

# Adjustable Precision Shunt Regulator

## FEATURES

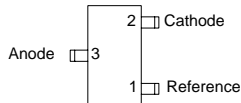
- Low voltage operation (2.5V)
- Adjustable output voltage from  $V_O = V_{REF}$  to 18V
- Wide operating current range from 0.4mA to 100mA
- Low dynamic output impedance 0.5Ω max.
- ESD rating is 5.5kV (per MIL-STD 883D)
- Pb-free lead finish available.**

## APPLICATIONS

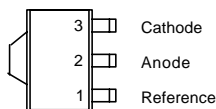
- Linear Regulators
- Adjustable Supplies
- Switching Power Supplies
- Battery Operated Computers
- Instrumentation
- Computer Disk Drives

## PIN CONFIGURATION

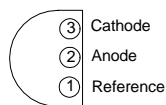
SOT-23 (Top view)



SOT-89 (Top view)



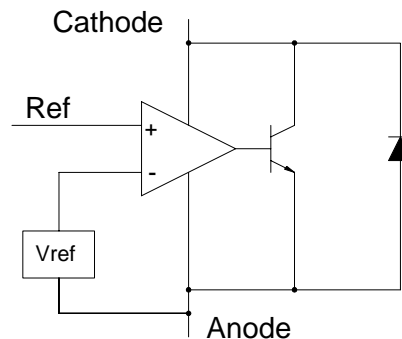
TO-92 (Top view)



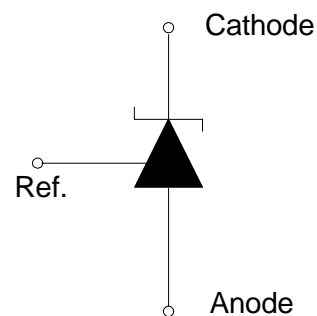
## DESCRIPTION

The SS431 is a low-voltage three-terminal adjustable shunt regulator with a guaranteed thermal stability over the applicable temperature range. The output voltage can be set to any value between  $V_{REF}$  (approximately 2.5V) to 18V with two external resistors (see application circuit). This device has a typical output impedance of 0.2 ohms. Active output circuitry provides very sharp turn-on characteristics, making this device an excellent replacement for Zener diodes in many applications. The SS431 is characterized for operation from 0°C to 105°C, and three package options (SOT-23, SOT-89, and TO-92) allow the designer the opportunity to select the proper package for his application. **The SS431 is available with a Pb-free lead finish - order as SS431Gxx.**

## BLOCK DIAGRAM



## SYMBOL



**ABSOLUTE MAXIMUM RATINGS over ambient temp. range.**

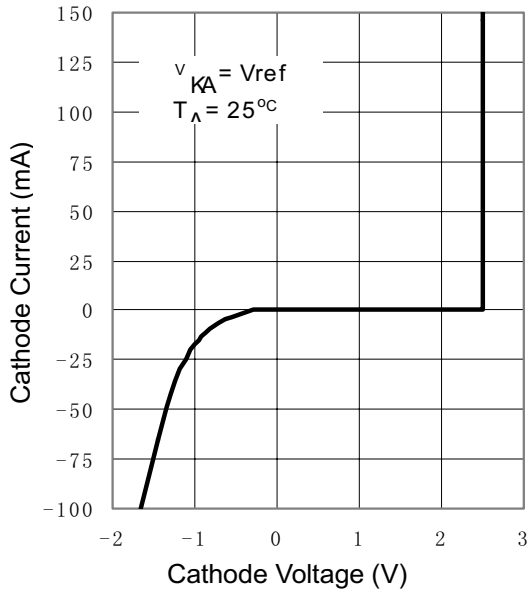
Parameter	Symbol	Maximum	Units
Cathode Voltage	$V_{KA}$	18	V
Continuous Cathode Current	$I_{KA}$	150	mA
Reference Current	$I_{REF}$	10	mA
Operating Junction Temperature	$T_J$	125	°C
Storage Temperature Range	$T_{STG}$	-65 to 150	°C
Thermal Resistance	$\theta_{JA}$	156	°C/W
Lead Temperature (Soldering) 10 seconds	$T_{LEAD}$	260	°C

**ELECTRICAL CHARACTERISTICS**
 **$T_A = 25^\circ\text{C}$** 

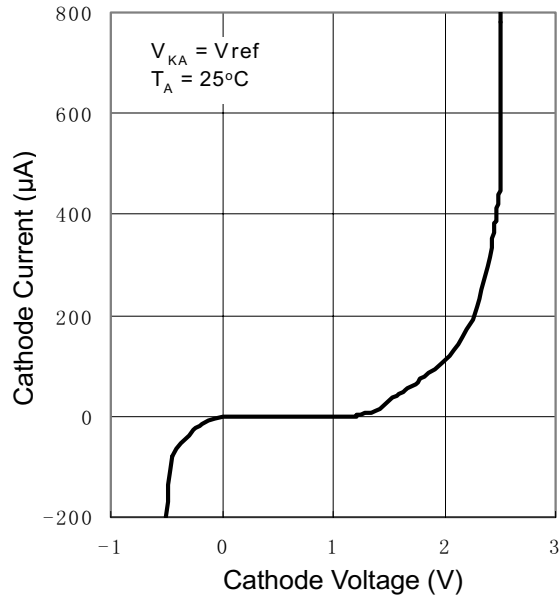
PARAMETER		TEST CIRCUIT	TEST CONDITIONS				UNIT	
				MIN	TYP	MAX		
Reference voltage	1%	$V_{ref}$	1	$V_{KA} = V_{ref}$ $I_{KA} = 10\text{mA}$	2470	2495	2520	mV
Deviation of reference voltage Over full temperature range		$V_{I(dev)}$	1	$V_{KA} = V_{ref}$ , $I_{KA} = 10\text{mA}$ $T_A = \text{full range}$		4	25	mV
Ratio of change in reference voltage to the change in cathode voltage		$\frac{\Delta V_{ref}}{\Delta V_{KA}}$	2	$I_{KA} = 10\text{mA}$ $\Delta V_{KA} = 10\text{V} - V_{ref}$		-1.4	-2.7	mV/V
Reference current		$I_{ref}$	2	$I_{KA} = 10\text{mA}$ , $R1=10k\Omega$ , $R2 = \infty$		2	4	$\mu\text{A}$
Deviation of Reference current over full temperature range		$I_{I(dev)}$	2	$I_{KA} = 10\text{mA}$ , $R1=10k\Omega$ , $R2 = \infty$ $T_A = \text{full range}$		0.4	1.2	$\mu\text{A}$
Minimum cathode current for regulation		$I_{min}$	1	$V_{KA} = V_{ref}$		0.4	1	mA
Off-state cathode current		$I_{off}$	3	$V_{KA} = 18\text{V}$ , $V_{ref} = 0$		0.1	1	$\mu\text{A}$
Dynamic impedance		$r_z$	1	$I_{KA} = 1\text{mA}$ to $100\text{mA}$ , $V_{KA} = V_{ref}$ $f \leq 1\text{kHz}$		0.2	0.5	$\Omega$

**TYPICAL PERFORMANCE CHARACTERISTICS**

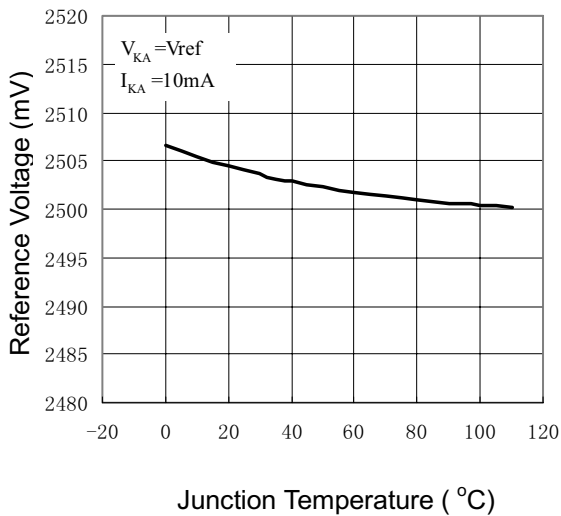
CATHODE CURRENT  
Vs.  
CATHODE VOLTAGE



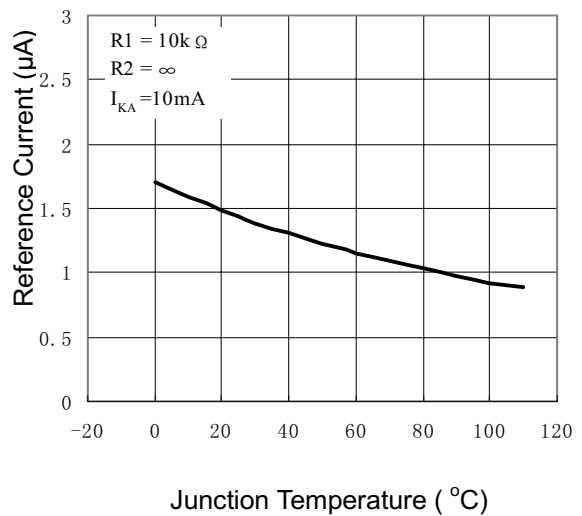
CATHODE CURRENT  
Vs.  
CATHODE VOLTAGE

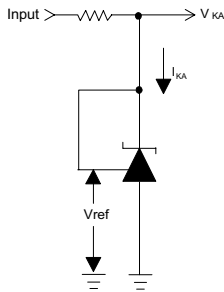


REFERENCE VOLTAGE  
Vs.  
JUNCTION TEMPERATURE

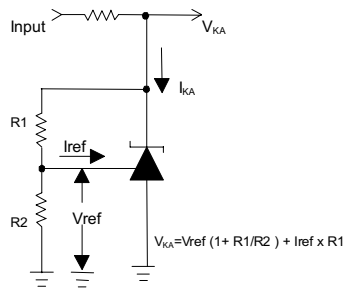


REFERENCE INPUT CURRENT  
Vs.  
JUNCTION TEMPERATURE

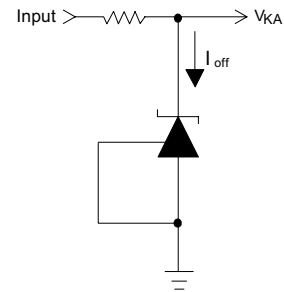


**TEST CIRCUITS**


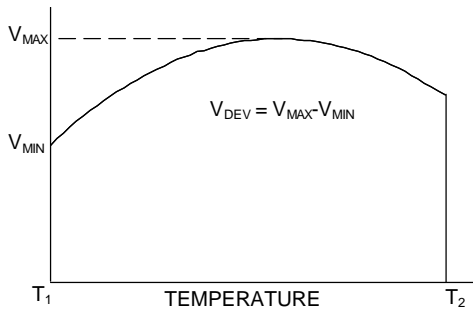
**Test Circuit 1:**  
 $V_{KA} = V_{ref}$



**Test Circuit 2:**  
 $V_{KA} > V_{ref}$



**Test Circuit 3:**  
**Off State Current**

**APPLICATION INFORMATION**


Deviation of reference input voltage,  $V_{DEV}$ , is defined as the maximum variation of the reference input voltage over the full temperature range.

The average temperature coefficient of the reference input voltage,  $\alpha V_{REF}$  is defined as:

$$\Delta V_{REF} \frac{\text{ppm}}{^{\circ}\text{C}} = \frac{\pm \left[ \frac{V_{MAX} - V_{MIN}}{V_{REF}(\text{at } 25^{\circ}\text{C})} \right] 10^6}{T_2 - T_1} = \frac{\pm \left[ \frac{V_{DEV}}{V_{REF}(\text{at } 25^{\circ}\text{C})} \right] 10^6}{T_2 - T_1}$$

Where:

$T_2 - T_1$  = full temperature change.

$\alpha V_{REF}$  can be positive or negative depending on whether the slope is positive or negative.

Example:  $V_{DEV} = 7.0\text{mV}$ ,  $V_{REF} = 2495\text{mV}$ ,

$T_2 - T_1 = 105^{\circ}\text{C}$ , slope is negative.

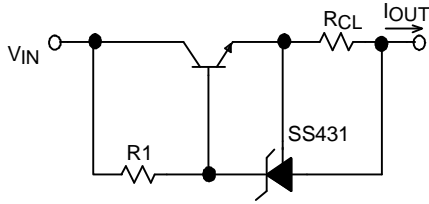
$$aV_{REF} = \frac{\left[ \frac{7.0\text{mV}}{2495\text{mV}} \right] 10^6}{105^{\circ}\text{C}} = -27\text{ppm}/^{\circ}\text{C}$$

**Note 4.** The dynamic output impedance,  $R_Z$ , is defined as:

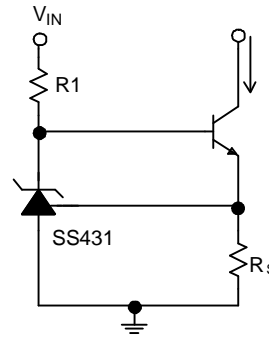
$$R_Z = \frac{\Delta V_Z}{\Delta I_Z}$$

When the device is programmed with two external resistors,  $R_1$  and  $R_2$ , (see Fig. 2), the dynamic output impedance of the overall circuit, is defined as:

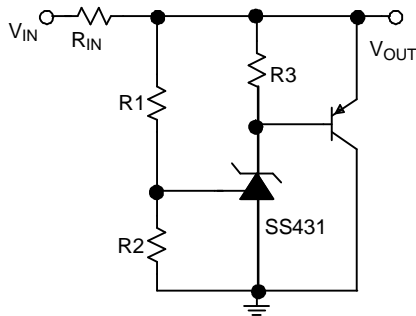
$$r_Z = \frac{\Delta V}{\Delta I} \cong R_Z \left[ 1 + \frac{R_1}{R_2} \right]$$

**APPLICATION EXAMPLES**


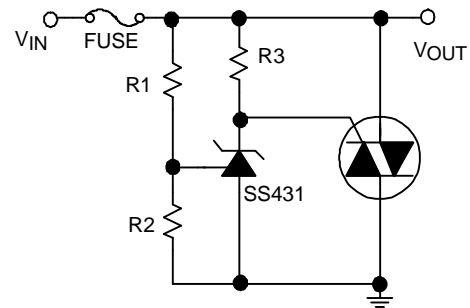
$$I_{OUT} = V_{REF} / R_{CL}$$

**Current Limiter or Current Source**


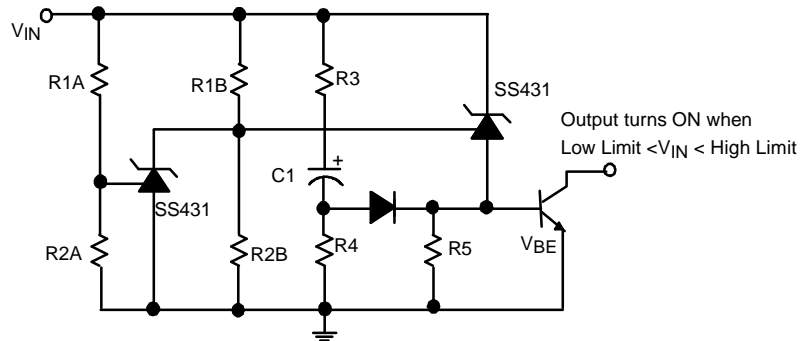
$$I_{OUT} = V_{REF} / R_S$$

**Constant-Current Sink**


$$V_{OUT} \cong (1 + R_1/R_2) \times V_{REF}$$

**Higher-Current Shunt Regulator**


$$V_{LIMIT} \cong (1 + R_1/R_2) \times V_{REF}$$

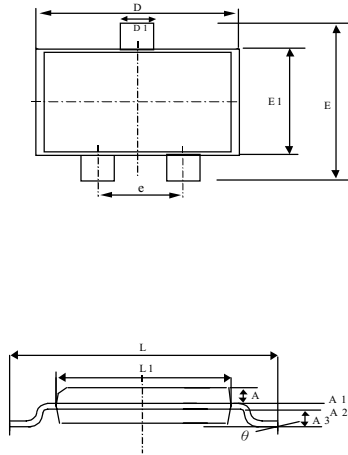
**Crow Bar**


$$\text{Low Limit} \cong V_{REF} (1 + R_{1B} / R_{2B}) + V_{BE}$$

$$\text{High Limit} \cong V_{REF} (1 + R_{1A} / R_{2A})$$

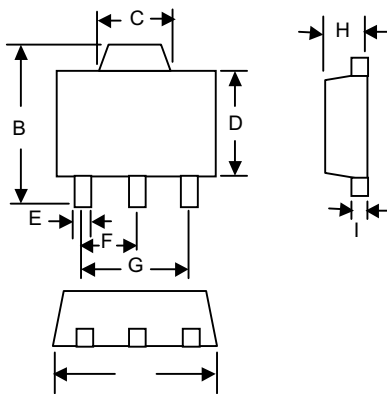
**Over-Voltage/Under-Voltage Protection Circuit**

Output turns ON when  
Low Limit < VIN < High Limit

**PHYSICAL DIMENSIONS SOT-23**


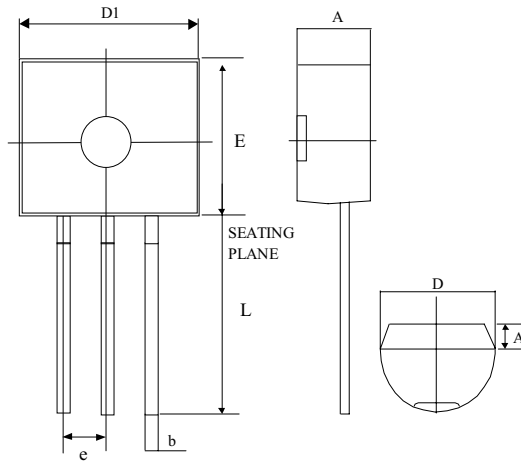
SYMBOL	MIN	NOM	MAX
A	0.20	0.30	0.40
A1	—	0.00	—
A2	—	0.10	—
A3	0.70	0.80	0.90
D1	0.30	0.40	0.50
e	1.70	2.00	2.30
D	2.80	2.90	3.00
E	2.25	2.50	2.75
E1	1.40	1.50	1.60
L	2.25	2.50	2.75
L1	1.40	1.50	1.60
$\theta$	—	2°	—

Unit: :mm

**SOT-89**


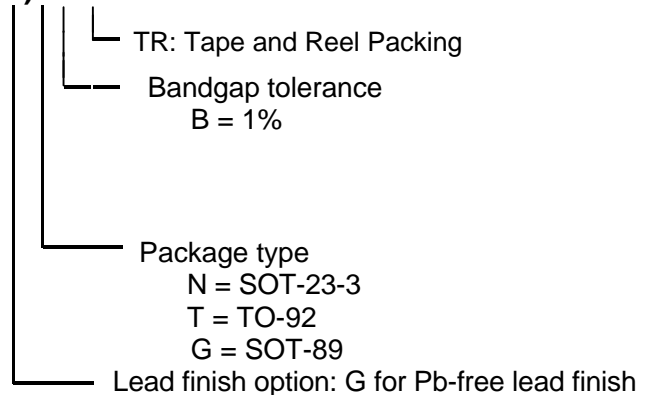
SYMBOL	MIN	MAX
A	4.40	4.60
B	4.05	4.25
C	1.50	1.70
D	2.40	2.60
E	0.31	0.46
F	1.48	1.52
G	2.96	3.04
H	1.40	1.60
I	0.35	0.41

Unit: :mm

**PHYSICAL DIMENSIONS TO-92**


SYMBOL	MIN	NOM	MAX
A	3.45	3.56	3.66
A1	1.22	1.3	1.37
b	-	0.38	-
D1	4.27	4.52	4.78
D	4.14	4.29	4.45
E	4.32	4.57	4.83
e	-	1.27	-
L	12.98	13.49	14.00

Unit: mm

**ORDERING INFORMATION**
**SS431(G)xBxx**


Example: SS431GNBTR

SS431 with 1% tolerance in SOT-23-3 with pb-free lead finish shipped on tape and reel

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