

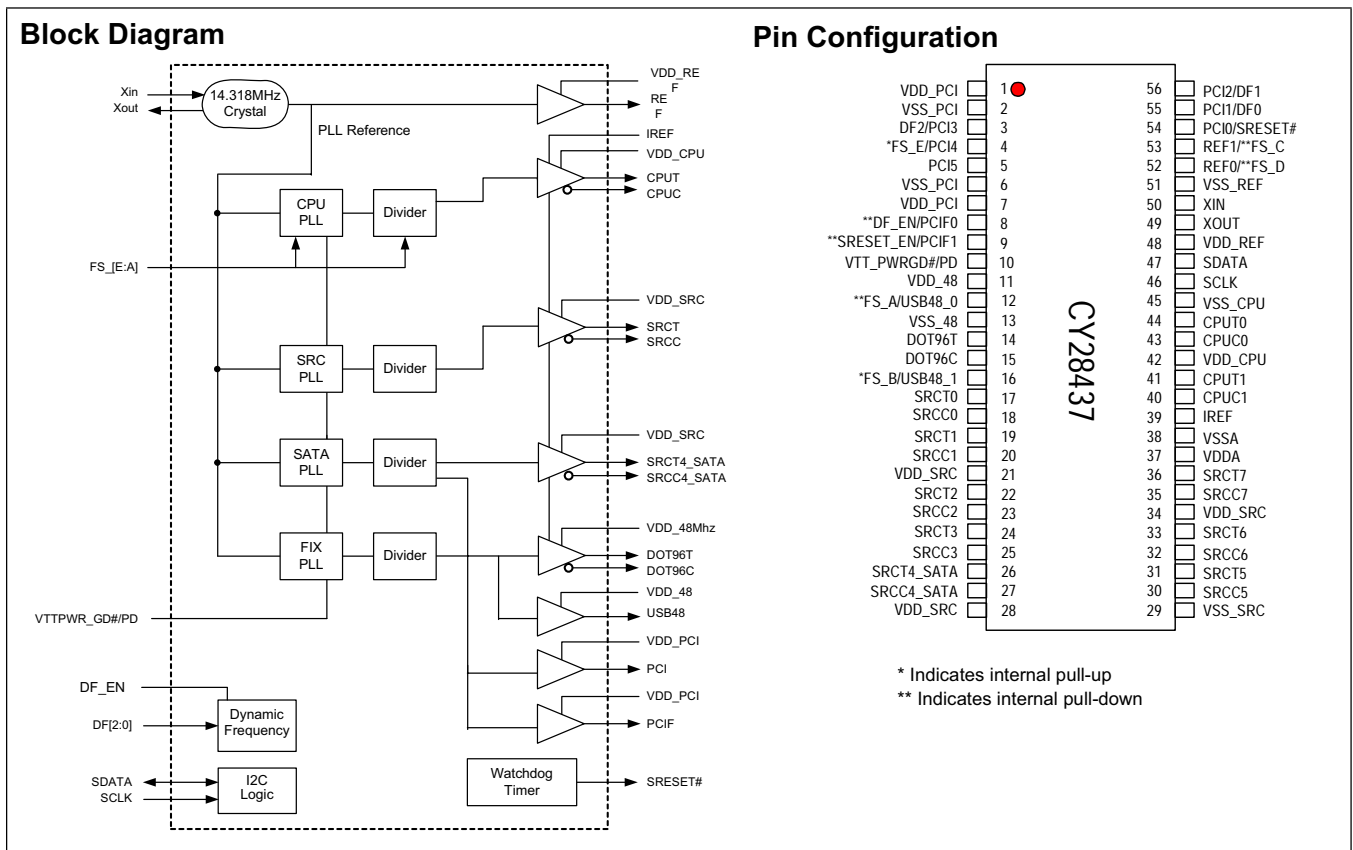
Clock Generator for Intel® Grantsdale Chipset

Features

- Compliant to Intel® CK410
- Supports Intel Prescott and Tejas CPU
- Selectable CPU frequencies
- Differential CPU clock pairs
- 100 MHz differential SRC clocks
- 96 MHz differential dot clock
- 48 MHz USB clocks
- 33 MHz PCI clock
- Dynamic Frequency Control

- Dial-A-Frequency®
- Watchdog timer
- Two Independent Overclocking PLLs
- Low-voltage frequency select input
- I²C support with readback capabilities
- Ideal Lexmark Spread Spectrum profile for maximum electromagnetic interference (EMI) reduction
- 3.3V power supply
- 56-pin SSOP and TSSOP packages

CPU	SRC	PCI	REF	DOT96	USB
x 2	x 8	x 8	x 2	x 1	x 2



Pin Description

Pin No.	Name	Type	Description
1,7	VDD_PCI	PWR	3.3V power supply for outputs.
2,6	VSS_PCI	GND	Ground for outputs.
4	FS_E/PCI4	I/O, SE, PU	3.3V-tolerant input for CPU frequency selection/33-MHz clock. <i>Refer to DC Electrical Specifications table for V_{il_FS} and V_{ih_FS} specifications.</i>
5	PCI	O, SE	33 MHz clocks.
8	DF_EN/PCIF0	I/O, SE PD	3.3V LVTTTL input to Enable DF pin/33-MHz Output. 1 = Enable, 0 = Disable. Intel Type-5 output buffer
9	SRESET_EN/PCIF1	I/O, PD, SE	3.3V LVTTTL input to enable Watchdog/33-MHz clocks. 1 = Enable, 0 = Disable
10	VTT_PWRGD#/PD	I, PD	3.3V LVTTTL input. This pin is a level sensitive strobe used to latch the FS_A, FS_B, FS_C, FS_D, and FSE inputs. After VTT_PWRGD# (active LOW) assertion, this pin becomes a real-time input for asserting power-down (active HIGH).
11	VDD_48	PWR	3.3V power supply for outputs.
12	FS_A/USB48_0	I/O, PD, SE	3.3V-tolerant input for CPU frequency selection/fixed 48-MHz clock output. <i>Refer to DC Electrical Specifications table for V_{il_FS} and V_{ih_FS} specifications.</i>
13	VSS_48	GND	Ground for outputs.
14,15	DOT96T, DOT96C	O, DIF	Fixed 96-MHz clock output.
16	FS_B/USB48_1	I/O, PU, SE	3.3V-tolerant input for CPU frequency selection/fixed 48-MHz clock output. <i>Refer to DC Electrical Specifications table for V_{il_FS} and V_{ih_FS} specifications.</i>
17,18,19,20,22,23,24,25,31,30,33,32,35,36	SRCT/C	O, DIF	Differential serial reference clocks. Outputs have overclocking capability.
21,28,34	VDD_SRC	PWR	3.3V power supply for outputs.
26,27	SRC4_SATAT, SRC4_SATAC	O, DIF	Differential serial reference clock. Recommended output for SATA.
29	VSS_SRC	GND	Ground for outputs.
37	VDDA	PWR	3.3V power supply for PLL.
38	VSSA	GND	Ground for PLL.
39	IREF	I	A precision resistor is attached to this pin, which is connected to the internal current reference.
42	VDD_CPU	PWR	3.3V power supply for outputs.
44,43,41,40	CPUT/C	O, DIF	Differential CPU clock outputs.
45	VSS_CPU	GND	Ground for outputs.
46	SCLK	I	SMBus-compatible SCLOCK.
47	SDATA	I/O	SMBus-compatible SDATA.
48	VDD_REF	PWR	3.3V power supply for outputs.
49	XOUT	O, SE	14.318 MHz crystal output.
50	XIN	I	14.318 MHz crystal input.
51	VSS_REF	GND	Ground for outputs.
52	FS_D/REF0	I/O, SE, PD	3.3V-tolerant input for CPU frequency selection/Reference clock. <i>Refer to DC Electrical Specifications table for V_{il_FS} and V_{ih_FS} specifications.</i>
53	FS_C/REF1	I/O, SE, PD	3.3V-tolerant input for CPU frequency selection/Reference clock. Selects test mode if pulled to V_{IHFS_C} when VTT_PWRGD# is asserted LOW. <i>Refer to DC Electrical Specifications table for V_{ILFS_C}, V_{IMFS_C}, V_{IHFS_C} specifications.</i>
54	PCI0/SRESET#	O PU	33 MHz clocks/3.3V LVTTTL output for Watchdog reset. When configured as SRESET# output this output becomes open drain type with a high (>100k Ω) internal pull-up resistor.
3,55,56	DF/PCI	I/O, SE	3.3V LVTTTL input for Dynamic Frequency/33-MHz clocks output.

Frequency Select Pins (FS_[A:E])

Host clock frequency selection is achieved by applying the appropriate logic levels to FS_A, FS_B, FS_C, FS_D, and FS_E inputs prior to VTT_PWRGD# assertion (as seen by the clock synthesizer). Upon VTT_PWRGD# being sampled LOW by the clock chip (indicating processor VTT voltage is stable), the clock chip samples the FS_A, FS_B, FS_C, FS_D, and FS_E input values. For all logic levels of FS_A, FS_B, FS_C, FS_D, and FS_E, VTT_PWRGD# employs a one-shot functionality in that once a valid LOW on VTT_PWRGD# has been sampled, all further VTT_PWRGD#, FS_A, FS_B, FS_C, FS_D, and FS_E transitions will be ignored, except in test mode.

FS_C is a three-level input, when sampled at a voltage greater than 2.1V by VTT_PWRGD#, the device will enter test mode as selected by the voltage level on the FS_B input.

Serial Data Interface

To enhance the flexibility and function of the clock synthesizer, a two-signal serial interface is provided. Through the Serial Data Interface, various device functions, such as individual clock output buffers, can be individually enabled or disabled. The registers associated with the Serial Data Interface initialize to their default setting upon power-up, and therefore use of this interface is optional. Clock device register changes are normally made upon system initialization, if any are required. The interface cannot be used during system operation for power management functions.

Input Conditions				Output Frequency									
FS_D	FS_C	FS_B	FS_A	CPU	SRC	CPU PLL Gear Constants	CPU M divider	CPU N DEFAULT	CPU N allowable range for DAF	SRC PLL Gear Constants	SRC M divider (not changeable by user)	SRC N DEFAULT	SRC N allowable range for DAF
FSEL_3	FSEL_2	FSEL_1	FSEL_0	(MHz)	(MHz)	(G)							
0	1	0	1	100	100	30	60	200	200-250	30	60	200	200-266
0	0	0	1	133.3333333	100	40	60	200	200-250	30	60	200	200-266
0	0	1	1	166.6666667	100	60	63	175	175-262	30	60	200	200-266
0	0	1	0	200	100	60	60	200	200-250	30	60	200	200-266
0	0	0	0	266.6666667	100	80	60	200	200-250	30	60	200	200-266
0	1	0	0	333.3333333	100	120	63	175	175-262	30	60	200	200-266
0	1	1	0	400	100	120	60	200	200-250	30	60	200	200-266
1	1	0	1	100.952381	100	30	63	212	212-262	30	60	200	200-266
1	0	0	1	133.968254	100	40	63	211	211-262	30	60	200	200-266
1	0	1	1	167	100	60	60	167	167-250	30	60	200	200-266
1	0	1	0	200.952381	100	60	63	211	211-262	30	60	200	200-266
1	0	0	0	266.6666667	100	80	60	200	200-250	30	60	200	200-266
1	1	0	0	334	100	120	60	167	167-250	30	60	200	167-266
1	1	1	0	400.6451613	100	120	62	207	207-258	30	60	200	167-266

X	HIGH	LOW	X	Tristate	Tristate	Tristate	Tristate	Tristate	Tristate
X	HIGH	HIGH	X	REF/N	REF/N	REF/N	REF/N	REF/N	REF/N

Figure 1. CPU and SRC Frequency Select Tables

Data Protocol

The clock driver serial protocol accepts byte write, byte read, block write, and block read operations from the controller. For block write/read operation, the bytes must be accessed in sequential order from lowest to highest byte (most significant bit first) with the ability to stop after any complete byte has been transferred. For byte write and byte read operations, the

system controller can access individually indexed bytes. The offset of the indexed byte is encoded in the command code, as described in *Table 1*.

The block write and block read protocol is outlined in *Table 2* while *Table 3* outlines the corresponding byte write and byte read protocol. The slave receiver address is 11010010 (D2h).

Table 1. Command Code Definition

Bit	Description
7	0 = Block read or block write operation, 1 = Byte read or byte write operation
(6:0)	Byte offset for byte read or byte write operation. For block read or block write operations, these bits should be '0000000'

Table 2. Block Read and Block Write Protocol

Block Write Protocol		Block Read Protocol	
Bit	Description	Bit	Description
1	Start	1	Start
8:2	Slave address – 7 bits	8:2	Slave address – 7 bits
9	Write	9	Write
10	Acknowledge from slave	10	Acknowledge from slave
18:11	Command Code – 8 bits	18:11	Command Code – 8 bits
19	Acknowledge from slave	19	Acknowledge from slave
27:20	Byte Count – 8 bits (Skip this step if I ² C_EN bit set)	20	Repeat start
28	Acknowledge from slave	27:21	Slave address – 7 bits
36:29	Data byte 1 – 8 bits	28	Read = 1
37	Acknowledge from slave	29	Acknowledge from slave
45:38	Data byte 2 – 8 bits	37:30	Byte Count from slave – 8 bits
46	Acknowledge from slave	38	Acknowledge
....	Data Byte /Slave Acknowledges	46:39	Data byte 1 from slave – 8 bits
....	Data Byte N – 8 bits	47	Acknowledge
....	Acknowledge from slave	55:48	Data byte 2 from slave – 8 bits
....	Stop	56	Acknowledge
		Data bytes from slave / Acknowledge
		Data Byte N from slave – 8 bits
		NOT Acknowledge
		Stop

Table 3. Byte Read and Byte Write Protocol

Byte Write Protocol		Byte Read Protocol	
Bit	Description	Bit	Description
1	Start	1	Start
8:2	Slave address – 7 bits	8:2	Slave address – 7 bits
9	Write	9	Write
10	Acknowledge from slave	10	Acknowledge from slave
18:11	Command Code – 8 bits	18:11	Command Code – 8 bits
19	Acknowledge from slave	19	Acknowledge from slave
27:20	Data byte – 8 bits	20	Repeated start

Table 3. Byte Read and Byte Write Protocol (continued)

Byte Write Protocol		Byte Read Protocol	
Bit	Description	Bit	Description
28	Acknowledge from slave	27:21	Slave address – 7 bits
29	Stop	28	Read
		29	Acknowledge from slave
		37:30	Data from slave – 8 bits
		38	NOT Acknowledge
		39	Stop

Control Registers

Byte 0: Control Register 0

Bit	@Pup	Name	Description
7	1	SRC[T/C]7	SRC[T/C]7 Output Enable 0 = Disable (Tri-state), 1 = Enable
6	1	SRC[T/C]6	SRC[T/C]6 Output Enable 0 = Disable (Tri-state), 1 = Enable
5	1	SRC[T/C]5	SRC[T/C]5 Output Enable 0 = Disable (Tri-state), 1 = Enable
4	1	SRC[T/C]4_SATA	SRC[T/C]4_SATA Output Enable 0 = Disable (Tri-state), 1 = Enable
3	1	SRC[T/C]3	SRC[T/C]3 Output Enable 0 = Disable (Tri-state), 1 = Enable
2	1	SRC[T/C]2	SRC[T/C]2 Output Enable 0 = Disable (Tri-state), 1 = Enable
1	1	SRC[T/C]1	SRC[T/C]1 Output Enable 0 = Disable (Tri-state), 1 = Enable
0	1	SRC[T/C]0	SRC[T/C]0 Output Enable 0 = Disable (Tri-state), 1 = Enable

Byte 1: Control Register 1

Bit	@Pup	Name	Description
7	1	PCIF0	PCIF0 Output Enable 0 = Disabled, 1 = Enabled
6	1	DOT96[T/C]	DOT96[T/C]MHz Output Enable 0 = Disable (Tri-state), 1 = Enabled
5	1	USB_0	USB_0 MHz Output Enable 0 = Disabled, 1 = Enabled
4	1	REF	REF Output Enable 0 = Disabled, 1 = Enabled
3	0	RESERVED	RESERVED, Set = 0
2	1	CPU[T/C]1	CPU[T/C]1 Output Enable 0 = Disable (Tri-state), 1 = Enabled
1	1	CPU[T/C]0	CPU[T/C]0 Output Enable 0 = Disable (Tri-state), 1 = Enabled
0	0	CPU	PLL1 (CPU PLL) Spread Spectrum Enable 0 = Spread off, 1 = Spread on

Byte 2: Control Register 2

Bit	@Pup	Name	Description
7	1	PCI5	PCI5 Output Enable 0 = Disabled, 1 = Enabled
6	1	PCI4	PCI4 Output Enable 0 = Disabled, 1 = Enabled
5	1	PCI3	PCI3 Output Enable 0 = Disabled, 1 = Enabled
4	1	PCI2	PCI2 Output Enable 0 = Disabled, 1 = Enabled
3	1	PCI1	PCI1 Output Enable 0 = Disabled, 1 = Enabled
2	1	PCI0	PCI0 Output Enable 0 = Disabled, 1 = Enabled
1	1	RESERVED	RESERVED, Set = 1
0	1	PCIF1	PCIF1 Output Enable 0 = Disabled, 1 = Enabled

Byte 3: Control Register 3

Bit	@Pup	Name	Description
7	0	SRC[T/C]7	Allow control of SRC[T/C]7 with assertion of SW PCI_STP# 0 = Free running, 1 = Stopped with PCI_STP#
6	0	SRC[T/C]6	Allow control of SRC[T/C]6 with assertion of SW PCI_STP# 0 = Free running, 1 = Stopped with PCI_STP#
5	0	SRC[T/C]5	Allow control of SRC[T/C]5 with assertion of SW PCI_STP# 0 = Free running, 1 = Stopped with PCI_STP#
4	0	SRC[T/C]4_SATA	Allow control of SRC[T/C]4_SATA with assertion of SW PCI_STP# 0 = Free running, 1 = Stopped with PCI_STP#
3	0	SRC[T/C]3	Allow control of SRC[T/C]3 with assertion of SW PCI_STP# 0 = Free running, 1 = Stopped with PCI_STP#
2	0	SRC[T/C]2	Allow control of SRC[T/C]2 with assertion of SW PCI_STP# 0 = Free running, 1 = Stopped with PCI_STP#
1	0	SRC[T/C]1	Allow control of SRC[T/C]1 with assertion of SW PCI_STP# 0 = Free running, 1 = Stopped with PCI_STP#
0	0	SRC[T/C]0	Allow control of SRC[T/C]0 with assertion of SW PCI_STP# 0 = Free running, 1 = Stopped with PCI_STP#

Byte 4: Control Register 4

Bit	@Pup	Name	Description
7	HW	FS_E	FS_E Reflects the value of the FS_E pin sampled on power-up. 0 = FS_E was LOW during VTT_PWRGD# assertion.
6	0	DOT96	DOT_PWRDWN Drive Mode 0 = Driven in PWRDWN, 1 = Tri-state
5	0	RESERVED	RESERVED, Set = 0
4	0	PCIF1	Allow control of PCIF1 with assertion of SW PCI_STP# 0 = Free running, 1 = Stopped with PCI_STP#
3	0	PCIF0	Allow control of PCIF0 with assertion of SW PCI_STP# 0 = Free running, 1 = Stopped with PCI_STP#
2	1	RESERVED	RESERVED, Set = 1
1	1	RESERVED	RESERVED, Set = 1
0	1	RESERVED	RESERVED, Set = 1

Byte 5: Control Register 5

Bit	@Pup	Name	Description
7	0	SRC	SRC Stop Drive Mode 0 = Driven when PCI_STP# asserted, 1 = Tri-state when PCI_STP# asserted
6	0	RESERVED	RESERVED, Set = 0
5	0	RESERVED	RESERVED, Set = 0
4	0	RESERVED	RESERVED, Set = 0
3	0	SRC[7:0]	SRC PWRDWN Drive Mode 0 = Driven when PD asserted, 1 = Tri-state when PD asserted
2	0	RESERVED	RESERVED, Set = 0
1	0	CPU[T/C]1	CPU[T/C]1 PWRDWN Drive Mode 0 = Driven when PD asserted, 1 = Tri-state when PD asserted
0	0	CPU[T/C]0	CPU[T/C]0 PWRDWN Drive Mode 0 = Driven when PD asserted, 1 = Tri-state when PD asserted

Byte 6: Control Register 6

Bit	@Pup	Name	Description
7	0	TEST_SEL	REF/N or Tri-state Select 0 = Tri-state, 1 = REF/N Clock
6	0	TEST_MODE	Test Clock Mode Entry Control 0 = Normal operation, 1 = REF/N or Tri-state mode,
5	HW	FS_D	FS_D reflects the value of the FS_D pin sampled on power-up. 0 = FS_D was LOW during VTT_PWRGD# assertion
4	1	REF0	REF Output Drive Strength 0 = High, 1 = Low
3	1	PCI, PCIF and SRC clock outputs except those set to free running	SW PCI_STP# Function 0=SW PCI_STP assert, 1= SW PCI_STP deassert When this bit is set to 0, all STOPPABLE PCI, PCIF and SRC outputs will be stopped in a synchronous manner with no short pulses. When this bit is set to 1, all STOPPED PCI, PCIF and SRC outputs will resume in a synchronous manner with no short pulses.
2	HW	FS_C	FS_C Reflects the value of the FS_C pin sampled on power-up 0 = FS_C was low during VTT_PWRGD# assertion
1	HW	FS_B	FS_B Reflects the value of the FS_B pin sampled on power-up 0 = FS_B was low during VTT_PWRGD# assertion
0	HW	FS_A	FS_A Reflects the value of the FS_A pin sampled on power-up 0 = FS_A was low during VTT_PWRGD# assertion

Byte 7: Vendor ID

Bit	@Pup	Name	Description
7	0	Revision Code Bit 3	Revision Code Bit 3
6	0	Revision Code Bit 2	Revision Code Bit 2
5	1	Revision Code Bit 1	Revision Code Bit 1
4	1	Revision Code Bit 0	Revision Code Bit 0
3	1	Vendor ID Bit 3	Vendor ID Bit 3
2	0	Vendor ID Bit 2	Vendor ID Bit 2
1	0	Vendor ID Bit 1	Vendor ID Bit 1
0	0	Vendor ID Bit 0	Vendor ID Bit 0

Byte 8: Control Register 8

Bit	@Pup	Name	Description
7	0	CPU_SS	Spread Selection for CPU PLL 0: -0.5% (peak to peak) 1: -1.0% (peak to peak)
6	0	CPU_DWN_SS	Spread Selection for CPU PLL 0: Down spread 1: Center spread
5	0	SRC_SS_OFF	SRC Spread Spectrum Enable 0 = Spread off, 1 = Spread on
4	0	SRC_SS	Spread Selection for SRC PLL 0: -0.5% (peak to peak) 1: -1.0% (peak to peak)
3	0	RESERVED	RESERVED, Set = 0
2	1	USB	48-MHz Output Drive Strength 0 = 2x, 1 = 1x
1	1	PCI	33-MHz Output Drive Strength 0 = 2x, 1 = 1x
0	0	RESERVED	RESERVED, Set = 0

Byte 9: Control Register 9

Bit	@Pup	Name	Description
7	0	DF_Limit2	Dynamic Frequency Max threshold. These three bits will set the max allowed CPU frequency for Dynamic Frequency
6	0	DF_Limit1	
5	0	DF_Limit0	
4	0	DF_EN	Dynamic Frequency Enable 0 = Disable, 1 = Enable
3	0	FSEL_D	SW Frequency selection bits. See <i>Table 1</i> .
2	0	FSEL_C	
1	0	FSEL_B	
0	0	FSEL_A	

Byte 10: Control Register 10

Bit	@Pup	Name	Description
7	0	Recovery_Frequency	This bit allows selection of the frequency setting that the clock will be restored to once the system is rebooted 0: Use HW settings, 1: Recovery N[8:0]
6	0	Timer_SEL	Timer_SEL selects the WD reset function at SRESET pin when WD time out. 0 = Reset and Reload Recovery_Frequency 1 = Only Reset
5	1	Time_Scale	Time_Scale allows selection of WD time scale 0 = 294 ms 1 = 2.34 s
4	0	WD_Alarm	WD_Alarm is set to "1" when the watchdog times out. It is reset to "0" when the system clears the WD_TIMER time stamp.
3	0	WD_TIMER2	Watchdog timer time stamp selection 000: Reserved (test mode) 001: 1 * Time_Scale 010: 2 * Time_Scale 011: 3 * Time_Scale 100: 4 * Time_Scale 101: 5 * Time_Scale 110: 6 * Time_Scale 111: 7 * Time_Scale
2	0	WD_TIMER1	
1	0	WD_TIMER0	

Byte 10: Control Register 10 (continued)

Bit	@Pup	Name	Description
0	0	WD_EN	Watchdog timer enable, when the bit is asserted, Watchdog timer is triggered and time stamp of WD_Timer is loaded 0 = Disable, 1 = Enable

Byte 11: Control Register 11

Bit	@Pup	Name	Description
7	0	CPU_DAF_N7	If Prog_CPU_EN is set, the values programmed in CPU_DAF_N[8:0] and CPU_DAF_M[6:0] will be used to determine the CPU output frequency. The setting of the FS_Override bit determines the frequency ratio for CPU and other output clocks. When it is cleared, the same frequency ratio stated in the Latched FS[E:A] register will be used. When it is set, the frequency ratio stated in the FSEL[3:0] register will be used.
6	0	CPU_DAF_N6	
5	0	CPU_DAF_N5	
4	0	CPU_DAF_N4	
3	0	CPU_DAF_N3	
2	0	CPU_DAF_N2	
1	0	CPU_DAF_N1	
0	0	CPU_DAF_N0	

Byte 12: Control Register 12

Bit	@Pup	Name	Description
7	0	CPU_DAF_N8	If Prog_CPU_EN is set, the values programmed is in CPU_FSEL_N[8:0] and CPU_FSEL_M[6:0] will be used to determine the CPU output frequency. The setting of the FS_Override bit determines the frequency ratio for CPU and other output clocks. When it is cleared, the same frequency ratio stated in the Latched FS[E:A] register will be used. When it is set, the frequency ratio stated in the FSEL[3:0] register will be used.
6	0	CPU_DAF_M6	
5	0	CPU_DAF_M5	
4	0	CPU_DAF_M4	
3	0	CPU_DAF_M3	
2	0	CPU_DAF_M2	
1	0	CPU_DAF_M1	
0	0	CPU_DAF_M0	

Byte 13: Control Register 13

Bit	@Pup	Name	Description
7	0	SRC_N7	SRC Dial-A-Frequency Bit N7
6	0	SRC_N6	SRC Dial-A-Frequency Bit N6
5	0	SRC_N5	SRC Dial-A-Frequency Bit N5
4	0	SRC_N4	SRC Dial-A-Frequency Bit N4
3	0	SRC_N3	SRC Dial-A-Frequency Bit N3
2	0	SRC_N2	SRC Dial-A-Frequency Bit N2
1	0	SRC_N1	SRC Dial-A-Frequency Bit N1
0	0	SRC_N0	SRC Dial-A-Frequency Bit N0

Byte 14: Control Register 14

Bit	@Pup	Name	Description
7	0	SRC_N8	SRC Dial-A-Frequency Bit N8
6	0	SW_RESET	Software Reset. When set the device will assert a reset signal on SRESET# upon completion of the block/word/byte write that set it. After asserting and deasserting the SRESET# this bit will self clear (set to 0). The SRESET# pin must be enabled by latching SRESET#_EN on VTT_PRWGD# to utilize this feature.
5	0	FS_[E:A]	FS_Override 0 = Select operating frequency by FS(D:A) input pins 1 = Select operating frequency by FSEL_(3:0) settings
4	0	SMSW_SEL	Smooth switch select 0: Select CPU_PLL 1: Select SRC_PLL
3	0	RESERVED	RESERVED, Set = 0
2	0	RESERVED	RESERVED, Set = 0
1	1	PCIF	Free running 33-MHz Output Drive Strength 0 = 2x, 1 = 1x
0	0	Recovery_N8	Watchdog Recovery Bit

Byte 15: Control Register 15

Bit	@Pup	Name	Description
7	0	Recovery N7	Watchdog Recovery Bit
6	0	Recovery N6	Watchdog Recovery Bit
5	0	Recovery N5	Watchdog Recovery Bit
4	0	Recovery N4	Watchdog Recovery Bit
3	0	Recovery N3	Watchdog Recovery Bit
2	0	Recovery N2	Watchdog Recovery Bit
1	0	Recovery N1	Watchdog Recovery Bit
0	0	Recovery N0	Watchdog Recovery Bit

Byte 16: Control Register 16

Bit	@Pup	Name	Description
7	1	REF1	REF1 Output Enable 0 = Disable, 1 = Enable
6	1	USB48_1	USB48_1 Output Enable 0 = Disable, 1 = Enable
5	0	SRC_FREQ_SEL	SRC Frequency selection 0: SRC frequency is selected via the FSE pin 1: SRC frequency is initially set to 167 MHz.
4	0	RESERVED	RESERVED, Set = 0
3	0	SRC_SATA	SATA PLL Spread Spectrum Enable 0 = Spread off, 1 = Spread on
2	0	Prog_SRC_EN	Programmable SRC frequency enable 0 = Disabled, 1 = Enabled.
1	0	Prog_CPU_EN	Programmable CPU frequency enable 0 = Disabled, 1 = Enabled.
0	1	Watchdog Autorecovery	Watchdog Autorecovery Mode 0 = Disable (Manual), 1= Enable (Auto)

Table 4. Crystal Recommendations

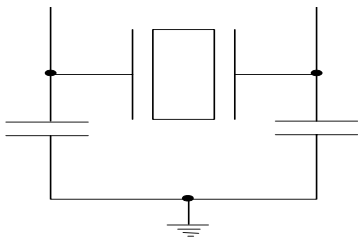
Frequency (Fund)	Cut	Loading	Load Cap	Drive (max.)	Shunt Cap (max.)	Motional (max.)	Tolerance (max.)	Stability (max.)	Aging (max.)
14.31818 MHz	AT	Parallel	20 pF	0.1 mW	5 pF	0.016 pF	35 ppm	30 ppm	5 ppm

The CY28437 requires a parallel resonance crystal. Substituting a series resonance crystal will cause the CY28437 to operate at the wrong frequency and violate the ppm specification. For most applications there is a 300-ppm frequency shift between series and parallel crystals due to incorrect loading.

Crystal Loading

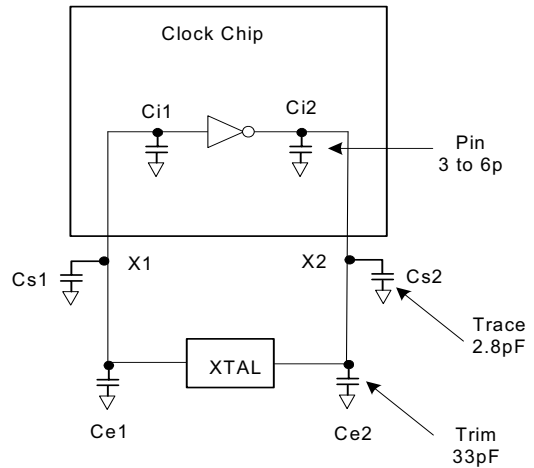
Crystal loading plays a critical role in achieving low ppm performance. To realize low ppm performance, the total capacitance the crystal will see must be considered to calculate the appropriate capacitive loading (CL).

Figure 2 shows a typical crystal configuration using the two trim capacitors. An important clarification for the following discussion is that the trim capacitors are in series with the crystal not parallel. It's a common misconception that load capacitors are in parallel with the crystal and should be approximately equal to the load capacitance of the crystal. This is not true.


Figure 2. Crystal Capacitive Clarification

Calculating Load Capacitors

In addition to the standard external trim capacitors, trace capacitance and pin capacitance must also be considered to correctly calculate crystal loading. As mentioned previously, the capacitance on each side of the crystal is in series with the crystal. This means the total capacitance on each side of the crystal must be twice the specified crystal load capacitance (CL). While the capacitance on each side of the crystal is in series with the crystal, trim capacitors (Ce1,Ce2) should be calculated to provide equal capacitive loading on both sides.


Figure 3. Crystal Loading Example

As mentioned previously, the capacitance on each side of the crystal is in series with the crystal. This means the total capacitance on each side of the crystal must be twice the specified load capacitance (CL). While the capacitance on each side of the crystal is in series with the crystal, trim capacitors (Ce1,Ce2) should be calculated to provide equal capacitive loading on both sides.

Use the following formulas to calculate the trim capacitor values for Ce1 and Ce2.

Load Capacitance (each side)

$$C_e = 2 * CL - (C_s + C_i)$$

Total Capacitance (as seen by the crystal)

$$CL_e = \frac{1}{\left(\frac{1}{C_{e1} + C_{s1} + C_{i1}} + \frac{1}{C_{e2} + C_{s2} + C_{i2}} \right)}$$

- CL.....Crystal load capacitance
- CL_e..... Actual loading seen by crystal using standard value trim capacitors
- C_e..... External trim capacitors
- C_s..... Stray capacitance (terraced)
- C_i Internal capacitance (lead frame, bond wires etc.)

Dynamic Frequency

Dynamic Frequency – Dynamic Frequency (DF) is a technique to increase the CPU frequency dynamically from any starting value. The user selects the starting point, either by HW, FSEL, or DAF, then enables DF. After that, DF will dynamically change as determined by the value on the DF[2:0] pins.

DF/PCI Pin – These PCI pins incorporate dual functions, either DF or PCI. The function is selected by the DF_EN pin. When used as DF, these three pins will map to eight entries that correspond to different “N” values for Dynamic Frequency. Below is a table that list the combinations along with the increase in “N”.

DOC[2:0]	DOC N value
000	Original Frequency
001	+2
010	+6
011	+10
100	+14
101	+18
110	+30
111	+40

DF_EN bit – This bit enables the DF mode. By default, it is not set. When set, the operating frequency is determined by DF[2:0] pins. **Default = 0, (No DF)**

DF_Limit bit – There are three bits that allow the user to set an upper limit to prevent CPU runaway. In the event that the user uses DAF with DF, this feature will provide some safeguard so the CPU won’t burn up.

Dial-A-Frequency (CPU and SRC)

This feature allows the user to overclock their system by slowly stepping up the CPU or SRC frequency. When the programmable output frequency feature is enabled, the CPU and SRC frequencies are determined by the following equation

$$F_{cpu} = G * N/M \text{ or } F_{cpu} = G2 * N, \text{ where } G2 = G / M$$

“N” and “M” are the values programmed in Programmable Frequency Select N-Value Register and M-Value Register, respectively. “G” stands for the PLL Gear Constant, which is determined by the programmed value of FS[E:A]. See *Figure 1* for the Gear Constant for each Frequency selection. The PCI Express only allows user control of the N register, the M value is fixed and documented in *Figure 1*.

In this mode, the user writes the desired N and M value into the DAF I2C registers. The user cannot change only the M value and must change both the M and the N values at the same time, if they require a change to the M value. The user may change only the N value if required.

Associated Register bits

CPU_DAF_Enable – This bit enables CPU DAF mode. By default, it is not set. When set, the operating frequency is determined by the values entered into the CPU_DAF_N register. **Note:** the CPU_DAF_N and M register must contain valid values before CPU_DAF is set. Default = 0 (No DAF)

CPU_DAF_N – There are 9 bits (for 512 values) to linearly change the CPU frequency (limited by VCO range). Default =

0 (0000) The allowable values for N are detailed in the frequency select table in *Figure 1*.

CPU_DAF_M – There are 7 bits (for 128 values) to linearly change the CPU frequency (limited by VCO range). Default = 0. The allowable values for M are detailed in the frequency select table in *Figure 1*.

SRC_DAF_Enable – This bit enables SRC DAF mode. By default, it is not set. When set, the operating frequency is determined by the values entered into the SRC_DAF_N register. **Note:** the SRC_DAF_N register must contain valid values before SRC_DAF is set. Default = 0 (No DAF)

SRC_DAF_N – There are 9 bits (for 512 values) to linearly change the CPU frequency (limited by VCO range). Default = 0 (0000) The allowable values for N are detailed in the frequency select table in *Figure 1*.

Recovery – The recovery mechanism during CPU DAF when the system locks up and the Watchdog timer is enabled is determined by the “Watchdog Recovery Mode” and “Watchdog Auto recovery Enable” bits. The possible recovery methods are: (A) Auto, (B) Manual (by Recovery N), (C) HW, and (D) No recovery—just send reset signal.

There is no recovery mode for SRC Dial-a-frequency.

Software Frequency Select

This mode allows the user to select the CPU output frequencies using the Software Frequency select bits in the SMBUS register.

FSEL – There are 4 bits (for 16 combinations) to select predetermined CPU frequencies from a table. The table selections are detailed in *Figure 1*.

FS_Override – This bit allows the CPU frequency to be selected from HW or FSEL settings. By default, this bit is not set and the CPU frequency is selected by HW. When this bit is set, the CPU frequency is selected by the FSEL bits. Default = 0

Recovery – The recovery mechanism during FSEL when the system locks up is determined by the “Watchdog Recovery Mode” and “Watchdog Auto recovery Enable” bits. The only possible recovery method is from the Hardware Settings. Auto recovery or manual recovery can cause a wrong output frequency because the output divider may have changed with the selected CPU frequency and these recovery methods will not recover the original output divider setting.

Smooth Switching

The device contains one smooth switch circuit that is shared by the CPU PLL and SRC PLL. The smooth switch circuit ensures that when the output frequency changes by overclocking, the transition from the old frequency to the new frequency is a slow, smooth transition containing no glitches. The rate of change of output frequency when using the smooth switch circuit is less than 1 MHz/0.667 μs. The frequency overshoot and undershoot will be less than 2%.

The Smooth Switch circuit can be assigned to either PLL via register byte 14 bit 4. By default the smooth switch circuit is assigned to the CPU PLL. Either PLL can still be overclocked when it does not have control of the smooth switch circuit but it is not guaranteed to transition to the new frequency without large frequency glitches.

It is not recommended to enable overclocking and change the N values of both PLLs in the same SMBUS block write.

Watchdog Timer

The Watchdog timer is used in the system in conjunction with overclocking. It is used to provide a reset to a system that has hung up due to overclocking the CPU and the Front side bus. The Watchdog is enabled by the user and if the system completes its checkpoints, the system will clear the timer. However, when the timer runs out, there will be a reset pulse generated on the SRESET# pin for 20 ms that is used to reset the system.

When the Watchdog is enabled (WD_EN = 1) the Watchdog timer will start counting down from a value of Watchdog_timer * time scale. If the Watchdog timer reaches 0 before the WD_EN bit is cleared then it will assert the SRESET# signal and set the Watchdog Alarm bit to 1.

To use the Watchdog the SRESET# pin must be enabled by SRESET_EN pin being sampled LOW by VTPWRGD# assertion during system boot-up.

At any point during the Watchdog timer countdown, if the time stamp or Watchdog timer bits are changed, the timer will reset and start counting down from the new value.

After the Reset pulse, the Watchdog will stay inactive until either:

1. A new time stamp or Watchdog timer value is loaded.
2. The WD_EN bit is cleared and then set again.

Watchdog Register Bits

The following register bits are associated with the Watchdog timer:

Watchdog Enable – This bit (by default) is not set, which disables the Watchdog. When set, the Watchdog is enabled. Also, when there is a transition from LOW to HIGH, the timer reloads. Default = 0, disable

Watchdog Timer – There are three bits (for seven combinations) to select the timer value. Default = 000. The Value '000' is a reserved test mode.

Watchdog Alarm – This bit is a flag and when it is set, it indicates that the timer has expired. This bit is not set by default. When the bit is set, the user is allowed to clear. Default = 0.

Watchdog Time Scale – This bit selects the multiplier. When this bit is not set, the multiplier will be 250 ms. When set (by default), the multiplier will be 3s. Default = 1.

Watchdog Reset Mode – This selects the Watchdog reset mode. When this bit is not set (by default), the Watchdog will send a reset pulse and reload the recovery frequency depends on Watchdog Recovery Mode setting. When set, it just send a reset pulse. Default = 0, Reset & Recover Frequency.

Watchdog Recovery Mode – This bit selects the location to recover from. One option is to recover from the HW settings (already stored in SMBUS registers for readback capability) and the second is to recover from a register called "Recovery N". Default = 0 (Recover from the HW setting).

Watchdog Autorecovery Enable – This bit is set by default and the recovered values are automatically written into the "Watchdog Recovery Register" and reloaded by the watchdog function. When this bit is not set, the user is allowed to write to the "Watchdog Recovery Register". The value stored in the "Watchdog Recovery Register" will be used for recovery. Default = 1, Autorecovery.

Watchdog Recovery Register – This is a nine-bit register to store the Watchdog N recovery value. This value can be written by the Auto recovery or User depending on the state of the "Watchdog Auto recovery Enable bit".

Watchdog Recovery Modes

There are three operating modes that require Watchdog recovery. The modes are Dial-A-Frequency (DAF), Dynamic Clocking (DF), or Frequency Select. There are 4 different recovery modes: The following section lists the operating mode and the recovery mode associated with it.

Recover to Hardware M,N, O

When this recovery mode is selected, in the event of a watchdog timeout, the original M, N, and O values that were latched by the HW FSEL pins at chip boot-up should be reloaded.

Autorecovery

When this recovery mode is selected, in the event of a Watchdog timeout, the M and N values stored in the Recovery M and N registers should be reloaded. The current values of M and N will be latched into the internal recovery M and N registers by the WD_EN bit being set.

Manual Recovery

When this recovery mode is selected, in the event of a Watchdog timeout, the N value as programmed by the user in the N recovery register, and the M value that is stored in the Recovery M register (not accessible by the user) should be restored. The current M value should be latched M recovery register by the WD_EN bit being set.

No Recovery

If no recovery mode is selected, in the event of a watchdog time out, the device should just assert the SRESET# and keep the current values of M and N

Software Reset

Software reset is a reset function which is used to send out a pulse from the SRESET# pin. It is controlled by the SW_RESET enable register bit. Upon completion of the byte/word/block write in which the SW_RESET bit was set, the device will send a RESET pulse on the SRESET# pin. The duration of the SRESET# pulse should be the same as the duration of the SRESET# pulse after a Watchdog timer time out.

After the SRESET# pulse is asserted the SW_RESET bit should be automatically cleared by the device.

PD (Power-down) Clarification

The VTT_PWRGD#/PD pin is a dual-function pin. During initial power-up, the pin functions as VTT_PWRGD#. Once VTT_PWRGD# has been sampled LOW by the clock chip, the pin assumes PD functionality. The PD pin is an asynchronous active HIGH input used to shut off all clocks cleanly prior to shutting off power to the device. This signal is synchronized internal to the device prior to powering down the clock synthesizer. PD is also an asynchronous input for powering up the system. When PD is asserted HIGH, all clocks need to be driven to a low value and held prior to turning off the VCOs and the crystal oscillator.

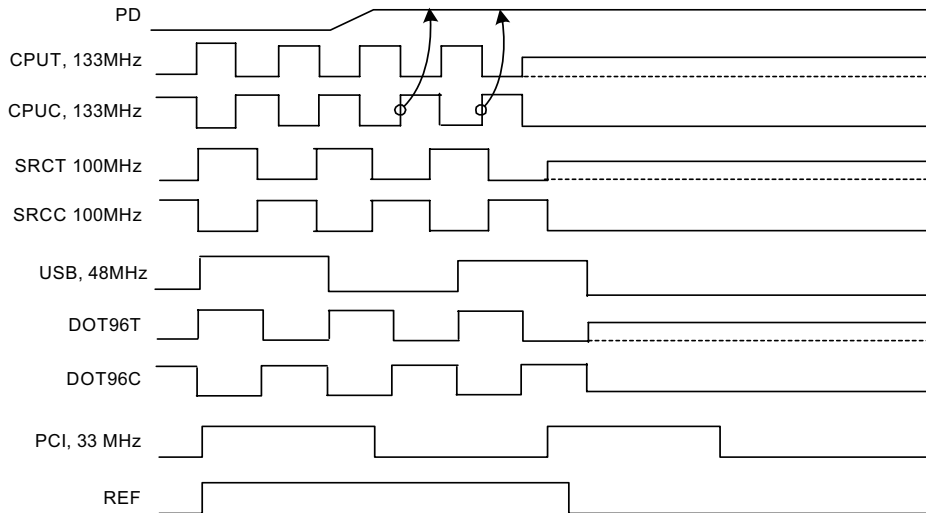
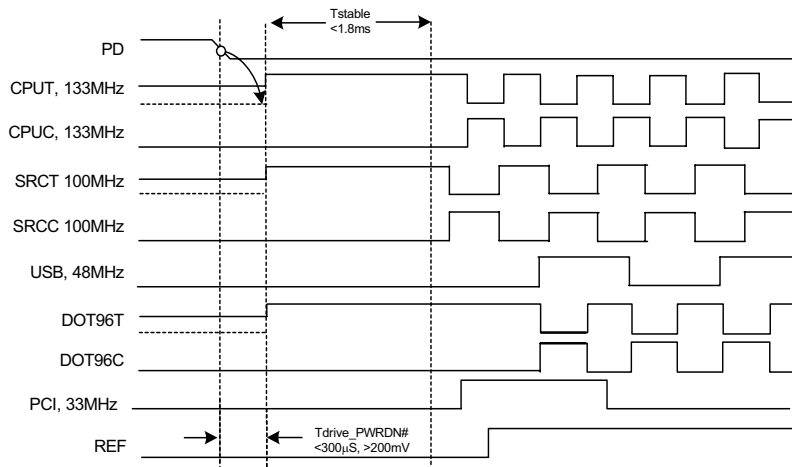
PD (Power-down) Assertion

When PD is sampled HIGH by two consecutive rising edges of CPUC, all single-ended outputs will be held LOW on their next HIGH-to-LOW transition and differential clocks must be held HIGH or tri-stated (depending on the state of the control register drive mode bit) on the next diff clock# HIGH-to-LOW transition within four clock periods. When the SMBus PD drive mode bit corresponding to the differential (CPU, SRC, and DOT) clock output of interest is programmed to '0', the clock

output are held with "Diff clock" pin driven high at $2 \times I_{ref}$, and "Diff clock#" tri-state. If the control register PD drive mode bit corresponding to the output of interest is programmed to "1", then both the "Diff clock" and the "Diff clock#" are tri-state. Note the example below shows CPU_T = 133 MHz and PD drive mode = '1' for all differential outputs. This diagram and description is applicable to valid CPU frequencies 100, 133, 166, 200, 266, 333, and 400 MHz. In the event that PD mode is desired as the initial power-on state, PD must be asserted high in less than 10 μ s after asserting Vtt_PwrGd#.

PD Deassertion

The power-up latency is less than 1.8 ms. This is the time from the deassertion of the PD pin or the ramping of the power supply until the time that stable clocks are output from the clock chip. All differential outputs stopped in a three-state condition resulting from power down will be driven high in less than 300 μ s of PD deassertion to a voltage greater than 200 mV. After the clock chip's internal PLL is powered up and locked, all outputs will be enabled within a few clock cycles of each other. *Figure 5* is an example showing the relationship of clocks coming up.


Figure 4. Power-down Assertion Timing Waveform

Figure 5. Power-down Deassertion Timing Waveform

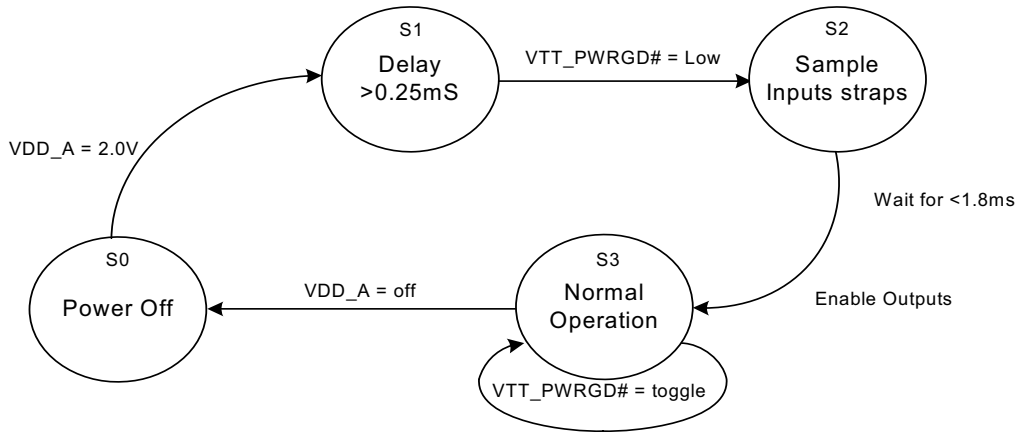


Figure 6. Clock Generator Power-up/Run State Diagram

Absolute Maximum Conditions

Parameter	Description	Condition	Min.	Max.	Unit
V _{DD}	Core Supply Voltage		-0.5	4.6	V
V _{DD_A}	Analog Supply Voltage		-0.5	4.6	V
V _{IN}	Input Voltage	Relative to V _{SS}	-0.5	V _{DD} + 0.5	VDC
T _S	Temperature, Storage	Non-functional	-65	150	°C
T _A	Temperature, Operating Ambient	Functional	0	70	°C
T _J	Temperature, Junction	Functional	-	150	°C
∅ _{JC}	Dissipation, Junction to Case	Mil-STD-883E Method 1012.1	-	20	°C/W
∅ _{JA}	Dissipation, Junction to Ambient	JEDEC (JESD 51)	-	60	°C/W
ESD _{HBM}	ESD Protection (Human Body Model)	MIL-STD-883, Method 3015	2000	-	V
UL-94	Flammability Rating	At 1/8 in.	V-0		
MSL	Moisture Sensitivity Level		1		

Multiple Supplies: The Voltage on any input or I/O pin cannot exceed the power pin during power-up. Power supply sequencing is NOT required.

DC Electrical Specifications

Parameter	Description	Condition	Min.	Max.	Unit
All V _{DDs}	3.3V Operating Voltage	3.3 ± 5%	3.135	3.465	V
V _{IL12C}	Input Low Voltage	SDATA, SCLK	-	1.0	V
V _{IH12C}	Input High Voltage	SDATA, SCLK	2.2	-	V
V _{IL_FS}	FS_[A:B,D:E] Input Low Voltage		V _{SS} - 0.3	0.35	V
V _{IH_FS}	FS_[A:B,D:E] Input High Voltage		0.7	V _{DD} + 0.5	V
V _{ILFS_C}	FS_C Low Range		0	0.35	V
V _{IMFS_C}	FS_C Mid Range		0.7	1.7	V
V _{IH_FS_C}	FS_C High Range		2.1	V _{DD}	V
V _{IL}	3.3V Input Low Voltage		V _{SS} - 0.3	0.8	V
V _{IH}	3.3V Input High Voltage		2.0	V _{DD} + 0.3	V
I _{IL}	Input Low Leakage Current	Except internal pull-up resistors, 0 < V _{IN} < V _{DD}	-5	-	μA
I _{IH}	Input High Leakage Current	Except internal pull-down resistors, 0 < V _{IN} < V _{DD}	-	5	μA
V _{OL}	3.3V Output Low Voltage	I _{OL} = 1 mA	-	0.4	V
V _{OH}	3.3V Output High Voltage	I _{OH} = -1 mA	2.4	-	V
I _{OZ}	High-impedance Output Current		-10	10	μA
C _{IN}	Input Pin Capacitance		3	5	pF
C _{OUT}	Output Pin Capacitance		3	5	pF
L _{IN}	Pin Inductance		-	7	nH
V _{XIH}	Xin High Voltage		0.7V _{DD}	V _{DD}	V
V _{XIL}	Xin Low Voltage		0	0.3V _{DD}	V
I _{DD3.3V}	Dynamic Supply Current	At max. load and freq. per <i>Figure 9</i>	-	500	mA
I _{PD3.3V}	Power-down Supply Current	PD asserted, Outputs Driven	-	70	mA
I _{PT3.3V}	Power-down Supply Current	PD asserted, Outputs Tri-state	-	2	mA

AC Electrical Specifications

Parameter	Description	Condition	Min.	Max.	Unit
Crystal					
T _{DC}	XIN Duty Cycle	The device will operate reliably with input duty cycles up to 30/70 but the REF clock duty cycle will not be within specification	47.5	52.5	%
T _{PERIOD}	XIN Period	When XIN is driven from an external clock source	69.841	71.0	ns
T _R / T _F	XIN Rise and Fall Times	Measured between 0.3V _{DD} and 0.7V _{DD}	–	10.0	ns
T _{CCJ}	XIN Cycle to Cycle Jitter	As an average over 1- μ s duration	–	500	ps
L _{ACC}	Long-term Accuracy	Over 150 ms	–	300	ppm
CPU at 0.7V (SSC refers to –0.5% spread spectrum)					
T _{DC}	CPUT and CPUC Duty Cycle	Measured at crossing point V _{OX}	45	55	%
T _{PERIOD}	100-MHz CPUT and CPUC Period	Measured at crossing point V _{OX}	9.997001	10.00300	ns
T _{PERIOD}	133-MHz CPUT and CPUC Period	Measured at crossing point V _{OX}	7.497751	7.502251	ns
T _{PERIOD}	166-MHz CPUT and CPUC Period	Measured at crossing point V _{OX}	5.998201	6.001801	ns
T _{PERIOD}	200-MHz CPUT and CPUC Period	Measured at crossing point V _{OX}	4.998500	5.001500	ns
T _{PERIOD}	266-MHz CPUT and CPUC Period	Measured at crossing point V _{OX}	3.748875	3.751125	ns
T _{PERIOD}	333-MHz CPUT and CPUC Period	Measured at crossing point V _{OX}	2.999100	3.000900	ns
T _{PERIOD}	400-MHz CPUT and CPUC Period	Measured at crossing point V _{OX}	2.499250	2.500750	ns
T _{PERIODSS}	100-MHz CPUT and CPUC Period, SSC	Measured at crossing point V _{OX}	9.997001	10.05327	ns
T _{PERIODSS}	133-MHz CPUT and CPUC Period, SSC	Measured at crossing point V _{OX}	7.497751	7.539950	ns
T _{PERIODSS}	166-MHz CPUT and CPUC Period, SSC	Measured at crossing point V _{OX}	5.998201	6.031960	ns
T _{PERIODSS}	200-MHz CPUT and CPUC Period, SSC	Measured at crossing point V _{OX}	4.998500	5.026634	ns
T _{PERIODSS}	266-MHz CPUT and CPUC Period, SSC	Measured at crossing point V _{OX}	3.748875	3.769975	ns
T _{PERIODSS}	333-MHz CPUT and CPUC Period, SSC	Measured at crossing point V _{OX}	2.999100	3.015980	ns
T _{PERIODSS}	400-MHz CPUT and CPUC Period, SSC	Measured at crossing point V _{OX}	2.499250	2.513317	ns
T _{PERIODAbs}	100-MHz CPUT and CPUC Absolute period	Measured at crossing point V _{OX}	9.912001	10.08800	ns
T _{PERIODAbs}	133-MHz CPUT and CPUC Absolute period	Measured at crossing point V _{OX}	7.412751	7.587251	ns
T _{PERIODAbs}	166-MHz CPUT and CPUC Absolute period	Measured at crossing point V _{OX}	5.913201	6.086801	ns
T _{PERIODAbs}	200-MHz CPUT and CPUC Absolute period	Measured at crossing point V _{OX}	4.913500	5.086500	ns
T _{PERIODAbs}	266-MHz CPUT and CPUC Absolute period	Measured at crossing point V _{OX}	3.663875	3.836125	ns
T _{PERIODAbs}	333-MHz CPUT and CPUC Absolute period	Measured at crossing point V _{OX}	2.914100	3.085900	ns
T _{PERIODAbs}	400-MHz CPUT and CPUC Absolute period	Measured at crossing point V _{OX}	2.414250	2.585750	ns
T _{PERIODSSAbs}	100-MHz CPUT and CPUC Absolute period, SSC	Measured at crossing point V _{OX}	9.912001	10.13827	ns
T _{PERIODSSAbs}	133-MHz CPUT and CPUC Absolute period, SSC	Measured at crossing point V _{OX}	7.412751	7.624950	ns
T _{PERIODSSAbs}	166-MHz CPUT and CPUC Absolute period, SSC	Measured at crossing point V _{OX}	5.913201	6.116960	ns
T _{PERIODSSAbs}	200-MHz CPUT and CPUC Absolute period, SSC	Measured at crossing point V _{OX}	4.913500	5.111634	ns
T _{PERIODSSAbs}	266-MHz CPUT and CPU C Absolute period, SSC	Measured at crossing point V _{OX}	3.663875	3.854975	ns
T _{PERIODSSAbs}	333-MHz CPUT and CPUC Absolute period, SSC	Measured at crossing point V _{OX}	2.914100	3.100980	ns
T _{PERIODSSAbs}	400-MHz CPUT and CPUC Absolute period, SSC	Measured at crossing point V _{OX}	2.414250	2.598317	ns
T _{SKEW}	CPU0 to CPU1	Measured at crossing point V _{OX}	–	100	ps

AC Electrical Specifications (continued)

Parameter	Description	Condition	Min.	Max.	Unit
T _{CCJ}	CPUT/C Cycle to Cycle Jitter	Measured at crossing point V _{OX}	–	65	ps
L _{ACC}	Long Term accuracy	Measured using frequency counter over 0.15 seconds.	–	300	ppm
T _R / T _F	CPUT and CPUC Rise and Fall Times	Measured from V _{OL} = 0.175 to V _{OH} = 0.525V	130	700	ps
T _{RFM}	Rise/Fall Matching	Determined as a fraction of 2*(T _R – T _F)/(T _R + T _F)	–	29	%
ΔT _R	Rise Time Variation		–	125	ps
ΔT _F	Fall Time Variation		–	125	ps
V _{HIGH}	Voltage High	Math averages <i>Figure 9</i>	660	850	mV
V _{LOW}	Voltage Low	Math averages <i>Figure 9</i>	–150	–	mV
V _{OX}	Crossing Point Voltage at 0.7V Swing		250	550	mV
V _{OVS}	Maximum Overshoot Voltage		–	V _{HIGH} + 0.3	V
V _{UDS}	Minimum Undershoot Voltage		–0.3	–	V
V _{RB}	Ring Back Voltage	See <i>Figure 9</i> . Measure SE	–	0.2	V
SRC					
T _{DC}	SRCT and SRCC Duty Cycle	Measured at crossing point V _{OX}	45	55	%
T _{PERIOD}	100-MHz SRCT and SRCC Period	Measured at crossing point V _{OX}	9.997001	10.00300	ns
T _{PERIODSS}	100-MHz SRCT and SRCC Period, SSC	Measured at crossing point V _{OX}	9.997001	10.05327	ns
T _{PERIODAbs}	100-MHz SRCT and SRCC Absolute Period	Measured at crossing point V _{OX}	9.872001	10.12800	ns
T _{PERIODSSAbs}	100-MHz SRCT and SRCC Absolute Period, SSC	Measured at crossing point V _{OX}	9.872001	10.17827	ns
T _{SKEW}	Any SRCT/C to SRCT/C Clock Skew	Measured at crossing point V _{OX}	–	250	ps
T _{CCJ}	SRCT/C Cycle to Cycle Jitter	Measured at crossing point V _{OX}	–	125	ps
L _{ACC}	SRCT/C Long Term Accuracy	Measured at crossing point V _{OX}	–	300	ppm
T _R / T _F	SRCT and SRCC Rise and Fall Times	Measured from V _{OL} = 0.175 to V _{OH} = 0.525V	130	700	ps
T _{RFM}	Rise/Fall Matching	Determined as a fraction of 2*(T _R – T _F)/(T _R + T _F)	–	20	%
ΔT _R	Rise Time Variation		–	125	ps
ΔT _F	Fall Time Variation		–	125	ps
V _{HIGH}	Voltage High	Math averages <i>Figure 9</i>	660	850	mV
V _{LOW}	Voltage Low	Math averages <i>Figure 9</i>	–150	–	mV
V _{OX}	Crossing Point Voltage at 0.7V Swing		250	550	mV
V _{OVS}	Maximum Overshoot Voltage		–	V _{HIGH} + 0.3	V
V _{UDS}	Minimum Undershoot Voltage		–0.3	–	V
V _{RB}	Ring Back Voltage	See <i>Figure 9</i> . Measure SE	–	0.2	V
PCI/PCIF					
T _{DC}	PCI Duty Cycle	Measurement at 1.5V	45	55	%
T _{PERIOD}	Spread Disabled PCIF/PCI Period	Measurement at 1.5V	29.99100	30.00900	ns
T _{PERIODSS}	Spread Enabled PCIF/PCI Period, SSC	Measurement at 1.5V	29.9910	30.15980	ns
T _{PERIODAbs}	Spread Disabled PCIF/PCI Period	Measurement at 1.5V	29.49100	30.50900	ns
T _{PERIODSSAbs}	Spread Enabled PCIF/PCI Period, SSC	Measurement at 1.5V	29.49100	30.65980	ns
T _{HIGH}	PCIF and PCI high time	Measurement at 2.4V	12.0	–	ns
T _{LOW}	PCIF and PCI low time	Measurement at 0.4V	12.0	–	ns

AC Electrical Specifications (continued)

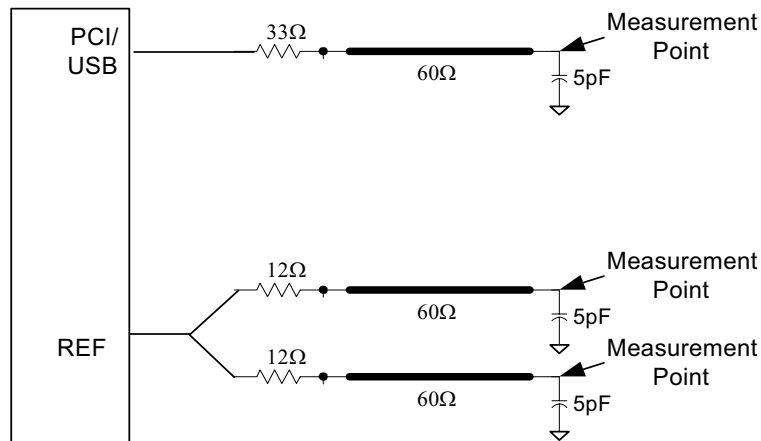
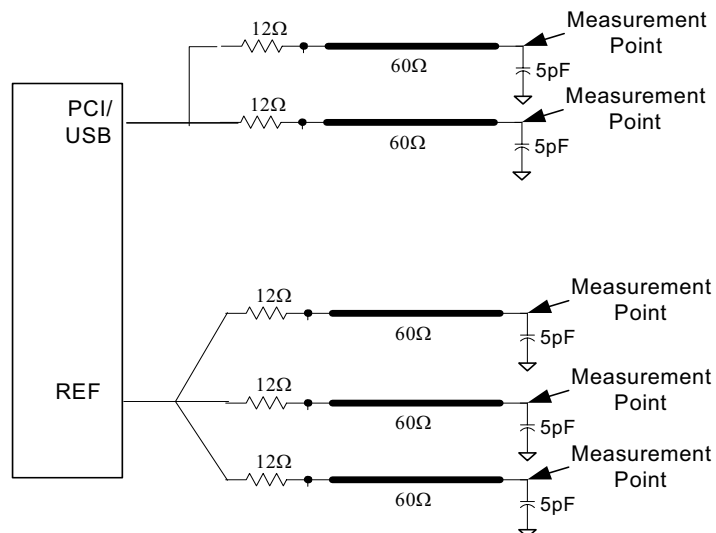
Parameter	Description	Condition	Min.	Max.	Unit
Edge Rate	Rising edge rate	Measured between 0.8V and 2.0V	0.85	6.0	V/ns
Edge Rate	Falling edge rate	Measured between 0.8V and 2.0V	0.85	6.0	V/ns
T _{SKEW}	Any PCI clock to Any PCI clock Skew	Measurement at 1.5V	–	500	ps
T _{CCJ}	PCIF and PCI Cycle to Cycle Jitter	Measurement at 1.5V	–	550	ps
DOT					
T _{DC}	DOT96T and DOT96C Duty Cycle	Measured at crossing point V _{OX}	45	55	%
T _{PERIOD}	DOT96T and DOT96C Period	Measured at crossing point V _{OX}	10.41354	10.41979	ns
T _{PERIODAbs}	DOT96T and DOT96C Absolute Period	Measured at crossing point V _{OX}	10.16354	10.66979	ns
T _{CCJ}	DOT96T/C Cycle to Cycle Jitter	Measured at crossing point V _{OX}	–	250	ps
L _{ACC}	DOT96T/C Long Term Accuracy	Measured at crossing point V _{OX}	–	100	ppm
T _{LTJ}	Long Term jitter	Measurement taken from cross point V _{OX} @1 μs	–	700	ps
		Measurement taken from cross point V _{OX} @10 μs	–	700	ps
T _R / T _F	DOT96T and DOT96C Rise and Fall Times	Measured from V _{OL} = 0.175 to V _{OH} = 0.525V	130	700	ps
T _{RFM}	Rise/Fall Matching	Determined as a fraction of *(T _R – T _F)/(T _R + T _F)	–	20	%
ΔT _R	Rise Time Variation		–	125	ps
ΔT _F	Fall Time Variation		–	125	ps
V _{HIGH}	Voltage High	Math averages <i>Figure 9</i>	660	850	mV
V _{LOW}	Voltage Low	Math averages <i>Figure 9</i>	–150	–	mV
V _{OX}	Crossing Point Voltage at 0.7V Swing		250	550	mV
V _{OVS}	Maximum Overshoot Voltage		–	V _{HIGH} + 0.3	V
V _{UDS}	Minimum Undershoot Voltage		–0.3	–	V
V _{RB}	Ring Back Voltage	See <i>Figure 9</i> . Measure SE	–	0.2	V
USB					
T _{DC}	Duty Cycle	Measurement at 1.5V In High Drive mode	45	55	%
T _{PERIOD}	Period	Measurement at 1.5V	20.83125	20.83542	ns
T _{PERIODAbs}	Absolute Period	Measurement at 1.5V	20.48125	21.18542	ns
T _{HIGH}	USB high time	Measurement at 2.4V	8.094	10.5	ns
T _{LOW}	USB low time	Measurement at 0.4V	7.694	10.5	ns
Edge Rate	Rising edge rate	Measured between 0.8V and 2.0V	0.52	2.4	V/ns
Edge Rate	Falling edge rate	Measured between 0.8V and 2.0V	0.52	2.4	V/ns
T _{CCJ}	Cycle to Cycle Jitter	Measurement at 1.5V	–	200	ps
T _{LTJ}	Long Term jitter	Measurement taken from cross point V _{OX} @1 μs	–	700	ps
		Measurement taken from cross point V _{OX} @10 μs	–	700	ps
		Measurement taken from cross point V _{OX} @125 μs	–	700	ps
REF					
T _{DC}	REF Duty Cycle	Measurement at 1.5V	45	55	%
T _{PERIOD}	REF Period	Measurement at 1.5V	69.8203	69.8622	ns
T _{PERIODAbs}	REF Absolute Period	Measurement at 1.5V	68.82033	70.86224	ns

AC Electrical Specifications (continued)

Parameter	Description	Condition	Min.	Max.	Unit
T_R / T_F	REF Rise and Fall Times	Measured between 0.8V and 2.0V	0.5	2.0	V/ns
Edge Rate	Rising edge rate	Measured between 0.8V and 2.0V	1.0	4.0	V/ns
Edge Rate	Falling edge rate	Measured between 0.8V and 2.0V	1.0	4.0	V/ns
T_{CCJ}	REF Cycle to Cycle Jitter	Measurement at 1.5V	–	1000	ps
ENABLE/DISABLE and SET-UP					
T_{STABLE}	Clock Stabilization from Power-up		–	1.8	ms

Test and Measurement Set-up
For PCI Single-ended Signals and Reference

The following diagram shows the test load configurations for the single-ended PCI, USB, and REF output signals.


Figure 7. Single-ended Load Configuration

Figure 8. Single-ended Load Configuration HIGH DRIVE OPTION

For Differential CPU, SRC and DOT96 Output Signals

The following diagram shows the test load configuration for the differential CPU and SRC outputs.

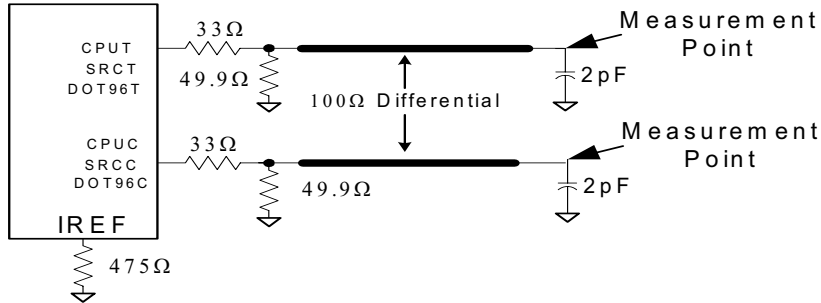


Figure 9. 0.7V Single-ended Load Configuration

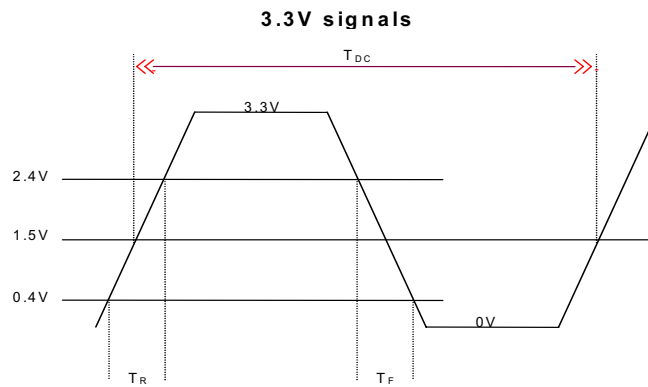
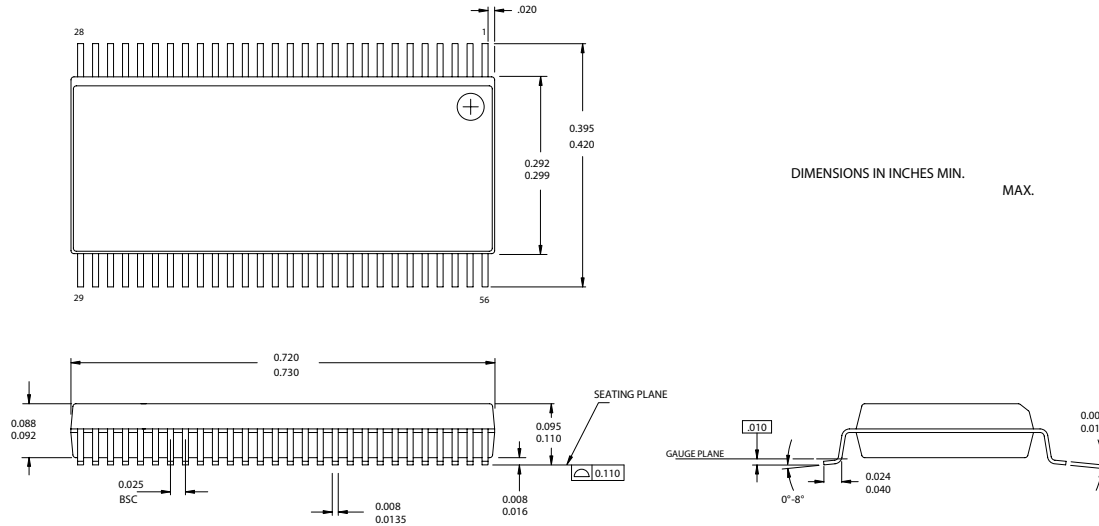
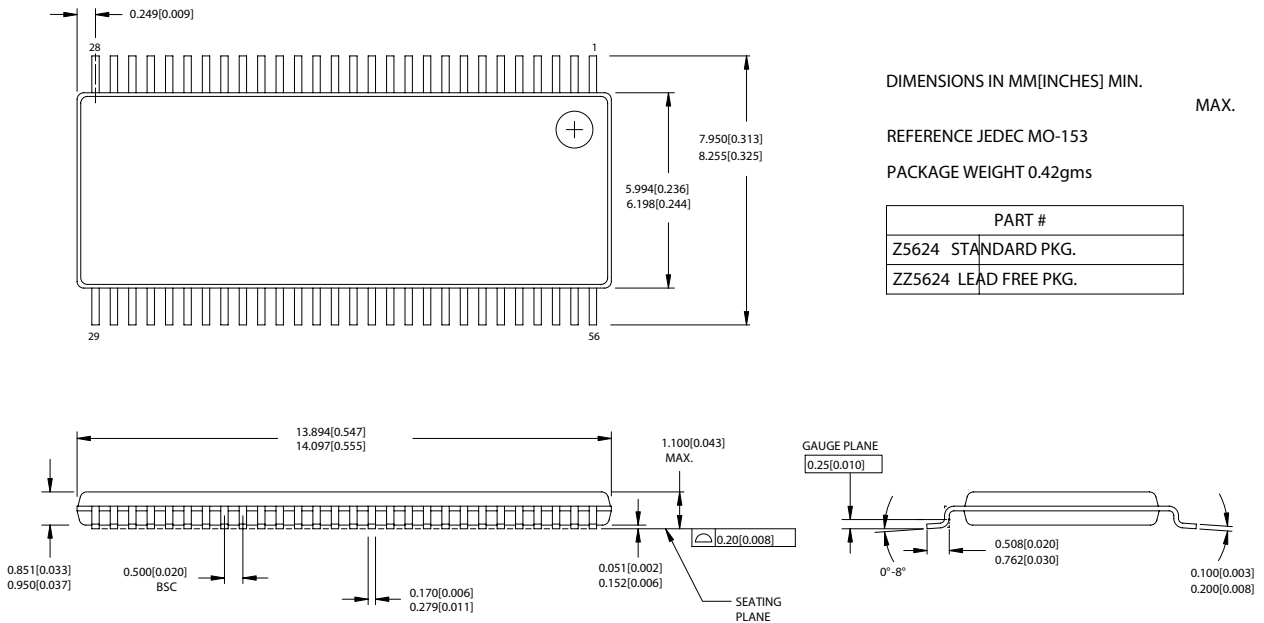


Figure 10. Single-ended Output Signals (for AC Parameters Measurement)

Ordering Information

Part Number	Package Type	Product Flow
Lead-free		
CY28437OXC	56-pin SSOP	Commercial, 0° to 85°C
CY28437OXCT	56-pin SSOP – Tape and Reel	Commercial, 0° to 85°C
CY28437ZXC	56-pin TSSOP	Commercial, 0° to 85°C
CY28437ZXCT	56-pin TSSOP – Tape and Reel	Commercial, 0° to 85°C

Package Diagrams
56-Lead Shrunk Small Outline Package O56

56-Lead Thin Shrunk Small Outline Package, Type II (6 mm x 12 mm) Z56


While SLI has reviewed all information herein for accuracy and reliability, Spectra Linear Inc. assumes no responsibility for the use of any circuitry or for the infringement of any patents or other rights of third parties which would result from each use. This product is intended for use in normal commercial applications and is not warranted nor is it intended for use in life support, critical medical instruments, or any other application requiring extended temperature range, high reliability, or any other extraordinary environmental requirements unless pursuant to additional processing by Spectra Linear Inc., and expressed written agreement by Spectra Linear Inc. Spectra Linear Inc. reserves the right to change any circuitry or specification without notice.