

# 80 mA, Tiny CMOS LDO With Shutdown

#### **Features**

- Space-Saving 5-Pin SC-70 and SOT-23 Packages
- Extremely Low Operating Current for Longer Battery Life: 53 µA (typ.)
- Very Low Dropout Voltage
- Rated 80 mA Output Current
- Requires only 1 µF Ceramic Output Capacitance
- High Output Voltage Accuracy: ±0.5% (typ.)
- 10 μsec (typ.) Wake-Up Time from SHDN
- Power-Saving Shutdown Mode: 0.05 μA(typ.)
- Overcurrent and Overtemperature Protection
- Pin Compatible Upgrade for Bipolar Regulators

#### **Applications**

- Cellular/GSM/PHS Phones
- Battery-operated Systems
- Portable Computers
- Medical Instruments
- Electronic Games
- Pagers

#### **General Description**

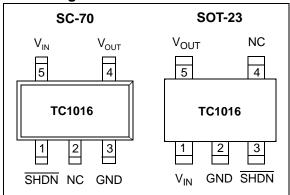
The TC1016 is a high-accuracy (typically ±0.5%), CMOS upgrade for bipolar low dropout regulators (LDOs). The TC1016 is offered in both the SC-70 and SOT-23 packages. The SC-70 package represents a 50% footprint reduction versus the popular SOT-23 package.

Developed specifically for battery-powered systems, the device's CMOS construction consumes only 53  $\mu$ A typical supply current over the entire 80 mA operating load range. This can be as much as 60 times less than the quiescent operating current consumed by bipolar LDOs.

With small-space requirements and cost in mind, the TC1016 was developed to be stable over the entire input voltage and output current operating range using low value (1  $\mu F$  ceramic), low Equivalent Series Resistance (ESR) output capacitors. Additional integrated features (such as shutdown, overcurrent and overtemperature protection) further reduce board space and cost of the entire voltage-regulating application.

Key performance parameters for the TC1016 are low drop out voltage (150 mV (typ.) at 80 mA output current), low supply current while shutdown (0.05  $\mu A$  typical) and fast stable response to sudden input voltage and load changes.

#### **Pin Configurations**



#### 1.0 **ELECTRICAL CHARACTERISTICS**

#### ABSOLUTE MAXIMUM RATINGS\*

Input Voltage ......6.5V Power Dissipation.....Internally Limited (Note 7) Operating Temperature .....-40°C < T<sub>J</sub> < 125°C Storage Temperature......-65°C to +150°C Maximum Voltage On Any Pin......V<sub>IN</sub> + 0.3V to -0.3V

\*Notice: Static-sensitive device. Unused devices must be stored in conductive material. Protect devices from static discharge and static fields. Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other conditions above those indicated in the operational sections of the specifications is not implied. Exposure to Absolute Maximum Rating Conditions for extended periods may affect device reliability

#### **ELECTRICAL CHARACTERISTICS**

 $V_{IN} = V_R + 1V$ ,  $I_L = 100 \,\mu\text{A}$ ,  $C_L = 1.0 \,\mu\text{F}$ ,  $\overline{\text{SHDN}} > V_{IH}$ ,  $T_A = 25 \,^{\circ}\text{C}$ , unless otherwise noted. **Boldface** type specifications apply for junction temperatures of  $-40 \,^{\circ}\text{C}$  to  $+125 \,^{\circ}\text{C}$ .

junction temperatures of – 40 C to +125 C.								
Parameter	Sym	Min	Тур	Max	Units	Test Conditions		
Input Operating Voltage	V <sub>IN</sub>	2.7		6.0	V	Note 1		
Maximum Output Current	I <sub>OUTMAX</sub>	80			mA			
Output Voltage	V <sub>OUT</sub>	V <sub>R</sub> - 2.5%	$V_R \pm 0.5\%$	V <sub>R</sub> + 2.5%	V	Note 2		
V <sub>OUT</sub> Temperature Coefficient	TCV <sub>OUT</sub>	1	40	1	ppm/°C	Note 3		
Line Regulation	$(\Delta V_{OUT}/\Delta V_{IN})/V_{R}$	ı	0.01	0.2	%/V	$(V_R + 1V) < V_{IN} < 6V$		
Load Regulation (Note 4)	$\Delta V_{OUT}/V_{R}$	1	0.23	1	%	$I_L = 0.1 \text{ mA to } I_{OUTMAX}$		
Dropout Voltage (Note 5)	V <sub>IN</sub> – V <sub>OUT</sub>		2 100 150	 200 300	mV	I <sub>L</sub> = 100 μA I <sub>L</sub> = 50 mA I <sub>L</sub> = 80 mA		
Supply Current	I <sub>IN</sub>	1	53	90	μA	$\overline{SHDN} = V_{IH}, I_L = 0$		
Shutdown Supply Current	I <sub>INSD</sub>	-	0.05	0.5	μΑ	SHDN = 0V		
Power Supply Rejection Ratio	PSRR		58	_	dB	f =1 kHz, I <sub>L</sub> = 50 mA		
Wake-Up Time (from Shutdown mode)	t <sub>WK</sub>		10		μs	$V_{IN} = 5V$ , $I_L = 60$ mA, $C_{IN} = 1$ $\mu$ F, $C_{OUT} = 1$ $\mu$ F, $f = 100$ Hz		
Settling Time (from Shutdown Mode)	t <sub>S</sub>	1	32	1	μs	$V_{IN}$ = 5V, $I_{L}$ = 60 mA, $C_{IN}$ = 1 $\mu$ F, $C_{OUT}$ = 1 $\mu$ F, f = 100 Hz		
Output Short Circuit Current	l <sub>outsc</sub>	1	120	_	mA	V <sub>OUT</sub> = 0V		
Thermal Regulation	V <sub>OUT</sub> /P <sub>D</sub>		0.04		V/W	Notes 6, 7		
Thermal Shutdown Die Temperature	T <sub>SD</sub>		160	I	ç			
Thermal Shutdown Hysteresis	$\DeltaT_{SD}$	ı	10	ı	°C			
Output Noise	eN		800	_	nV/√Hz	f = 10 kHz		
SHDN Input High Threshold	V <sub>IH</sub>	60			%V <sub>IN</sub>	$V_{IN} = 2.7V \text{ to } 6.0V$		
SHDN Input Low Threshold	V <sub>IL</sub>	_		15	%V <sub>IN</sub>	$V_{IN} = 2.7V \text{ to } 6.0V$		

- The minimum  $V_{IN}$  has to meet two conditions:  $V_{IN} \ge 2.7V$  and  $V_{IN} \ge (V_R + 2.5\%) + V_{DROPOUT} + V_R$  is the regulator voltage setting. For example:  $V_R = 1.8V$ , 2.7V, 2.8V, 3.0V. Note 1:

3: 
$$TCV_{OUT} = \frac{(V_{OUTMAX} - V_{OUTMIN}) \times 10^6}{V_{OUT} \times \Delta T}$$

- 4: Regulation is measured at a constant junction temperature using low duty cycle pulse testing. Load regulation is tested over a load range from 0.1 mA to the maximum specified output current. Changes in output voltage due to heating effects are covered by the Thermal Regulation specification.
- 5: Dropout voltage is defined as the input-to-output differential at which the output voltage drops 2% below its nominal value at a 1V differential.
- 6: Thermal regulation is defined as the change in output voltage at a time T after a change in power dissipation is applied, excluding load or line regulation effects. Specifications are for a current pulse equal to Ilmax at  $V_{IN} = 6V$  for t = 10 msec.
- The maximum allowable power dissipation is a function of ambient temperature, the maximum allowable juction temperature and the thermal resistance from junction-to-air (i.e.  $T_A$ ,  $T_J$ ,  $\theta_{JA}$ ). Exceeding the maximum allowable power dissipation causes the device to initiate thermal shutdown. Please see **Section 5.0** "**Thermal Considerations**" of this data sheet for more details.

#### 2.0 TYPICAL PERFORMANCE CURVES

**Note:** The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only. The performance characteristics listed herein are not tested or guaranteed. In some graphs or tables, the data presented may be outside the specified operating range (e.g., outside specified power supply range) and therefore outside the warranted range.

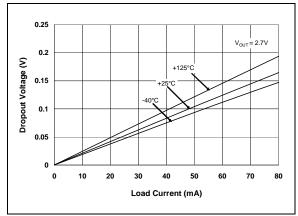
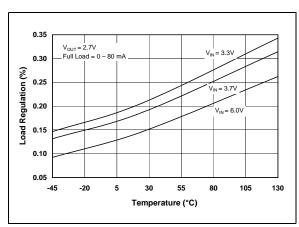
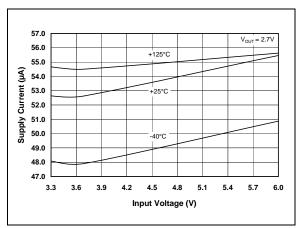


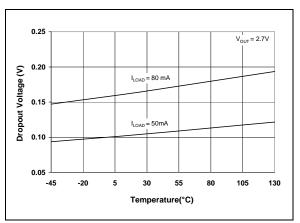
FIGURE 2-1: Dropout Voltage vs. Output Current.



**FIGURE 2-2:** Load Regulation vs. Temperature.



**FIGURE 2-3:** Supply Current vs. Input Voltage.



**FIGURE 2-4:** Dropout Voltage vs. Temperature.

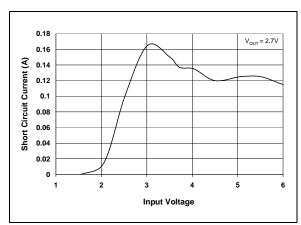


FIGURE 2-5: Short Circuit Current vs. Input Voltage.

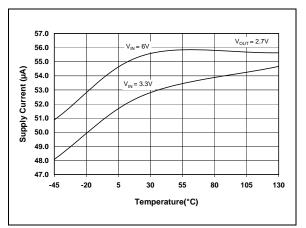
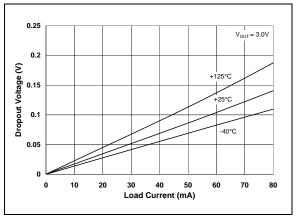


FIGURE 2-6: Supply Current vs. Temperature.



**FIGURE 2-7:** Dropout Voltage vs. Output Current.

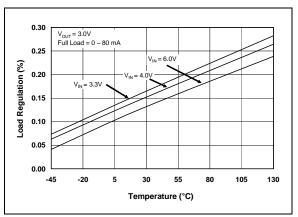
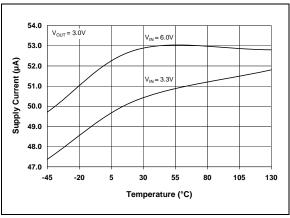
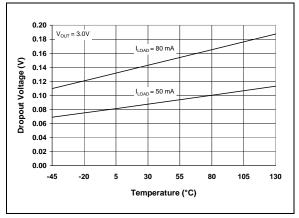


FIGURE 2-8: Load Regulation vs. Temperature.



**FIGURE 2-9:** Supply Current vs. Temperature.



**FIGURE 2-10:** Dropout Voltage vs. Temperature.

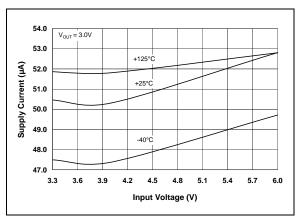
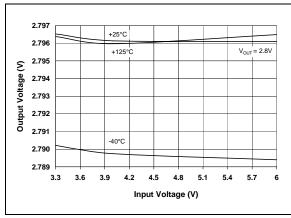
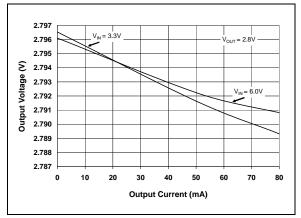


FIGURE 2-11: Supply Current vs. Input Voltage



**FIGURE 2-12:** Output Voltage vs. Supply Voltage.



**FIGURE 2-13:** Output Voltage vs. Output Current.

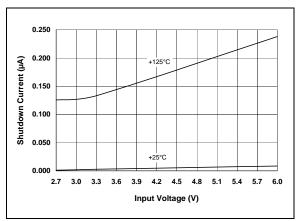
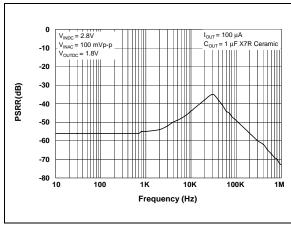
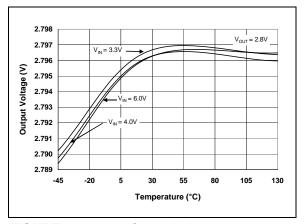


FIGURE 2-14: Shutdown Current vs. Input Voltage.



**FIGURE 2-15:** Power Supply Rejection Ratio vs. Frequency.



**FIGURE 2-16:** Output Voltage vs. Temperature.

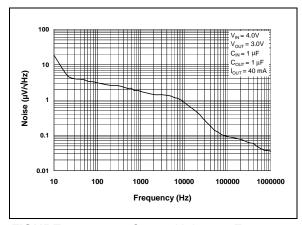
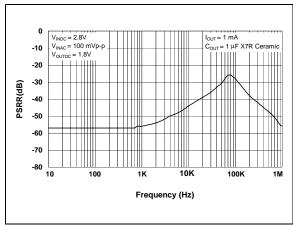
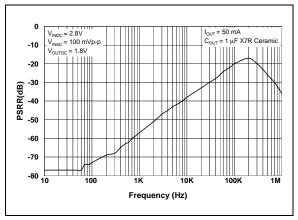


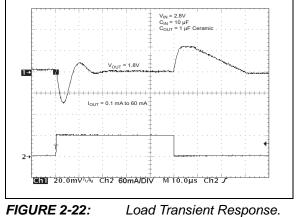
FIGURE 2-17: Output Noise vs. Frequency.

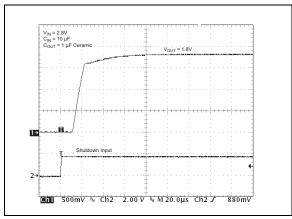


**FIGURE 2-18:** Power Supply Rejection Ratio vs. Frequency.

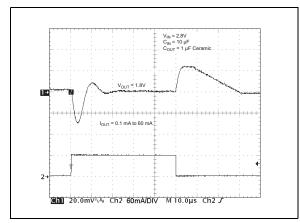


**FIGURE 2-19:** Power Supply Rejection Ratio vs. Frequency.

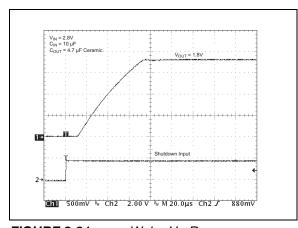




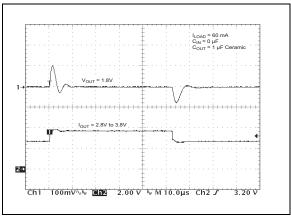
**FIGURE 2-20:** Wake-Up Response.



**FIGURE 2-23:** Load Transient Response.



**FIGURE 2-21:** Wake-Up Response.



**FIGURE 2-24:** Line Transient Response.

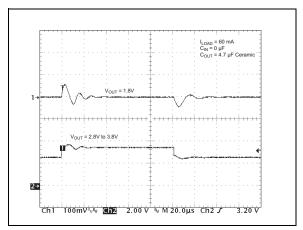


FIGURE 2-25: Line Transient Response.

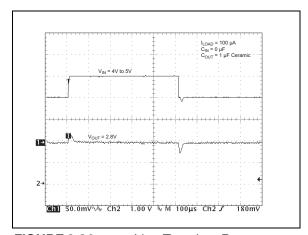


FIGURE 2-26: Line Transient Response.

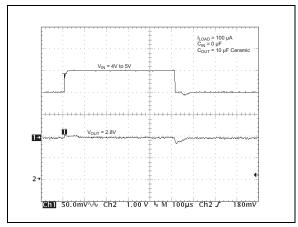


FIGURE 2-27: Line Transient Response.

#### 3.0 PIN DESCRIPTIONS

The descriptions of the pins are listed in Table 3-1.

TABLE 3-1: PIN FUNCTION TABLE

5-Pin SC-70	Pin No. 5-Pin SOT-23	Name	Function
1	3	SHDN	Shutdown control input
2	4	NC	No connect
3	2	GND	Ground terminal
4	5	V <sub>OUT</sub>	Regulated voltage output
5	1	V <sub>IN</sub>	Unregulated supply input

# 3.1 Shutdown Control Input (SHDN)

The regulator is fully enabled when a logic-high is applied to  $\overline{SHDN}$ . The regulator enters shutdown when a logic-low is applied to this input. During shutdown, the output voltage falls to zero and the supply current is reduced to  $0.05~\mu A$  (typ.)

#### 3.2 Ground Terminal (GND)

For best performance, it is recommended that the ground pin be tied to a ground plane.

## 3.3 Regulated Voltage Output (V<sub>OUT</sub>)

Bypass the regulated voltage output to GND with a minimum capacitance of 1  $\mu F$ . A ceramic bypass capacitor is recommended for best performance.

## 3.4 Unregulated Supply Input (V<sub>IN</sub>)

The minimum  $V_{IN}$  has to meet two conditions in order to ensure that the output maintains regulation:  $V_{IN} \geq 2.7 V$  and  $V_{IN} \geq [(V_R + 2.5\%) + V_{DROPOUT}]$ . The maximum  $V_{IN}$  should be less than or equal to 6V. Power dissipation may limit  $V_{IN}$  to a lower potential in order to maintain a junction temperature below 125°C. Refer to **Section 5.0** "Thermal Considerations", for determining junction temperature.

It is recommended that  $V_{\mbox{\scriptsize IN}}$  be bypassed to GND with a ceramic capacitor.

#### 4.0 DETAILED DESCRIPTION

The TC1016 is a precision, fixed-output, linear voltage regulator. The internal linear pass element is a P-channel MOSFET. As with all P-channel CMOS LDOs, there is a body drain diode, with the cathode connected to  $V_{IN}$  and the anode connected to  $V_{OLT}$  (Figure 4-1).

As shown in Figure 4-1, the output voltage of the LDO is sensed and divided down internally to reduce external component count. The internal error amplifier has a fixed, band gap reference on the inverting input, with the sensed output voltage on the non-inverting input. The error amplifier output will pull the gate voltage down until the inputs of the error amplifier are equal in order to regulate the output voltage.

By sensing the current in the P-channel MOSFET, the maximum current delivered to the load is limited to a typical value of 120 mA, preventing excessive current from damaging the Printed Circuit Board (PCB) in the event of a shorted or faulted load.

An internal thermal sensing device is used to monitor the junction temperature of the LDO. When the sensed temperature is over the set threshold of 160°C (typ.), the P-channel MOSFET is turned off. When the MOSFET is off, the power dissipation internal to the device is almost zero. The device cools until the junction temperature is approximately 150°C and the

MOSFET is turned on. If the internal power dissipation is still high enough for the junction to rise to 160°C, it will again shut off and cool. The maximum operating junction temperature of the device is 125°C. Steady-state operation at or near the 160°C overtemperature point can lead to permanent damage of the device.

The output voltage ( $V_{OUT}$ ) remains stable over the entire input operating voltage range (2.7V to 6.0V), as well as the entire load range (0 mA to 80 mA). The output voltage is sensed through an internal resistor divider and compared with a precision internal voltage reference. Several fixed-output voltages are available by changing the value of the internal resistor divider.

Figure 4-2 shows a typical application circuit. The regulator is enabled anytime the shutdown input pin is at or above  $V_{\text{IH}},$  and shutdown (disabled) anytime the shutdown input pin is below  $V_{\text{IL}}.$  For applications where the  $\overline{\text{SHDN}}$  feature is not used, tie the  $\overline{\text{SHDN}}$  pin directly to the input supply voltage source. While in shutdown, the supply current decreases to 0.05  $\mu\text{A}$  (typ.) and the P-channel MOSFET is turned off.

As shown in Figure 4-2, batteries have internal source impedance. An input capacitor in used to lower the input impedance of the LDO. In some applications, high input impedance can cause the LDO to become unstable. Adding more input capacitance can compensate for this.

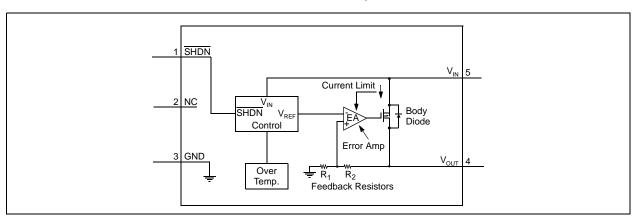


FIGURE 4-1: TC1016 Block Diagram.

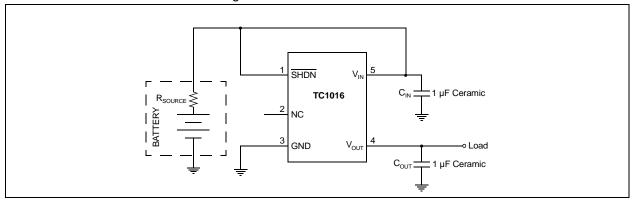


FIGURE 4-2: Typical Application Circuit.

#### 4.1 Input Capacitor

Low input source impedance is necessary for the LDO to operate properly. When operating from batteries, or in applications with long lead length (> 10") between the input source and the LDO, some input capacitance is required. A minimum of 0.1  $\mu F$  is recommended for most applications and the capacitor should be placed as close to the input of the LDO as is practical. Larger input capacitors will help reduce the input impedance and further reduce any high-frequency noise on the input and output of the LDO.

#### 4.2 Output Capacitor

A minimum output capacitance of 1  $\mu F$  for the TC1016 is required for stability. The ESR requirements on the output capacitor are between 0 and 2 ohms. The output capacitor should be located as close to the LDO output as is practical. Ceramic materials X7R and X5R have low temperature coefficients and are well within the acceptable ESR range required. A typical 1  $\mu F$  X5R 0805 capacitor has an ESR of 50 milli-ohms. Larger output capacitors can be used with the TC1016 to improve dynamic behavior and input ripple rejection performance.

Ceramic, aluminum electrolytic or tantalum capacitor types can be used. Since many aluminum electrolytic capacitors freeze at approximately -30°C, ceramic or solid tantalums are recommended for applications operating below -25°C. When operating from sources other than batteries, supply noise rejection and transient response can be improved by increasing the value of the input and output capacitors, and by employing passive filtering techniques.

#### 4.3 Turn-On Response

The turn on response is defined as two separate response categories, Wake-up Time ( $t_{WK}$ ) and Settling Time ( $t_{S}$ ).

The TC1016 has a fast  $t_{WK}$  (10 µsec, typ.) when released from shutdown. Figure 4-3 provides the TC1016's  $t_{WK}$ . The  $t_{WK}$  is defined as the time it takes for the output to rise to 2% of the  $V_{OUT}$  value after being released from shutdown.

The total turn-on response is defined as the  $t_{\rm S}$  (see Figure 4-3). The  $t_{\rm S}$  (inclusive with  $t_{\rm WK}$ ) is defined as the condition when the output is within 98% of its fully enabled value (42 µsec, typ.) when released from shutdown. The settling time of the output voltage is dependent on load conditions and output capacitance on  $V_{\rm OUT}$  (RC response).

Table 4-1 demonstrates the typical turn-on response timing for different input voltage power-up frequencies:  $V_{OUT} = 2.8V$ ,  $V_{IN} = 5.0V$ ,  $I_{OUT} = 60$  mA and  $C_{OUT} = 1$   $\mu F$ .

TABLE 4-1: TYPICAL TURN-ON RESPONSE TIMING

Frequency	Typical (t <sub>WK</sub> )	Typical (t <sub>S</sub> )		
1000 Hz	5.3 µsec	14 µsec		
500 Hz	5.9 µsec	16 µsec		
100 Hz	9.8 µsec	32 µsec		
50 Hz	14.5 µsec	52 µsec		
10 Hz	17.2 µsec	77 µsec		

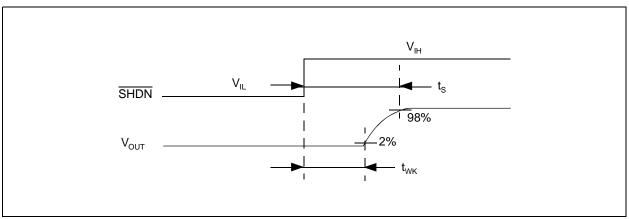


FIGURE 4-3: Wake-Up Time from Shutdown.

#### 5.0 THERMAL CONSIDERATIONS

#### 5.1 Thermal Shutdown

Integrated thermal-protection circuitry shuts the regulator off when die temperature exceeds approximately 160°C. The regulator remains off until the die temperature drops to approximately 150°C.

#### 5.2 Power Dissipation

The TC1016 is available in the SC-70 package. The thermal resistance for the SC-70 package is approximately 450°C/W when the copper area used in the PCB layout is similar to the JEDEC J51-7 high thermal conductivity or Semi G42-88 standards. For applications with larger or thicker copper areas, the thermal resistance can be lowered. See AN792 "A Method to Determine How Much Power a SOT23 Can Dissipate in an Application" (DS00792), for a method to determine the thermal resistance for a particular application.

The TC1016 power dissipation capability is dependant upon several variables: input voltage, output voltage, load current, ambient temperature and maximum junction temperature. The absolute maximum steady-state junction temperature is rated at 125°C. The power dissipation within the device is equal to:

#### **EQUATION 5-1:**

$$P_D = (V_{IN} - V_{OUT}) \times I_{LOAD} + V_{IN} \times I_{GND}$$

The  $V_{IN}$  x  $I_{GND}$  term is typically very small when compared to the  $(V_{IN}-V_{OUT})$  x  $I_{LOAD}$  term simplifying the power dissipation within the LDO to be:

#### **EQUATION 5-2:**

$$P_D = (V_{IN} - V_{OUT}) \times I_{LOAD}$$

To determine the maximum power dissipation capability, the following equation is used:

#### **EQUATION 5-3:**

$$P_{DMAX} = \frac{(T_{\text{J\_MAX}} - T_{\text{A\_MAX}})}{R\theta_{JA}}$$

Where:

 $T_{J MAX}$  = maximum junction temperature allowed

 $T_{A\_MAX}$  = the maximum ambient temperature allowed

 $R\theta_{JA}$  = the thermal resistance from junction-to-air

Given the following example:

 $V_{IN}$  = 3.0V to 4.1V  $V_{OUT}$  = 2.8V ±2.5%

 $I_{LOAD}$  = 60 mA (output current)  $T_{AMAX}$  = 55°C (max. ambient temp.)

#### Find:

1. Internal power dissipation:

$$\begin{split} P_{DMAX} &= (V_{\text{IN\_MAX}} - V_{\text{OUT\_MIN}}) \times I_{LOAD} \\ &= (4.1V - 2.8 \times (0.975)) \times 60mA \\ &= 82.2mW \end{split}$$

2. Junction temperature:

$$T_{\text{J\_MAX}} = P_{DMAX} \times R\theta_{JA}$$

$$= 82.2 mWatts \times 450^{\circ} C/W + T_{AMAX}$$

$$= 37^{\circ} C + 55^{\circ} C$$

$$= 92^{\circ} C$$

3. Maximum allowable dissipation:

$$P_D = \frac{T_{\text{J\_MAX}} - T_{\text{A\_MAX}}}{R\theta_{JA}}$$
$$= \frac{125^{\circ}C - 55^{\circ}C}{450^{\circ}C/W}$$
$$= 155mW$$

In this example, the TC1016 dissipates approximately 82.2 mW and the junction temperature is raised 37°C over the 55°C ambient to 92°C. The absolute maximum power dissipation is 155 mW when given a maximum ambient temperature of 55°C.

Input voltage, output voltage or load current limits can also be determined by substituting known values in Equation 5-2 and Equation 5-3.

#### 5.3 Layout Considerations

The primary path for heat conduction out of the SC-70 package is through the package leads. Using heavy, wide traces at the pads of the device will facilitate the removal of heat within the package, thus lowering the thermal resistance  $R\theta_{JA}.$  By lowering the thermal resistance, the maximum internal power dissipation capability of the package is increased.

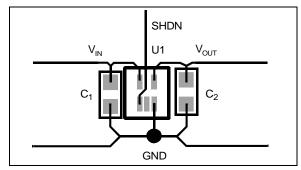
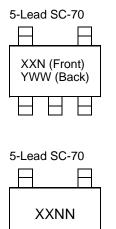


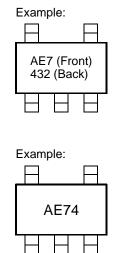
FIGURE 5-1: Suggested layout

### 6.0 PACKAGE INFORMATION

## 6.1 Package Marking Information



Part Number	Code
TC1016 - 1.8VLT	AE
TC1016 - 1.85VLT	AW
TC1016 - 2.6VLT	AF
TC1016 - 2.7VLT	AG
TC1016 - 2.8VLT	AH
TC1016 - 2.85VLT	AJ
TC1016 - 2.9VLT	AK
TC1016 - 3.0VLT	AL
TC1016 - 3.3VLT	AM
TC1016 - 4.0VLT	AP
·	·

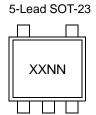


Legend: XX...X Customer-specific information\*
Y Year code (last digit of calendar year)
YY Year code (last 2 digits of calendar year)
WW Week code (week of January 1 is week '01')
NNN Alphanumeric traceability code

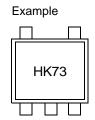
(a) Pb-free JEDEC designator for Matte Tin (Sn)
\* This package is Pb-free. The Pb-free JEDEC designator ((a))
can be found on the outer packaging for this package.

**Note:** In the event the full Microchip part number cannot be marked on one line, it will be carried over to the next line, thus limiting the number of available characters for customer-specific information.

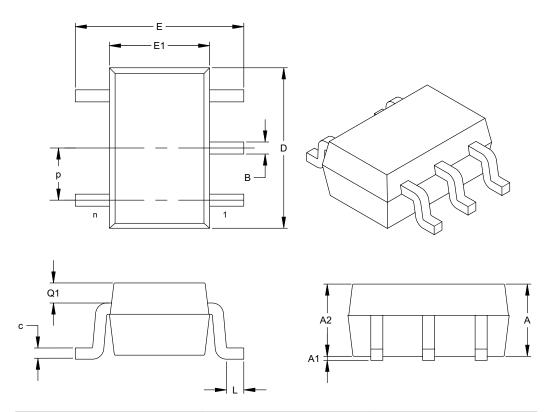
# 6.1 Package Marking Information (Continued)



Part Number	Code
<b>TC1016</b> – 1.8VCT	HK
<b>TC1016</b> – 1.85VCT	HW
<b>TC1016</b> – 2.6VCT	HL
<b>TC1016</b> – 2.7VCT	HM
<b>TC1016</b> – 2.8VCT	HP
<b>TC1016</b> – 2.85VCT	HQ
<b>TC1016</b> – 2.9VCT	HR
<b>TC1016</b> – 3.0VCT	HS
<b>TC1016</b> – 3.3VCT	HT
<b>TC1016</b> – 4.0VCT	HU



# 5-Lead Plastic Small Outline Transistor (LT) (SC-70)



	Units	INCHES			MILLIMETERS*		
Dimension	MIN	NOM	MAX	MIN	MOM	MAX	
Number of Pins	n	5 5					
Pitch	р		.026 (BSC)		0.65 (BSC)		
Overall Height	Α	.031		.043	0.80		1.10
Molded Package Thickness	A2	.031		.039	0.80		1.00
Standoff	A1	.000		.004	0.00		0.10
Overall Width	E	.071		.094	1.80		2.40
Molded Package Width	E1	.045		.053	1.15		1.35
Overall Length	D	.071		.087	1.80		2.20
Foot Length	L	.004		.012	0.10		0.30
Top of Molded Pkg to Lead Shoulder	Q1	.004		.016	0.10		0.40
Lead Thickness	С	.004		.007	0.10		0.18
Lead Width	В	.006		.012	0.15		0.30

<sup>\*</sup>Controlling Parameter

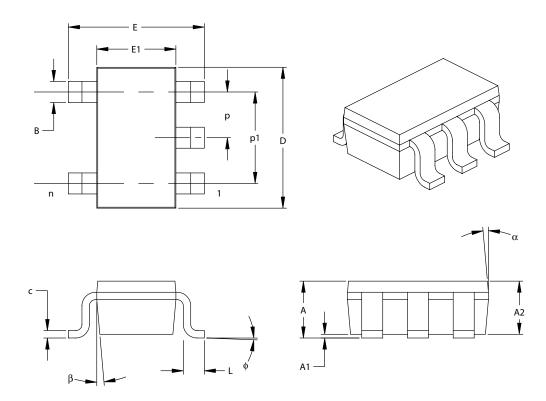
Notes:

Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .005" (0.127mm) per side.

JEITA (EIAJ) Standard: SC-70

Drawing No. C04-061

# 5-Lead Plastic Small Outline Transistor (OT) (SOT-23)



	Units	INCHES*			MILLIMETERS		
Dimension Limits		MIN	NOM	MAX	MIN	NOM	MAX
Number of Pins	n		5		5		
Pitch	р		.038			0.95	
Outside lead pitch (basic)	p1		.075			1.90	
Overall Height	Α	.035	.046	.057	0.90	1.18	1.45
Molded Package Thickness	A2	.035	.043	.051	0.90	1.10	1.30
Standoff	A1	.000	.003	.006	0.00	0.08	0.15
Overall Width	E	.102	.110	.118	2.60	2.80	3.00
Molded Package Width	E1	.059	.064	.069	1.50	1.63	1.75
Overall Length	D	.110	.116	.122	2.80	2.95	3.10
Foot Length	L	.014	.018	.022	0.35	0.45	0.55
Foot Angle	ф	0	5	10	0	5	10
Lead Thickness	С	.004	.006	.008	0.09	0.15	0.20
Lead Width	В	.014	.017	.020	0.35	0.43	0.50
Mold Draft Angle Top	α	0	5	10	0	5	10
Mold Draft Angle Bottom	β	0	5	10	0	5	10

\*Controlling Parameter

Notes:

Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .005" (0.127mm) per side.

EIAJ Equivalent: SC-74A Drawing No. C04-091

NOTES:

### APPENDIX A: REVISION HISTORY

### Revision B (March 2005)

- Updated Section 6.0 "Package Information" to include old and new packaging examples, as well as replaced SC-70 package diagram with up-todate version. Added additional voltage options
- Added SOT-23 package and voltage options.
- Applied new template and rearranged sections to be consistent with current documentation.

## .Revision A (October 2001)

• Original Release of this Document.

NOTES:

# PRODUCT IDENTIFICATION SYSTEM

To order or obtain information, e.g., on pricing or delivery, refer to the factory or the listed sales office.

PART NO. X	.xx x xxxx	Exa	amples:	
	lage Temperature Package tions Range	a)	TC1016-1.8VCTTR:	80 mA Tiny CMOS LDO with Shutdown, SOT-23 Package.
Device:	TC1016: 80 mA Tiny CMOS LDO with Shutdown	a)	TC1016-1.8VLTTR:	80 mA Tiny CMOS LDO with Shutdown, SC-70 Package.
Voltage Options*: (Standard)	1.8V 1.85V 2.6V	b)	TC1016-1.85VCTTR:	80 mA Tiny CMOS LDO with Shutdown, SOT-23 Package.
	2.7V 2.8V 2.85V	c)	TC1016-1.85VLTTR:	80 mA Tiny CMOS LDO with Shutdown, SC-70 Package.
	2.9V 3.0V 3.3V	d)	TC1016-2.6VCTTR:	80 mA Tiny CMOS LDO with Shutdown, SOT-23 Package.
	Other voltage options available. Please contact your local Microchip sales office for details.	e)	TC1016-2.6VLTTR:	80 mA Tiny CMOS LDO with Shutdown, SC-70 Package.
Temperature Range:	V = -40°C to +125°C	f)	TC1016-2.7VCTTR:	80 mA Tiny CMOS LDO with Shutdown, SOT-23 Package.
Packages:	LTTR = 5-pin SC-70 (Tape and Reel)  CTTR = 5-pin SOT-23 (Tape and Reel)	g)	TC1016-2.7VLTTR:	80 mA Tiny CMOS LDO with Shutdown, SC-70 Package.
		h)	TC1016-2.8VCTTR:	80 mA Tiny CMOS LDO with Shutdown, SOT-23 Package.
		i)	TC1016-2.8VLTTR:	80 mA Tiny CMOS LDO with Shutdown, SC-70 Package.
		j)	TC1016-2.85VLTTR:	80 mA Tiny CMOS LDO with Shutdown, SC-70 Package.

NOTES:

#### Note the following details of the code protection feature on Microchip devices:

- · Microchip products meet the specification contained in their particular Microchip Data Sheet.
- Microchip believes that its family of products is one of the most secure families of its kind on the market today, when used in the intended manner and under normal conditions.
- There are dishonest and possibly illegal methods used to breach the code protection feature. All of these methods, to our knowledge, require using the Microchip products in a manner outside the operating specifications contained in Microchip's Data Sheets. Most likely, the person doing so is engaged in theft of intellectual property.
- Microchip is willing to work with the customer who is concerned about the integrity of their code.
- Neither Microchip nor any other semiconductor manufacturer can guarantee the security of their code. Code protection does not
  mean that we are guaranteeing the product as "unbreakable."

Code protection is constantly evolving. We at Microchip are committed to continuously improving the code protection features of our products. Attempts to break Microchip's code protection feature may be a violation of the Digital Millennium Copyright Act. If such acts allow unauthorized access to your software or other copyrighted work, you may have a right to sue for relief under that Act.

Information contained in this publication regarding device applications and the like is provided only for your convenience and may be superseded by updates. It is your responsibility to ensure that your application meets with your specifications. MICROCHIP MAKES NO REPRESENTATIONS OR WAR-RANTIES OF ANY KIND WHETHER EXPRESS OR IMPLIED, WRITTEN OR ORAL, STATUTORY OR OTHERWISE, RELATED TO THE INFORMATION, INCLUDING BUT NOT LIMITED TO ITS CONDITION, QUALITY, PERFORMANCE, MERCHANTABILITY OR FITNESS FOR PURPOSE. Microchip disclaims all liability arising from this information and its use. Use of Microchip's products as critical components in life support systems is not authorized except with express written approval by Microchip. No licenses are conveyed, implicitly or otherwise, under any Microchip intellectual property rights.

#### **Trademarks**

The Microchip name and logo, the Microchip logo, Accuron, dsPIC, KEELOQ, microID, MPLAB, PIC, PICmicro, PICSTART, PRO MATE, PowerSmart, rfPIC, and SmartShunt are registered trademarks of Microchip Technology Incorporated in the U.S.A. and other countries.

AmpLab, FilterLab, Migratable Memory, MXDEV, MXLAB, PICMASTER, SEEVAL, SmartSensor and The Embedded Control Solutions Company are registered trademarks of Microchip Technology Incorporated in the U.S.A.

Analog-for-the-Digital Age, Application Maestro, dsPICDEM, dsPICDEM.net, dsPICworks, ECAN, ECONOMONITOR, FanSense, FlexROM, fuzzyLAB, In-Circuit Serial Programming, ICSP, ICEPIC, MPASM, MPLIB, MPLINK, MPSIM, PICkit, PICDEM, PICDEM.net, PICLAB, PICtail, PowerCal, PowerInfo, PowerMate, PowerTool, rfLAB, rfPICDEM, Select Mode, Smart Serial, SmartTel, Total Endurance and WiperLock are trademarks of Microchip Technology Incorporated in the U.S.A. and other countries.

 $\ensuremath{\mathsf{SQTP}}$  is a service mark of Microchip Technology Incorporated in the U.S.A.

All other trademarks mentioned herein are property of their respective companies.

© 2005, Microchip Technology Incorporated, Printed in the U.S.A., All Rights Reserved.

Printed on recycled paper.

QUALITY MANAGEMENT SYSTEM

CERTIFIED BY DNV

ISO/TS 16949:2002

Microchip received ISO/TS-16949:2002 quality system certification for its worldwide headquarters, design and wafer fabrication facilities in Chandler and Tempe, Arizona and Mountain View, California in October 2003. The Company's quality system processes and procedures are for its PICmicro® 8-bit MCUs, KEELOQ® code hopping devices, Serial EEPROMs, microperipherals, nonvolatile memory and analog products. In addition, Microchip's quality system for the design and manufacture of development systems is ISO 9001:2000 certified.



# WORLDWIDE SALES AND SERVICE

#### **AMERICAS**

**Corporate Office** 

2355 West Chandler Blvd. Chandler, AZ 85224-6199 Tel: 480-792-7200

Tel: 480-792-7200 Fax: 480-792-7277 Technical Support:

http://support.microchip.com

Web Address: www.microchip.com

**Atlanta** 

Alpharetta, GA Tel: 770-640-0034 Fax: 770-640-0307

**Boston** 

Westborough, MA Tel: 774-760-0087 Fax: 774-760-0088

Chicago Itasca, IL

Tel: 630-285-0071 Fax: 630-285-0075

**Dallas** 

Addison, TX Tel: 972-818-7423 Fax: 972-818-2924

Detroit

Farmington Hills, MI Tel: 248-538-2250 Fax: 248-538-2260

Kokomo

Kokomo, IN Tel: 765-864-8360 Fax: 765-864-8387

Los Angeles

Mission Viejo, CA Tel: 949-462-9523 Fax: 949-462-9608

San Jose

Mountain View, CA Tel: 650-215-1444 Fax: 650-961-0286

**Toronto** 

Mississauga, Ontario,

Canada

Tel: 905-673-0699 Fax: 905-673-6509

#### ASIA/PACIFIC

Australia - Sydney

Tel: 61-2-9868-6733 Fax: 61-2-9868-6755

China - Beijing

Tel: 86-10-8528-2100 Fax: 86-10-8528-2104

China - Chengdu

Tel: 86-28-8676-6200 Fax: 86-28-8676-6599

China - Fuzhou

Tel: 86-591-8750-3506 Fax: 86-591-8750-3521

China - Hong Kong SAR

Tel: 852-2401-1200 Fax: 852-2401-3431

**China - Shanghai** Tel: 86-21-5407-5533 Fax: 86-21-5407-5066

China - Shenyang

Tel: 86-24-2334-2829 Fax: 86-24-2334-2393

China - Shenzhen

Tel: 86-755-8203-2660 Fax: 86-755-8203-1760

China - Shunde

Tel: 86-757-2839-5507 Fax: 86-757-2839-5571

China - Qingdao

Tel: 86-532-502-7355 Fax: 86-532-502-7205

#### ASIA/PACIFIC

India - Bangalore

Tel: 91-80-2229-0061 Fax: 91-80-2229-0062

India - New Delhi

Tel: 91-11-5160-8631 Fax: 91-11-5160-8632

Japan - Kanagawa

Tel: 81-45-471-6166 Fax: 81-45-471-6122

Korea - Seoul

Tel: 82-2-554-7200 Fax: 82-2-558-5932 or

82-2-558-5934

**Singapore** Tel: 65-6334-8870 Fax: 65-6334-8850

Taiwan - Kaohsiung

Tel: 886-7-536-4818 Fax: 886-7-536-4803

Taiwan - Taipei

Tel: 886-2-2500-6610 Fax: 886-2-2508-0102

Taiwan - Hsinchu

Tel: 886-3-572-9526 Fax: 886-3-572-6459

#### **EUROPE**

Austria - Weis

Tel: 43-7242-2244-399 Fax: 43-7242-2244-393

Denmark - Ballerup

Tel: 45-4450-2828 Fax: 45-4485-2829

France - Massy

Tel: 33-1-69-53-63-20 Fax: 33-1-69-30-90-79

**Germany - Ismaning** 

Tel: 49-89-627-144-0 Fax: 49-89-627-144-44

Italy - Milan

Tel: 39-0331-742611 Fax: 39-0331-466781

Netherlands - Drunen Tel: 31-416-690399

Fax: 31-416-690340 **England - Berkshire** Tel: 44-118-921-5869

Fax: 44-118-921-5820

03/01/05