2067556 CHERRY SEMICONDUCTOR CORP

85D 00770

DT-58-11-31



CS-1524 CS-2524 CS-3524 CS-1524/2524/3524

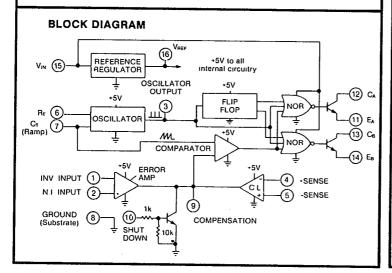
PULSE WIDTH MODULATOR CONTROL CIRCUIT DUAL OUTPUT FOR SINGLE-ENDED OR PUSH-PULL APPLICATIONS

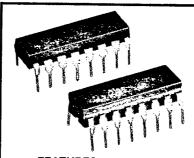
DESCRIPTION

The CS-1524, CS-2524 and CS-3524 incorporate all the functions required for the control of regulating power supplies, inverters or switching regulators. They can also be used as the control element for high-power-output applications. The CS-1524 family was designed for switching regulators of either polarity, transformer-coupled dc-to-dc converters, transformerless voltage doublers and polarity converter applications employing fixed-frequency, pulse-width modulation techniques. The dual alternating outputs allow either single-ended or push-pull applications. Each device includes an on-chip reference, error amplifier, programmable oscillator, pulse-steering flip-flop, two uncommitted output transistors, a high-gain comparator, and current-limiting and shut-down circuitry. The CS-1524 is characterized for operation over the full military temperature range of -55° C to +125° C. The CS-2524 and CS-3524 are designed for operation from -25° C to +85° C and 0° C to +70° C respectively.

RECOMMENDED OPERATING CONDITIONS

Supply Voltage, Vcc	8V to 40V
Reference Output Current	0 to 20mA
Current through C _T Terminal	0.03mA to -2mA
Timing Resistor, R ₁	1.8KΩ to 100KΩ
Timing Capacitor, CT	0.001 µF to 0.1 µF

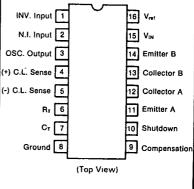




FEATURES:

- Complete PWM Control Circuit
- Single Ended or Push-Pull Outputs
- Low Standby Current (8mA Typical)
- 5V ± 4% Reference
- 1% Temperature Stability, V_{ref}
- Operation to 300 KHz
- Current Limiting

PIN CONNECTIONS



89

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MIN. TYP. MAX. MIN. TYP. MAX. MAX. MIN. TYP. MAX. MAX. MIN. TYP. MAX. MIN. TYP. MAX. MIN. TYP. MAX. MAX. MIN. TYP. MAX. MIN. TYP. MAX. MAX. MIN. TYP. MAX. MAX. MIN. TYP. MAX. MAX. MAX. MAX. MIN. TYP. MAX.	Reference Output Current		ting Jur	perature inction Te	remperati	ture	• • • • • • • •	55	55 to +150
PARAMETER	LECTHICAL CHARACTERISTICS	(Unless otherwise stated, these specifica -25°C to +85°C for the CS2524 and 0°C	ations a to +70°	apply for °C for th	T _A = -55° ie CS352	°C to +1 4, V _{IN} = 1	25° C fo 20V, and	or the CS d f = 20k!	i1524, Hz)
MIN. TYP. MAX. TYP. TYP. MAX. TYP.	PARAMETER		CS	1524/CS	S2524	Ţ <u></u>	CS3524	24	UNITS
Output Voltage 4.8 5.0 5.2 4.6 5.0 5.4 V Line Regulation V _{Nx} = 8 to 40V 10 20 10 30 mV Load Regulation I _L = 0 to 20mA 20 50 20 50 mV Ripple Rejection I = 120H₂, T₂ = 25°C 66 6 66 de de Short Circuit Current Limit V _{Nx} = 0, T₂ = 25°C 100 1 0.3 1 0.3 1 0.3 1 0.3 1 0.3 1 0.3 1 0.3 1 0.3 1 % Long Term Stability Var = 25° C, t = 1000 Hrs. 20 0 0 0 0 0 0 0 mv Oscillator Section Ta = 25° C, t = 1000 Hrs. 300 N 1 0 3 1 1 0 3 1 1 0 % Oscillator Section Ta = 25° C 5 1 1 0 0 5 </th <th>Reference Section</th> <th></th> <th>MIN.</th> <th>TYP.</th> <th>MAX.</th> <th>MIN.</th> <th>TYP.</th> <th>MAX.</th> <th></th>	Reference Section		MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	
Line Regulation		T	TAB	1 50	T 6 2	T 16	T.,	-,-	,
Load Regulation I ₁ = 0 to 20mA 20 50 20 50 50 50 50 50	· · · · · · · · · · · · · · · · · · ·	V _{IV} = 8 to 40V	14.0	+		4.0	+-		
Rippie Rejection	···		 	+	+				mV
Short Circuit Current Limit	 		 		30	 	+	50)	mV
Temperature Stability		<u> </u>		1	 			+	dB m^
Long Term Stability			 	+	+	+	+-	+	mA %
Casillator Section Casillator Casillator Section Casillator Section Casillator Casillator Section Casillator Casillator Casillator Casillator Casillator Casillator Casillator Casillator Ca					+-			+-	
Initial Accuracy	*** <u>***</u>	10 0,		<u> </u>	ىــــــــــــــــــــــــــــــــــــــ		20		mV
Initial Accuracy	Maximum Frequency	$C_{\tau} = .001 \mu F$, $R_{\tau} = 2k\Omega$		1 300	F		T 200		1-1.1
Voltage Stability				+	 		-}	+	kHz
Temperature Stability	······································			<u> </u>	+		+	++	
Output Amplitude Pin 3. T _A = 25°C 3.5 2 2 % Output Pulse Width C _T = .01 μF, T _A = 25°C 0.5 0.5 .05 μs Error Amplifier Section Input Offset Voltage V _{CM} = 2.5V 0.5 5 2 10 mV Input Offset Voltage V _{CM} = 2.5V 2 10 2 10 μA Open Loop Voltage Gain 72 80 60 80 dB Common Mode Voltage T _A = 25°C 1.8 3.4 1.8 3.4 V Common Mode Rejection Ratio T _A = 25°C 70 70 70 dB Small Signal Bandwidth A _V = 0dB, T _A = 25°C 3 3.8 0.5 3.8 V Comparator Section T _A = 25°C 0.5 3.8 0.5 3.8 V Duty-Cycle % Each Output On 0 45 0 45 % Input Threshold Zero Duty-Cycle 1 1 1 V I							 	+	<u> </u>
Output Pulse Width C _T = .01μF, T _A = 25°C 0.5 0.5 μs Error Amplifier Section Input Offset Voltage V _{CM} = 2.5V 0.5 5 2 10 mV Input Diffset Voltage V _{CM} = 2.5V 2 10 2 10 μA Open Loop Voltage Gain V _{CM} = 2.5V 2 10 2 10 μA Open Loop Voltage Gain T _A = 25°C 1.8 3.4 1.8 3.4 V Common Mode Voltage T _A = 25°C 70 70 70 dB Small Signal Bandwidth A _V = 0dB, T _A = 25°C 3 3.8 0.5 3.8 V Comparator Section Duty-Cycle % Each Output On 0 45 0 45 % Input Threshold Zero Duty-Cycle 1 1 1 V Input Threshold Maximum Duty-Cycle 3.5 3.5 V Input Threshold Maximum Duty-Cycle 3.5 3.5 V C				35			125	++	
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Input Bias Current V _{CM} = 2.5V 2 10 2 10 μA	Input Offset Voltage	V _{CM} = 2.5V		0.5	5		Γ,	1 +0]	1/
Copen Loop Voltage Gain 72 80 60 80 60 80 dB	Input Bias Current				+		-		
Common Mode Voltage TA = 25°C 1.8 3.4 1.8 3.4 V	Open Loop Voltage Gain		72	 	- "	60	+	10	
Common Mode Rejection Ratio T _A = 25°C 70 70 70 dB		T _A = 25° C			24		80	+ 34+	
Small Signal Bandwidth Av = 0dB, T _A = 25°C 3 3 MHz Output Voltage T _A = 25°C 0.5 3.8 0.5 3.8 V Comparator Section Duty-Cycle % Each Output On 0 45 0 45 % Input Threshold Zero Duty-Cycle 1 1 1 V Input Bias Current 3.5 3.5 3.5 V Current Limiting Section Sense Voltage Pin 9 = 2V with Error Amplifier Set for Maximum Out, T _A = 25°C 190 200 210 180 200 220 mV Sense Voltage Pin 9 = 2V with Error Amplifier Set for Maximum Out, T _A = 25°C 190 200 210 180 200 220 mV Sense Voltage Pin 9 = 2V with Error Amplifier Set for Maximum Out, T _A = 25°C 0.2 0.2 mV 0.1 0.2 mV Common Mode Voltage -1 +1 -1 +1 V V 0.1 0.2 mV 0.1 50 </td <td></td> <td>· </td> <td>, ''``</td> <td>70</td> <td></td> <td></td> <td>70</td> <td>3.4</td> <td></td>		·	, ''``	70			70	3.4	
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Duty-Cycle % Each Output On 0		1			3.0		<u> </u>	3.6	v
Input Threshold Zero Duty-Cycle 1	Duty-Cycle	% Each Output On	<u> </u>		45	,		T-45	04
Input Threshold	, , , , , , , , , , , , , , , , , , , 	† 	 	1	-40		 	45	
Input Bias Current		 						+	
Current Limiting Section Sense Voltage Pin 9 = 2V with Error Amplifier Set for Maximum Out, T _A = 25°C 190 200 210 180 200 220 mV Sense Voltage T.C. 0.2 0.2 0.2 mV/°C Common Mode Voltage -1 +1 -1 +1 V Collector Each Output) Collector-Emitter Voltage 40 40 V V Collector Leakage Current Voe = 40V 0.1 50 0.1 50 μA Saturation Voltage Ic = 50mA 1 2 1 2 V Emitter Output Voltage Vin = 20V 17 18 17 18 V Fall Time Rc = 2K ohm, T _A = 25°C 0.2 0.2 μs Fotal Standby Current Vin = 40V 8 10 8 10 8 10 0.1 0.1 0.1 0.1 0.1 0.1 0.1 0.1 0.1 0.1 0.1 0.1 0.1								-	
Pin 9 = 2V with Error Amplifier Set for Maximum Out, T _A = 25°C 190 200 210 180 200 220 mV	Current Limiting Section				—— L				μΑ
Sense Voltage T.C. 0.2 0.2 mV/° C Common Mode Voltage -1 +1 -1 +1 V Output Section (Each Output) Collector-Emitter Voltage 40 40 V Collector Leakage Current VoE = 40V 0.1 50 0.1 50 μA Saturation Voltage Ic = 50mA 1 2 1 2 V Emitter Output Voltage Vin = 20V 17 18 17 18 V Rise Time Rc = 2K ohm, TA = 25° C 0.2 0.2 μs Fall Time Rc = 2K ohm, TA = 25° C 0.1 0.1 μs Fotal Standby Current Vin = 40V 8 10 8 10 8 10 8 10 8 10 8 10 8 10 8 10 8 10 8 10 8 10 8 10 8 10 8 10 8 10 8 10 8 10 8 10 8 10 10 10 10	Sense Voltage	Pin 9 = 2V with Error Amplifier Set for Maximum Out, T _A = 25°C	190	200	210	180	200	220	mV
Common Mode Voltage -1 +1 -1 +1 V Output Section (Each Output) Collector-Emitter Voltage 40 40 V Collector Leakage Current V _{CE} = 40V 0.1 50 0.1 50 μA Saturation Voltage I _C = 50mA 1 2 1 2 V Emitter Output Voltage V _{IN} = 20V 17 18 17 18 V Rise Time R _C = 2K ohm, T _A = 25°C 0.2 0.2 μs Fall Time R _C = 2K ohm, T _A = 25°C 0.1 0.1 μs Fotal Standby Current V _{IN} = 40V 8 10 8 10 8 10 8 10 7 7	Sense Voltage T.C.			0.2	,——		0.2		mV/° C
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Saturation Voltage I _C = 50mA 1 2 1 2 V Emitter Output Voltage V _{IN} = 20V 17 18 17 18 V Rise Time R _C = 2K ohm, T _A = 25°C 0.2 0.2 μs Fall Time R _C = 2K ohm, T _A = 25°C 0.1 0.1 μs Total Standby Current V _{IN} = 40V 8 10 8 10 8 10 πA		V _{CE} = 40V		0.1	50		0.1	50	
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Rise Time R _C = 2K ohm, T _A = 25° C 0.2 0.2 μs Fall Time R _C = 2K ohm, T _A = 25° C 0.1 0.1 μs Fotal Standby Current V _{IN} = 40V 8 10 8 10 πο		V _{IN} = 20V	17			17	-		
Fall Time R _C = 2K ohm, T _A = 25°C 0.1 0.1 μs Total Standby Current V _{IN} = 40V 8 10 8 10 πΔ	· · · · · · · · · · · · · · · · · · ·	Rc = 2K ohm, T _A = 25°C				"		,	
Total Standby Current $V_{IN} = 40V$ 8 10 8 10 mA	all Time		1			-		-+	
(Excluding oscillator charging current, error and current limit	- · · · · · · · · · · · · · · · · · · ·	V _{IN} = 40V	_	-	10	-		10	μs mA

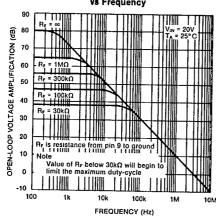
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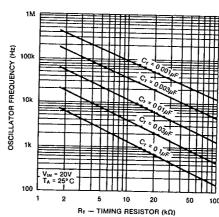
CS-1524 CS-2524 CS-3524

TYPICAL CHARACTERISTICS

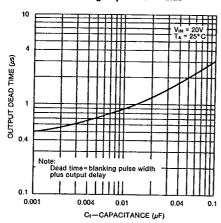
Open-Loop Voltage Amplification of Error Amplifier vs Frequency



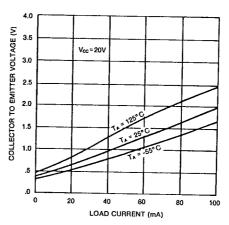
Oscillator Frequency vs Timing Components



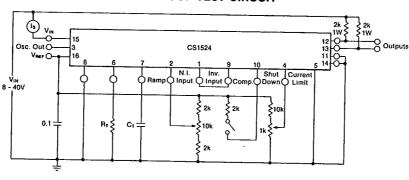
Output Dead Time vs Timing Capacitance Value



Output Saturation Voltage vs Load Current



OPEN LOOP TEST CIRCUIT



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CS-1524 CS-2524 CS-3524

PRINCIPLES OF OPERATION

The CS1524 is a fixed-frequency pulse-width-modulation voltage regulator control circuit. The regulator operates at a frequency that is programmed by one timing resistor (R_T) and one timing capacitor (C_T). Rr establishes a constant charging current for C_T. This results in a linear voltage ramp at C_T, which is fed to the comparator providing linear control of the output pulse width by the error amplifier. The CS1524 contains an on-board 5V regulator that serves as a reference as well as powering the CS1524's internal control circuitry and is also useful in supplying external support functions. This reference voltage is lowered externally by a resistor divider to provide a reference within the common-mode range of the error amplifier or an external reference may be used. The power supply output is sensed by a second resistor divider network to generate a feedback signal to the error amplifier. The amplifier output voltage is then compared to the linear voltage ramp at C_T.

The resulting modulated pulse out of the high-gain comparator is then steered to the appropriate output pass transistor $(Q_1\ or\ Q_2)$ by the pulse-steering flip-flop, which is synchronously toggled by the oscillator output. The oscillator output pulse also serves as a blanking pulse to assure both outputs are never on simultaneously during the transition times. The width of the blanking pulse is controlled by the value of C_T . The outputs may be applied in a pushpull configuration in which their frequency is half that of the base oscillator, or paralleled for single-ended applications in which the frequency is equal to the oscillator. The output of the error amplifier shares a common input to the comparator with the current limiting and shutdown circuitry and can be overridden by signals from either of these inputs. This common point is also available externally and may be employed to control the gain of, or to compensate, the error amplifier, or to provide additional control to the regulator.

TYPICAL APPLICATIONS DATA Oscillator

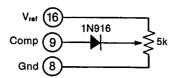
The oscillator controls the frequency of the CS1524 and is programmed by R_{T} and C_{T} according to the approximate formula:

 $\begin{array}{cc} \text{where} & R_T \text{ is in kilhoms} \\ C_T \text{ is in microfarads} \\ \text{f is in kilohertz} \end{array}$

Practical values of C_T fall between 0.001 and 0.1 microfarad. Practical values of R_T fall between 1.8 and 100 kilohms. This results in a frequency range typically from 120 hertz to 500 kilohertz.

Blanking

The output pulse of the oscillator is used as a blanking pulse at the output. This pulse width is controlled by the value of C_T . If small values of C_T are required for frequency control, the oscillator output pulse width may still be increased by applying a shunt capacitance of up to 100pF from pin 3 to ground. If still greater dead-time is required, it should be accomplished by limiting the maximum duty cycle by clamping the output of the error amplifier. This can easily be done with the circuit below:



Synchronous Operation

When an external clock is desired, a clock pulse of approximately 3V can be applied directly to the oscillator output terminal. The impedance to ground at this point is approximately 2 kilohms. In this configuration R_T C_T must be selected for a clock period slightly greater than that of the external clock.

If two or more CS1524 regulators are to be operated synchronously, all oscillator output terminals should be tied together, all C_T terminals connected to a single timing capacitor, and the timing resistor connected to a single R_T terminal. The other R_T terminals can be left open or shorted $V_{\text{REF}}.$ Minimum lead lengths should be used between the C_T terminals.

Current Limiting

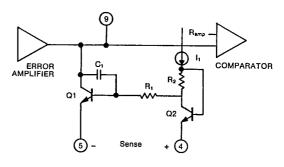
The current limiting circuitry of the CS1524 is shown below. By matching the base-emitter voltages of Q1 and Q2 and assuming negligible voltage drop across R_1 .

Threshold =
$$V_{BE}$$
 (Q1) + $I_1 R_2 - V_{BE}$ (Q2) = $I_1 R_2$
 $\approx 200 \text{ mV}$

Although this circuit provides a relatively small threshold with a negligible temperature coefficient, there are some limitations to its use, the most important of which is the ± 1 volt common mode range which requires sensing in the ground line. Another factor to consider is that the frequency compensation provided by R₁C₁ and Q1 provides a roll-off pole at approximately 300 Hertz.

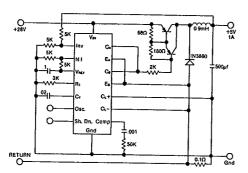
Since the gain of this circuit is relatively low, there is a transition region as the current limit amplifier takes over pulse width control from the error amplifier. For testing purposes, threshold is defined as the input voltage to get 25% duty cycle with the error amplifier signaling maximum duty cycle.

If this current limit circuitry is unused, pins 4 and 5 should both be grounded.



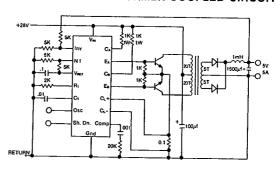
Current Limiting Circuitry of the CS1524

SINGLE-ENDED LC CIRCUIT



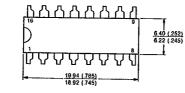
In this conventional single-ended regulator circuit, the two outputs of the CS1524 are connected in parallel for effective 0-90% duty-cycle modulation. The use of an output inductor requires an R-C phase compensation network for loop stability.

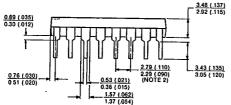
PUSH-PULL TRANSFORMER COUPLED CIRCUIT

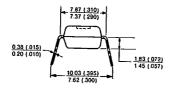


Push-pull outputs are used in this transformer-coupled DC-DC regulating converter. Note that the oscillator must be set at twice the desired output frequency as the CS1524's internal flip-flop divides the frequency by 2 as it switches the P.W.M. signal from one output to the other. Current limiting is done here in the primary so that the pulse width will be reduced should transformer saturation occur.

CS-2524, CS-3524 MECHANICAL SPECIFICATIONS: PLASTIC 16 PIN N PACKAGE



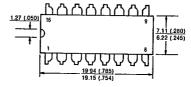


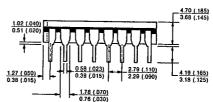


- NOTES:

 1. DIMENSIONS SHOWN ARE IN MILLIMETERS.
 THOSE IN PARENTHESES ARE IN INCHES.
 2. TOLERANCES ARE NON-ACCUMULATIVE.

CS-1524, CS-2524, CS-3524 MECHANICAL SPECIFICATIONS: CERAMIC (CERDIP) 16 PIN J PACKAGE







- NOTES:

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 THOSE IN PARENTHESES ARE IN INCHES.
 2. TOLERANCES ARE NON-ACCUMULATIVE.

2000 South County Trail, East Greenwich, Rhode Island 02818 (401) 885-3600 Telex WUI 6817157

Our Sales Representative in Your Area is: