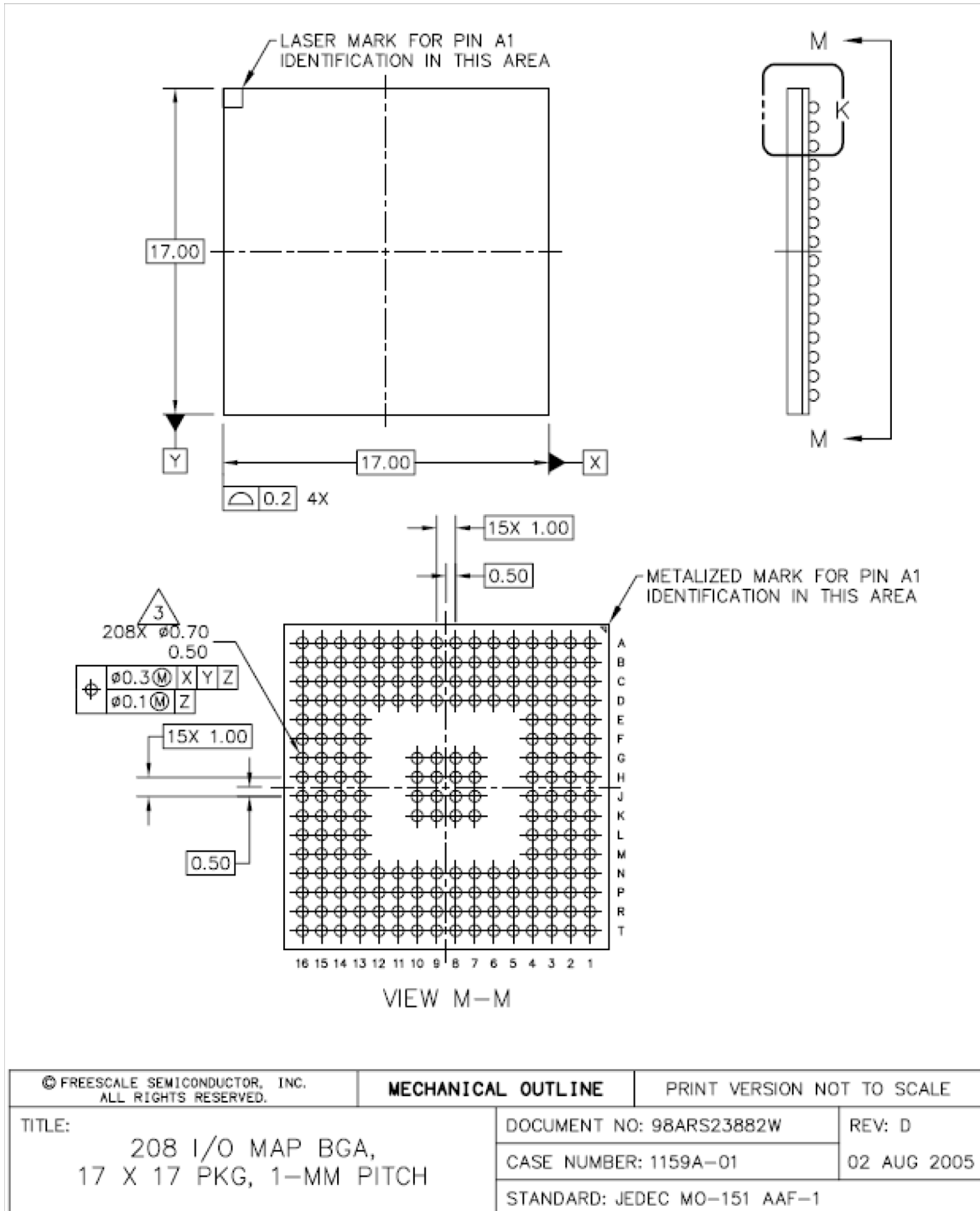
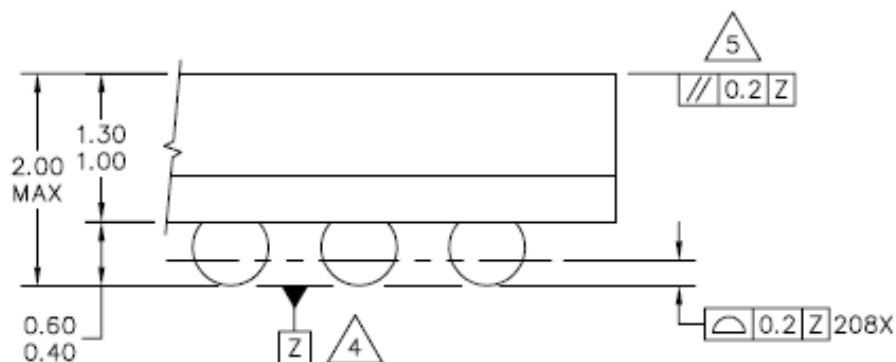


Figure 41. MPC5553 208 MAP BGA Package



DETAIL K  
(ROTATED 90° CLOCKWISE)

NOTES:

1. ALL DIMENSIONS ARE IN MILLIMETERS.
2. INTERPRET DIMENSIONS AND TOLERANCES PER ASME Y14.5M-1994.
3. DIMENSION b IS MEASURED AT THE MAXIMUM SOLDER BALL DIAMETER, PARALLEL TO DATUM PLANE Z.
4. DATUM Z (SEATING PLANE) IS DEFINED BY THE SPHERICAL CROWNS OF THE SOLDER BALLS.
5. PARALLELISM MEASUREMENT SHALL EXCLUDE ANY EFFECT OF MARK ON TOP SURFACE OF PACKAGE.

© FREESCALE SEMICONDUCTOR, INC. ALL RIGHTS RESERVED.	<b>MECHANICAL OUTLINE</b>	PRINT VERSION NOT TO SCALE	
TITLE: 208 I/O MAP BGA, 17 X 17 PKG, 1-MM PITCH	DOCUMENT NO: 98ARS23882W	REV: D	
	CASE NUMBER: 1159A-01	02 AUG 2005	
	STANDARD: JEDEC MO-151 AAF-1		

Figure 41. MPC5553 208 MAP BGA Package (continued)

## 5 MPC5553 Revision History

Table 32 provides a revision history of the MPC5553 Data Sheet.

**Table 32. MPC5553 Revision History**

Revision	Location(s)	Substantive Change(s)
Rev. 0		This is the first released version of this document.
Rev. 1	Table 1	Footnote added to Freescale Part Number column.
	Table 2	Footnotes 6, 8, and 9 changed from 1mA to 2mA.
	Figure 39, Figure 40, Figure 41	Second page of package drawings added.
	Figure 37	Removed note about pin R1 in the figure and added a Note above it instead.
Rev. 1.1	Throughout	Editorial changes: subscripting, simplifying language.

Table 33 is the new format for the Revision History and changes continue from Table 32.

**Table 33. MPC5553 Revision History (continued)**

Revision	Author	Date	Substantive Change(s)
Rev. 1.1	NH	02/02/07	<p>Changes per RD initial review:</p> <ul style="list-style-type: none"> <li>• Changed the values in <a href="#">Table 14</a> for the H7Fa Flash pre-program and erase times. Typical and Initial Max values changed.</li> <li>• Typical Values — <ul style="list-style-type: none"> <li>16 Kbytes: from 265 to 325</li> <li>48 Kbytes: from 340 to 435</li> <li>64 Kbytes: from 400 to 525</li> <li>128 Kbytes: from 500 to 675</li> </ul> </li> <li>• Initial Max Values — <ul style="list-style-type: none"> <li>16 Kbytes: from 400 to 525</li> <li>48 Kbytes: from 400 to 525</li> <li>64 Kbytes: from 500 to 675</li> <li>128 Kbytes: from 1250 to 1800</li> </ul> </li> </ul>
Rev. 1.1	NH	02/06/07	<p>Changes per RD second review:</p> <ul style="list-style-type: none"> <li>• Added <a href="#">Figure 3</a> to show interpolated <math>IDD_{STBY}</math> values listed in <a href="#">Table 9</a>.</li> <li>• <a href="#">Table 9</a> DC Electrical Specifications: Changed wording of footnote 3. Spec 28: Corrected conditional text error showing wrong frequency. Spec 29: Deleted frequency information.</li> <li>• <a href="#">Table 6</a> FMPLL Electrical Characteristics: Grouped (2 x CI) in Specs 12 and 13.</li> <li>• <a href="#">Table 7</a> Power Sequence Pin Status for Fast Pads, updated paragraph.</li> <li>• <a href="#">Table 8</a> Power Sequence Pin Status, updated preceding paragraph.</li> <li>• <a href="#">Section 3.7.1</a>, "Input Value of Pins During POR Dependent on VDD33 Updated paragraph to remove redundancy,</li> <li>• <a href="#">Table 16</a> Flash BIU Settling: Changed wording of footnote from "Can be changed after Analysis and Characterization" to "These values may change after characterization."</li> <li>• <a href="#">Table 17</a> and <a href="#">Table 18</a>: Deleted the words 'not' from footnote 2. Changed from 'This parameter is supplied for reference and is not guaranteed by design and not tested' to 'This parameter is supplied for reference and is guaranteed by design and tested.'</li> <li>• <a href="#">Table 22</a> Bus Operation Timing: Specs 5 and 6: corrected format to show the bus timing values for various frequencies with EBTS bit = 0 and EBTS bit = 1. Specs 6 and 7: Added the calibrations signals: CAL_ADDR, CAL_<math>\overline{WE/BE}</math>, CAL_<math>\overline{CS}</math>, CAL_DATA.</li> <li>• <a href="#">Table 26</a> DSPI Timing: Added to beginning of footnote 1 'All DSPI timing specifications use the fastest slew rate (SRC=0b11) on pad type M or MH. DSPI signals using pad types of S or SH have an additional delay based on the slew rate.'</li> <li>• <a href="#">Table 27</a> EQADC SS Timing Characteristics: combined footnotes 1 and 2. Moved footnotes 1 and 2 to Spec 2 and deleted Spec 1.</li> </ul>

Table 33. MPC5553 Revision History (*continued*)

Revision	Author	Date	Substantive Change(s)
Rev 2.0	NH	02/07/07	<p>Changes per RD sign-off review:</p> <ul style="list-style-type: none"> <li>Changed paragraph preceding <a href="#">Table 7</a> Power Sequence Pin Status for the Fast Pad: <ul style="list-style-type: none"> <li>From: Although there are no power up/down sequencing requirements to prevent issues like latch-up, excessive current spikes, etc., the state of the I/O pins during power up/down varies depending on power. Prior to exiting POR, the pads are in a high impedance state (Hi-Z).</li> <li>To: There are no power up/down sequencing requirements to prevent issues such as latch-up, excessive current spikes, and so on. Therefore, the state of the I/O pins during power up/down varies depending on which supplies are powered.</li> </ul> </li> <li><a href="#">Section 3.7.1, "Input Value of Pins During POR Dependent on VDD33,"</a> changed <ul style="list-style-type: none"> <li>From: <ul style="list-style-type: none"> <li>To avoid accidentally selecting the bypass clock because PLLCFG[0:1] and RSTCFG are not treated as ones (1s) when POR negates, VDD33 must not lag VDDSYN and the RESET pin power (VDDEH6) when powering the device by more than the VDD33 lag specification in <a href="#">Table 6</a>. VDD33 individually can lag either VDDSYN or the RESET power pin (VDDEH6) by more than the VDD33 lag specification. VDD33 can lag one of the VDDSYN or VDDEH6 supplies, but cannot lag both by more than the VDD33 lag specification. This VDD33 lag specification only applies during power up. VDD33 has no lead or lag requirements when powering down.</li> <li>To: <ul style="list-style-type: none"> <li>When powering the device, VDD33 must not lag VDDSYN and the RESET power pin (VDDEH6) by more than the VDD33 lag specification listed in <a href="#">Table 6</a>. This avoids accidentally selecting the bypass clock mode because the internal versions of PLLCFG[0:1] and RSTCFG are not powered and therefore cannot read the default state when POR negates. VDD33 can lag VDDSYN or the RESET power pin (VDDEH6), but cannot lag both by more than the VDD33 lag specification. This VDD33 lag specification only applies during power up. VDD33 has no lead or lag requirements when powering down.</li> </ul> </li> </ul> </li> </ul> </li> <li><a href="#">Table 22</a> Bus Operation Timing: <ul style="list-style-type: none"> <li>Added the correct pins to the calibration signals: CAL_ADDR[10:11, 27:30], CAL_WE/BE[0:1], CAL_CS[0, 2:3], and CAL_DATA[0:15]. Added calibration signals to Specs 5 and 8.</li> </ul> </li> <li>Corrected the following EBI signals: <ul style="list-style-type: none"> <li>Specs 7 and 8: Added the following signals to Specs 7 and 8 the EBI section: OE, RD_WR, and BDIP. Broke out Spec 6 CLKOUT Posedge to output signal valid into Spec 6 for the EBI signals, and Spec 6a for the calibration signals, Broke out Spec 7 Input Signal Valid to CLKOUT Posedge into Spec 7 for the EBI signals, and Spec 7a for the calibration signals.</li> </ul> </li> <li><a href="#">Section 3.7.3, "Power-Down Sequence (VRC33 Grounded)"</a> Deleted the underscore in ORed_POR to become ORed POR.</li> </ul>
Rev 2.0	NH	2/09/07	<p><a href="#">Table 22</a> Bus Operation Timing:</p> <p>Removed references to CAL_OE, CAL_RD_WR, and CAL_TS because they really use the EBI signals OE, RD_WR, and TS on the MPC5553.</p>



Table 33. MPC5553 Revision History (*continued*)

Revision	Author	Date	Substantive Change(s)
Rev 2.0	NH	2/27/07	Per RD comments: <a href="#">Table 2</a> Absolute Maximum Ratings: changed footnote 6 from: Keep the negative DC current greater than 0.6 V on eTPUB[15] and SINB during the internal power-on reset (POR) state. To: Keep the negative DC current greater than –0.6 V on eTPUB[15] and SINB during the internal power-on reset (POR) state. <a href="#">Figure 38</a> MPC5553 208 Map BGA Pinout: Deleted two lines referring to the $\overline{CS}[0]$ signal ball assignment for the 208.
Rev 2.0	NH	3/1/07	Corrected the signal names in <a href="#">Section 3.14</a> , “Fast Ethernet AC Timing Specifications” to include the FEC_ prefix for the signal name. Waiting on response from Jim Eifert, Randy Dees, Jeffery Hopkins, and Bill Terry about the following Bugs filed against the Data Sheets: 1474, 1480, 1482, 1483, 1811, 1815, 1884, 2254, 2419, 2717, 2873 before preparing for final sign-off again.

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+33 1 69 35 48 48 (French)  
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Freescale Semiconductor Japan Ltd.  
Headquarters  
ARCO Tower 15F  
1-8-1, Shimo-Meguro, Meguro-ku,  
Tokyo 153-0064  
Japan  
0120 191014 or +81 3 5437 9125  
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