



## VFM STEP-UP DC/DC CONVERTER

### RH5RI××1B/××2B/××3B SERIES

#### OUTLINE

The RH5RI××1B/××2B/××3B Series are VFM (Chopper) Step-up DC/DC converter ICs with ultra low supply current by CMOS process.

The RH5RI××1B IC consists of an oscillator, a VFM control circuit, a driver transistor (Lx switch), a reference voltage unit, an error amplifier, resistors for voltage detection, and an Lx switch protection circuit. A low ripple, high efficiency step-up DC/DC converter can be constructed of this RH5RI××1B IC with only three external components, that is, an inductor, a diode and a capacitor.

The RH5RI××2B IC uses the same chip as that employed in the RH5RI××1B IC and is provided with a drive pin (EXT) for an external transistor instead of an Lx pin, so that a power transistor with a low saturation voltage can be externally provided, whereby a large current can be caused to flow through the inductor and accordingly a large current can be obtained. Therefore, the RH5RI××2B IC is recommendable to the user who need a current as large as several tens mA to several hundreds mA.

The RH5RI××3B IC also includes an internal chip enable circuit so that it is possible to set the standby supply current at MAX. 0.5μA.

These RH5RI××1B/××2B/××3B ICs are suitable for use with battery-powered instruments with low noise and ultra low supply current.

#### FEATURES

- Small Number of External Components ..... Only an inductor, a diode and a capacitor (RH5RI××1B)
- Ultra Low Input Current ..... TYP. 4μA (RH5RI301B/303B at no load, with 1.5V input)
- High Output Voltage Accuracy ..... ±2.5%
- Low Ripple and Low Noise
- Low Start-up Voltage (When the output current is 1mA) ..... MAX. 0.9V
- High Efficiency ..... TYP. 80%
- Low Temperature-Drift Coefficient of Output Voltage ..... TYP. ±50 ppm/°C
- Small Packages ..... SOT-89 (RH5RI××1B, RH5RI××2B)  
SOT-89-5 (RH5RI××3B)

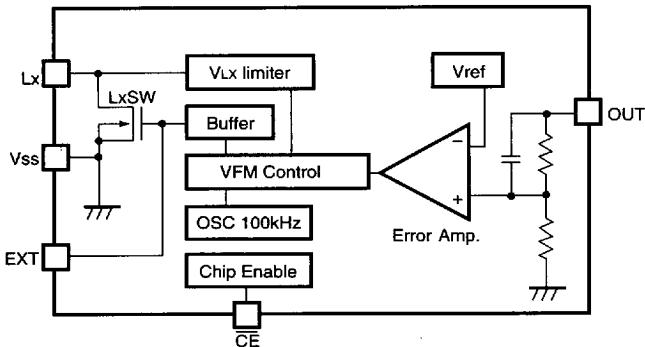
#### APPLICATIONS

- Power source for battery-powered equipment.
- Power source for cameras, camcorders, VCRs, PDAs, electronic data banks, and hand-held communication equipment.
- Power source for appliances which require higher cell voltage than that of batteries used in the appliances.

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## BLOCK DIAGRAM



(Note) Lx Pin.....only for RH5RIxx1B and RH5RIxx3B  
 EXT Pin.....only for RH5RIxx2B and RH5RIxx3B  
 CE Pin.....only for RH5RIxx3B

## SELECTION GUIDE

In RH5RI Series, the output voltage, the driver, and the taping type for the ICs can be selected at the user's request. The selection can be made by designating the part number as shown below :

RH5RIxxxx-xx ← Part Number  
 ↑   ↑   ↑  
 a   b   c

Code	Contents
a	Setting Output Voltage (V <sub>OUT</sub> ): Stepwise setting with a step of 0.1V in the range of 2.5V to 7.5V is possible.
b	Designation of Driver: 1B: Internal Lx Tr. Driver (Oscillator Frequency 100kHz) 2B: External Tr. Driver (Oscillator Frequency 100kHz) 3B: Internal Tr./External Tr. (selectively available) (Oscillator Frequency 100kHz, with chip enable function)
c	Designation of Taping type : Ex. SOT-89 : T1, T2 SOT-89-5 : T1, T2 (refer to Taping Specification) "T1" is prescribed as a standard.

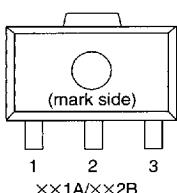
For example, the product with Output Voltage 5.0V, the External Driver (the Oscillator Frequency 100kHz) and Taping Type T1, is designated by Part Number RH5RI502B-T1.

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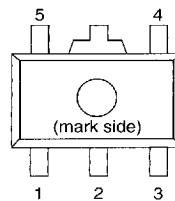
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## PIN CONFIGURATION

• SOT-89



• SOT-89-5



## PIN DESCRIPTION

Pin No.			Symbol	Description
xx1B	xx2B	xx3B		
1	1	5	Vss	Ground Pin
2	2	2	OUT	Step-up Output Pin, Power Supply (for device itself)
3	—	4	Lx	Switching Pin (Nch Open Drain)
—	3	3	EXT	External Tr. Drive Pin (CMOS Output)
—	—	1	CE	Chip Enable Pin (Active Low)

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**ABSOLUTE MAXIMUM RATINGS**V<sub>SS</sub>=0V

Symbol	Item	Rating	Unit	Note
V <sub>OUT</sub>	Output Pin Voltage	+12	V	
V <sub>LX</sub>	Lx Pin Voltage	+12	V	Note1
V <sub>EXT</sub>	EXT Pin Voltage	-0.3 to V <sub>OUT</sub> +0.3	V	Note2
V <sub>CE</sub>	C <sub>E</sub> Pin Voltage	-0.3 to V <sub>OUT</sub> +0.3	V	Note3
I <sub>LX</sub>	Lx Pin Output Current	250	mA	Note1
I <sub>EXT</sub>	EXT Pin Current	±50	mA	Note2
P <sub>D</sub>	Power Dissipation	500	mW	
T <sub>opt</sub>	Operating Temperature Range	-30 to +80	°C	
T <sub>stg</sub>	Storage Temperature Range	-55 to +125	°C	
T <sub>solder</sub>	Lead Temperature (Soldering)	260°C,10s		

(Note 1) Applicable to RH5RI××1A and RH5RI××3B.

(Note 2) Applicable to RH5RI××2B and RH5RI××3B.

(Note 3) Applicable to RH5RI××3B.

**ABSOLUTE MAXIMUM RATINGS**

Absolute Maximum ratings are threshold limit values that must not be exceeded even for an instant under any conditions. Moreover, such values for any two items must not be reached simultaneously. Operation above these absolute maximum ratings may cause degradation or permanent damage to the device. These are stress ratings only and do not necessarily imply functional operation below these limits.

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## ELECTRICAL CHARACTERISTICS

### • RH5RI301B

VOUT=3.0V

Symbol	Item	Conditions	MIN.	TYP.	MAX.	Unit	Note
VOUT	Output Voltage		2.925	3.000	3.075	V	
VIN	Input Voltage				8	V	
Vstart	Start-up Voltage	IOUT=1mA, VIN:0→2V		0.8	0.9	V	
Vhold	Hold-on Voltage	IOUT=1mA, VIN:2→0V	0.7			V	
IIN1	Input Current 1	To be measured at VIN at no load		4	8	µA	
IIN2	Input Current 2	To be measured at VIN VIN=3.5V		2	5	µA	
ILX	Lx Switching Current	VLX=0.4V	60			mA	
ILXleak	Lx Leakage Current	VLX=6V, VIN=3.5V			0.5	µA	
fosc	Maximum Oscillator Frequency		80	100	120	kHz	
Maxdty	Oscillator Duty Cycle	on (VLX "L" ) side	65	75	85	%	
η	Efficiency		70	80		%	
VLXlim	VLX Voltage Limit	Lx Switch On	0.65	0.8	1.0	V	Note

Unless otherwise provided, VIN=1.8V, Vss=0V, IOUT=10mA, T<sub>opt</sub>=25°C, and use External Circuit of Typical Application (FIG. 1).

(Note) ILx is gradually increased by the external inductor after Lx Switch is turned ON. In accordance with the increase of ILx, VLx is also increased. When VLx reaches VLXlim, Lx Switch is turned OFF by Lx Switch Protection Circuit. The time period from the time at which VLx reaches VLXlim to the time at which Lx Switch is turned OFF is about 3µs.

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## • RH5RI501B

Vout=5.0V

Symbol	Item	Conditions	MIN.	TYP.	MAX.	Unit	Note
VOUT	Output Voltage		4.875	5.000	5.125	V	
VIN	Input Voltage				8	V	
Vstart	Start-up Voltage	IOUT=1mA, VIN:0→2V		0.8	0.9	V	
Vhold	Hold-on Voltage	IOUT=1mA, VIN:2→0V	0.7			V	
IIN1	Input Current 1	To be measured at VIN at no load		6	12	µA	
IIN2	Input Current 2	To be measured at VIN VIN=5.5V		2	5	µA	
ILX	Lx Switching Current	VLX=0.4V	80			mA	
ILXleak	Lx Leakage Current	VLX=6V, VIN=5.5V			0.5	µA	
fosc	Maximum Oscillator Frequency		80	100	120	kHz	
Maxdty	Oscillator Duty Cycle	on (VLX "L" ) side	65	75	85	%	
η	Efficiency		70	80		%	
VLXlim	VLX Voltage Limit	Lx Switch On	0.65	0.8	1.0	V	Note2

Unless otherwise provided, VIN=3V, Vss=0V, IOUT=10mA, Topt=25°C, and use External Circuit of Typical Application (FIG. 1).

(Note) ILx is gradually increased by the external inductor after Lx Switch is turned ON. In accordance with the increase of ILx, VLx is also increased. When VLx reaches VLxlim, Lx Switch is turned OFF by Lx Switch Protection Circuit. The time period from the time at which VLx reaches VLxlim to the time at which Lx Switch is turned OFF is about 3µs.

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## • RH5RI302B

VOUT=3.0V

Symbol	Item	Conditions	MIN.	TYP.	MAX.	Unit	Note
VOUT	Output Voltage		2.925	3.000	3.075	V	
VIN	Input Voltage				8	V	
Vstart	Oscillator Start-up Voltage	EXT at no load,VOUT:0→2V		0.7	0.8	V	
IDD1	Supply Current 1	EXT at no load,VOUT=2.88V		30	50	μA	
IDD2	Supply Current 2	EXT at no load,VOUT=3.5V		2	5	μA	
IEXTH	EXT "H" Output Current	VEXT=VOUT-0.4V	-1.5			mA	
IEXTL	EXT "L" Output Current	VEXT=0.4V	1.5			mA	
fosc	Maximum Oscillator Frequency		80	100	120	kHz	
Maxdty	Oscillator Duty Cycle	VEXT "H" side	65	75	85	%	

Unless otherwise provided, VIN=1.8V, Vss=0V, IOUT=10mA, Topt=25°C, and use External Circuit of Typical Application (FIG. 2).

## • RH5RI502B

VOUT=5.0V

Symbol	Item	Conditions	MIN.	TYP.	MAX.	Unit	Note
VOUT	Output Voltage		4.875	5.000	5.125	V	
VIN	Input Voltage				8	V	
Vstart	Oscillator Start-up Voltage	EXT at no load,VOUT:0→2V		0.7	0.8	V	
IDD1	Supply Current 1	EXT at no load,VOUT=4.8V		60	90	μA	
IDD2	Supply Current 2	EXT at no load,VOUT=5.5V		2	5	μA	
IEXTH	EXT "H" Output Current	VEXT=VOUT-0.4V	-2			mA	
IEXTL	EXT "L" Output Current	VEXT=0.4V	2			mA	
fosc	Maximum Oscillator Frequency		80	100	120	kHz	
Maxdty	Oscillator Duty Cycle	VEXT "H" side	65	75	85	%	

Unless otherwise provided, VIN=3V, Vss=0V, IOUT=10mA, Topt=25°C, and use External Circuit of Typical Application (FIG. 2).

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## • RH5RI303B

V<sub>OUT</sub>=3.0V

Symbol	Item	Conditions	MIN.	TYP.	MAX.	Unit	Note
V <sub>OUT</sub>	Output Voltage		2.925	3.000	3.075	V	
V <sub>IN</sub>	Input Voltage				8	V	
V <sub>start</sub>	Start-up Voltage	I <sub>OUT</sub> =1mA, V <sub>IN</sub> :0→2V		0.8	0.9	V	
V <sub>hold</sub>	Hold-on Voltage	I <sub>OUT</sub> =1mA, V <sub>IN</sub> :2→0V	0.7			V	
η	Efficiency		70	80		%	
I <sub>IN1</sub>	Input Current 1	To be measured at V <sub>IN</sub> at no load		4	8	μA	
I <sub>IN2</sub>	Input Current 2	To be measured at V <sub>IN</sub> V <sub>IN</sub> =3.5V		2	5	μA	
I <sub>LX</sub>	Lx Switching Current	V <sub>LX</sub> =0.4V	60			mA	
I <sub>LXleak</sub>	Lx Leakage Current	V <sub>LX</sub> =6V, V <sub>IN</sub> =3.5V			0.5	μA	
I <sub>EXTH</sub>	EXT "H" Output Current	V <sub>EXT</sub> =V <sub>OUT</sub> -0.4V	-1.5			mA	
I <sub>EXTL</sub>	EXT "L" Output Current	V <sub>EXT</sub> =0.4V	1.5			mA	
V <sub>CEH1</sub>	CE "H" Level 1	V <sub>OUT</sub> ≥1.5V	V <sub>OUT</sub> -04			V	
V <sub>CEL1</sub>	CE "L" Level 1	V <sub>OUT</sub> ≥1.5V			0.4	V	
V <sub>C EH2</sub>	CE "H" Level 2	0.8V≤V <sub>OUT</sub> <1.5V	V <sub>OUT</sub> -0.1			V	
V <sub>CEL2</sub>	CE "L" Level 2	0.8V≤V <sub>OUT</sub> <1.5V			0.1	V	
I <sub>CEH</sub>	CE "H" Input Current	CE=3V			0.5	μA	
I <sub>CEL</sub>	CE "L" Input Current	CE=0V	-0.5			μA	
fosc	Maximum Oscillator Frequency		80	100	120	kHz	
Maxdty	Oscillator Duty Cycle	on (V <sub>LX</sub> "L" )side	65	75	85	%	
V <sub>LXlim</sub>	V <sub>LX</sub> Voltage Limit	Lx Switch on	0.65	0.8	1.0	V	Note

Unless otherwise provided, V<sub>IN</sub>=1.8V, V<sub>SS</sub>=0V, I<sub>OUT</sub>=10mA, T<sub>opt</sub>=25°C, and use External Circuit of Typical Application (FIG. 3).

(Note) I<sub>LX</sub> is gradually increased by the external inductor after Lx Switch is turned ON. In accordance with the increase of I<sub>LX</sub>, V<sub>LX</sub> is also increased. When V<sub>LX</sub> reaches V<sub>LXlim</sub>, Lx Switch is turned OFF by Lx Switch Protection Circuit. The time period from the time at which V<sub>LX</sub> reaches V<sub>LXlim</sub> to the time at which Lx Switch is turned OFF is about 3μs.

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## • RH5RH503B

V<sub>OUT</sub>=5.0V

Symbol	Item	Conditions	MIN.	TYP.	MAX.	Unit	Note
V <sub>OUT</sub>	Output Voltage		4.875	5.000	5.125	V	
V <sub>IN</sub>	Input Voltage				8	V	
V <sub>start</sub>	Start-up Voltage	I <sub>OUT</sub> =1mA, V <sub>IN</sub> :0→2V		0.8	0.9	V	
V <sub>hold</sub>	Hold-on Voltage	I <sub>OUT</sub> =1mA, V <sub>IN</sub> :2→0V	0.7			V	
η	Efficiency		70	85		%	
I <sub>IN1</sub>	Input Current 1	To be measured at V <sub>IN</sub> at no load		6	12	μA	
I <sub>IN2</sub>	Input Current 2	To be measured at V <sub>IN</sub> V <sub>IN</sub> =5.5V		2	5	μA	
I <sub>LX</sub>	Lx Switching Current	V <sub>LX</sub> =0.4V	80			mA	
I <sub>LXleak</sub>	Lx Leakage Current	V <sub>LX</sub> =6V, V <sub>IN</sub> =5.5V			0.5	μA	
I <sub>EXTH</sub>	EXT "H" Output Current	V <sub>EXT</sub> =V <sub>OUT</sub> -0.4V	-2.0			mA	
I <sub>EXTL</sub>	EXT "L" Output Current	V <sub>EXT</sub> =0.4V	2.0			mA	
V <sub>CEH1</sub>	CE "H" Level 1	V <sub>OUT</sub> ≥1.5V	V <sub>OUT</sub> -0.4			V	
V <sub>CEL1</sub>	CE "L" Level 1	V <sub>OUT</sub> ≥1.5V			0.4	V	
V <sub>CEH2</sub>	CE "H" Level 2	0.8V≤V <sub>OUT</sub> <1.5V	V <sub>OUT</sub> -0.1			V	
V <sub>CEL2</sub>	CE "L" Level 2	0.8V≤V <sub>OUT</sub> <1.5V			0.1	V	
I <sub>CEH</sub>	CE "H" Input Current	CE=5V			0.5	μA	
I <sub>CEL</sub>	CE "L" Input Current	CE=0V	-0.5			μA	
f <sub>osc</sub>	Maximum Oscillator Frequency		80	100	120	kHz	
Maxdty	Oscillator Duty Cycle	on (V <sub>LX</sub> "L" )side	65	75	85	%	
V <sub>LXlim</sub>	V <sub>LX</sub> Voltage Limit	Lx Switch on	0.65	0.8	1.0	V	Note

Unless otherwise provided, V<sub>IN</sub>=3V, V<sub>SS</sub>=0V, I<sub>OUT</sub>=10mA, T<sub>opt</sub>=25°C and use External Circuit of Typical Application (FIG. 3).

(Note) I<sub>LX</sub> is gradually increased by the external inductor after Lx Switch is turned ON. In accordance with the increase of I<sub>LX</sub>, V<sub>LX</sub> is also increased. When V<sub>LX</sub> reaches V<sub>LXlim</sub>, Lx Switch is turned OFF by Lx Switch Protection Circuit. The time period from the time at which V<sub>LX</sub> reaches V<sub>LXlim</sub> to the time at which Lx Switch is turned OFF is about 3μs.

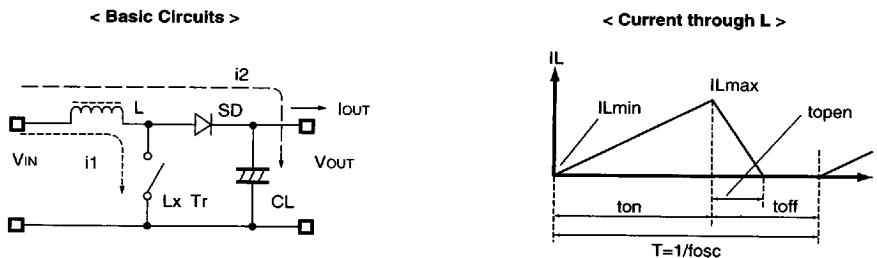
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## OPERATION OF STEP-UP DC/DC CONVERTER

Step-up DC/DC Converter charges energy in the inductor when Lx Transistor (LxTr) is on, and discharges the energy with the addition of the energy from Input Power Source thereto, so that a higher output voltage than the input voltage is obtained.

The operation will be explained with reference to the following diagrams :



Step 1 :  $Lx Tr$  is turned ON and current  $IL$  ( $=i_1$ ) flows, so that energy is charged in  $L$ . At this moment,  $IL$  ( $=i_1$ ) is increased from  $IL_{min}$  ( $=0$ ) to reach  $IL_{max}$  in protection to the on-time period ( $ton$ ) of  $Lx Tr$ .

Step 2 : When  $Lx Tr$  is turned OFF, Schottky diode ( $SD$ ) is turned on in order that  $L$  maintains  $IL$  at  $IL_{max}$ , so that current  $IL$  ( $=i_2$ ) is released.

Step 3 :  $IL$  ( $=i_2$ ) is gradually decreased, and  $IL$  reaches  $IL_{min}$  ( $=0$ ) after a time period of  $t_{open}$ , so that  $SD$  is turned OFF.

In the case of VFM control system, the output voltage is maintained constant by controlling the oscillator frequency ( $fosc$ ) with the on-time period ( $ton$ ) being maintained constant.

In the above two diagrams, the maximum value ( $IL_{max}$ ) and the minimum value ( $IL_{min}$ ) of the current which flows through the inductor are the same as those when  $Lx Tr$  is ON and also when  $Lx Tr$  is OFF.

The difference between  $IL_{max}$  and  $IL_{min}$ , which is represented by  $\Delta I$ , is:

$$\Delta I = IL_{max} - IL_{min} = V_{IN} \cdot ton/L = (V_{OUT} - V_{IN}) \cdot t_{open}/L \quad \text{Equation 1}$$

wherein  $T = 1/fosc = ton + toff$

$$\text{duty (\%)} = ton/T \cdot 100 = ton \cdot fosc \cdot 100$$

$$t_{open} \leq toff$$

In Equation 1,  $V_{IN} \cdot ton/L$  and  $(V_{OUT} - V_{IN}) \cdot t_{open}/L$  are respectively the change in the current at ON, and the change in the current at OFF.

In the VFM system,  $t_{open} < toff$  as illustrated in the above diagram. In this case, the energy charged in the inductor during the time period of  $ton$  is discharged in its entirely during the time period of  $toff$ , so that  $IL_{min}$  becomes zero ( $IL_{min}=0$ ).

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## SELECTION OF PERIPHERAL COMPONENTS

When LxTr is on, the energy PON charged in the inductor is provided by Equation 2 as follows :

$$\begin{aligned} PON &= \int_0^{ton} (VIN \cdot IL(t)) dt = \int_0^{ton} (VIN^2 \cdot t/L) dt \\ &= VIN^2 \cdot ton^2 / (2 \cdot L) \end{aligned} \quad \text{Equation 2}$$

In the case of the step-up DC/DC converter, the energy is also supplied from the input power source at the time of OFF.

$$\begin{aligned} \text{Thus, } POFF &= \int_0^{topen} (VIN \cdot IL(t)) dt = \int_0^{topen} (VIN \cdot (VOUT - VIN) \cdot t/L) dt \\ &= VIN \cdot (VOUT - VIN) \cdot topen^2 / (2 \cdot L) \end{aligned}$$

Here,  $topen = VIN \cdot ton / (VOUT - VIN)$  from Equation 1, and when this is substituted into the above equation.

$$= VIN^3 \cdot ton^2 / (2 \cdot L \cdot (VOUT - VIN)) \quad \text{Equation 3}$$

Input power PIN is  $(PON + POFF)/T$ . When this is converted in its entirely to the output.

$$PIN = (PON + POFF)/T = VOUT \cdot IOUT = POUT \quad \text{Equation 4}$$

Equation 5 can be obtained as follows by solving Equation 4 for IOUT by substituting Equation 2 and 3 into Equation 4 :

$$\begin{aligned} IOUT &= VIN^2 \cdot ton^2 / (2 \cdot L \cdot T \cdot (VOUT - VIN)) \\ &= VIN^2 \cdot maxdty^2 / (20000 \cdot fosc \cdot L \cdot (VOUT - VIN)) \end{aligned} \quad \text{Equation 5}$$

The peak current which flows through  $L \cdot LxTr \cdot SD$  is

$$ILmax = VIN \cdot ton / L \quad \text{Equation 6}$$

Therefore, it is necessary that the setting of the input/output conditions and the selection of peripheral components be made with ILmax taken into consideration.

### HINTS

The above explanation is directed to the calculation in an ideal case where it is supposed that there is no energy loss in the external components and LxSW. In an actual case, the maximum output current will be 50 to 80% of the above calculated maximum output current. In particular, care must be taken because VIN is decreased in an amount corresponding to the voltage reduction caused by LxSW when IL is large or VIN is small. Furthermore, It is required that with respect to VOUT, Vf of the diode (about 0.3V in the case of a Schottky type diode) be taken into consideration.

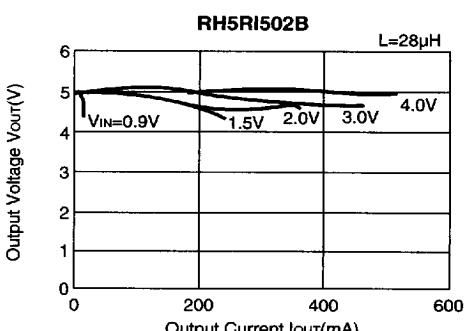
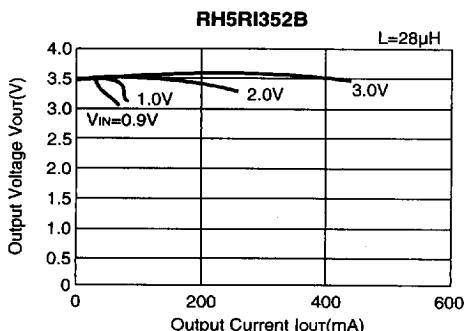
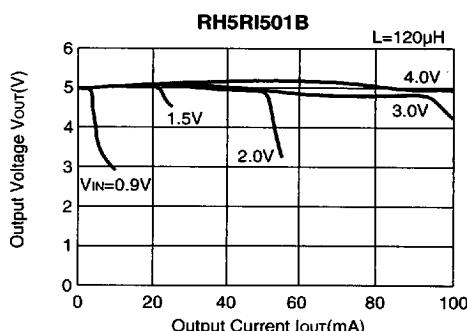
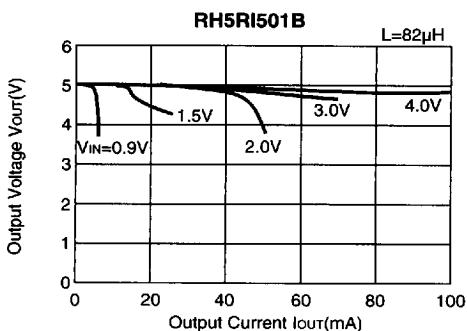
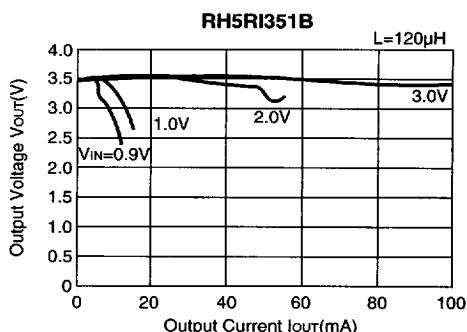
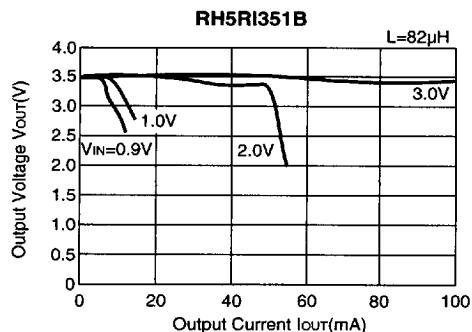
When ILX and VLX exceed their respective ratings, use RH5RI××2B and RH5RI××3B ICs with the attachment of an external transistor with a low saturation voltage thereto.

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## TYPICAL CHARACTERISTICS

### 1) Output Voltage vs. Output Current

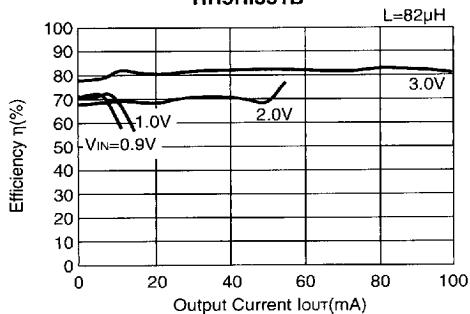


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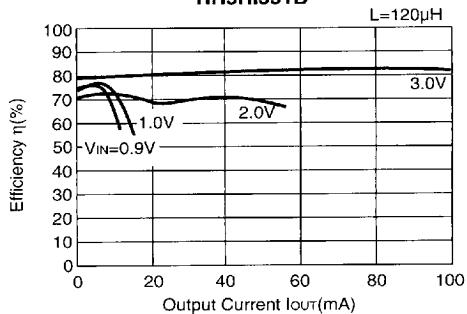
**RICOH**

## 2) Efficiency vs. Output Current

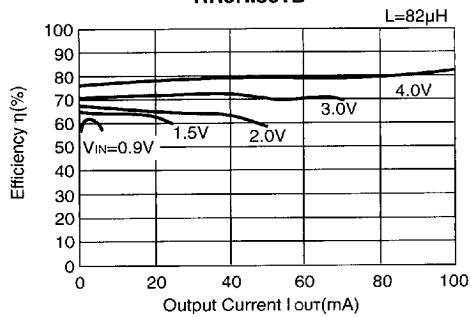
RH5RI351B



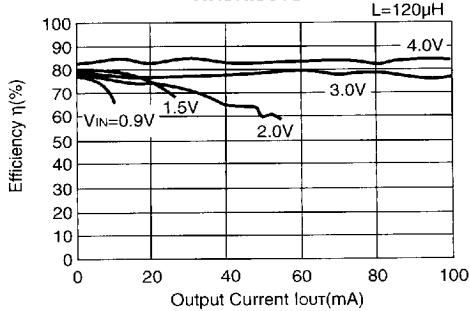
RH5RI351B



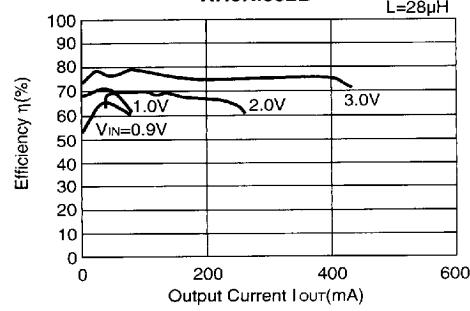
RH5RI501B



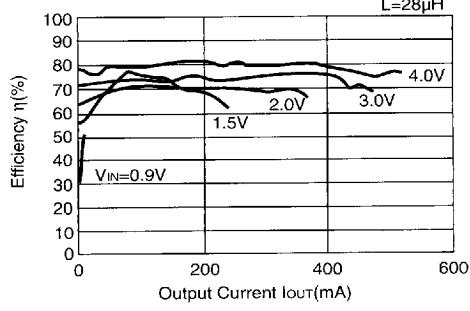
RH5RI501B



RH5RI352B

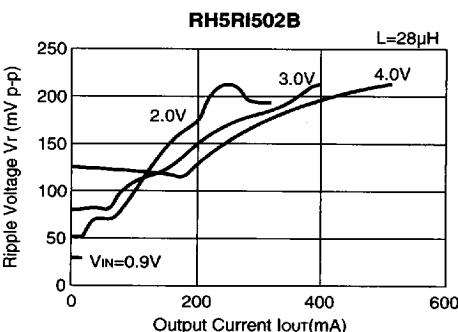
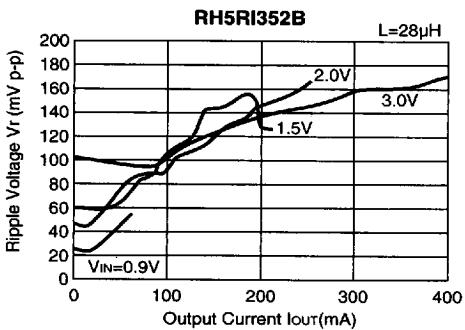
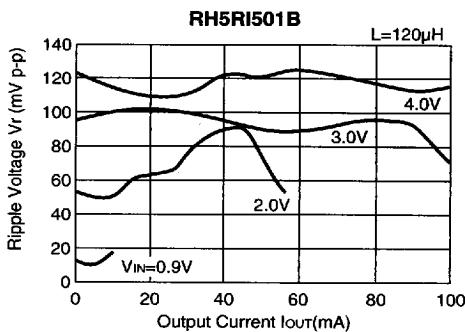
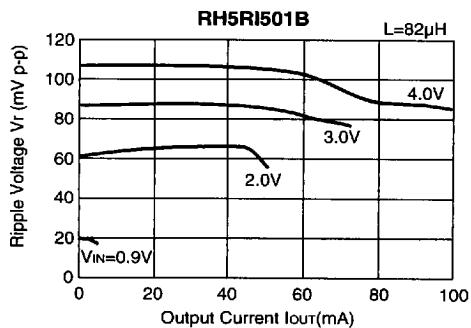
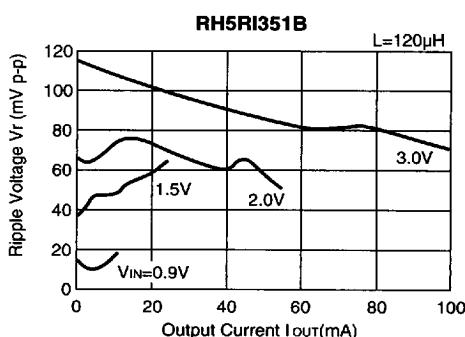
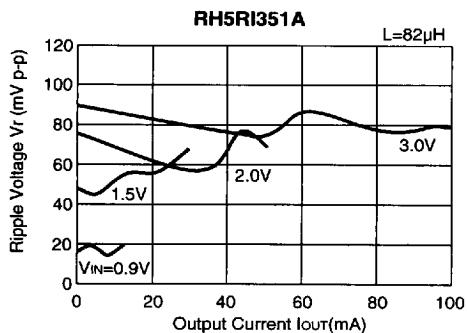


RH5RI502B



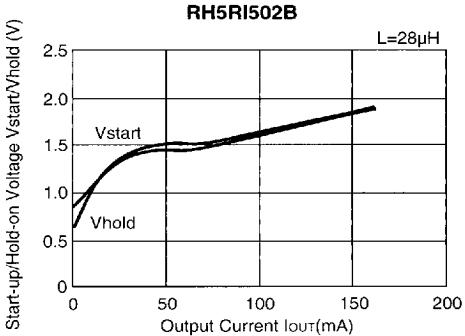
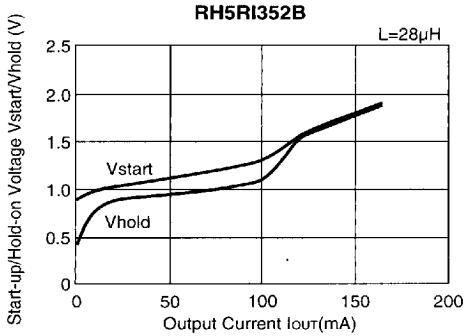
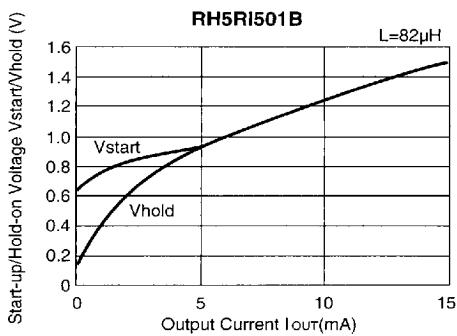
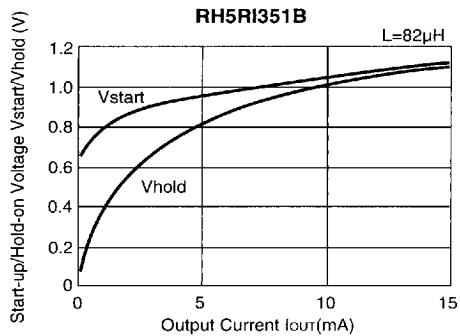
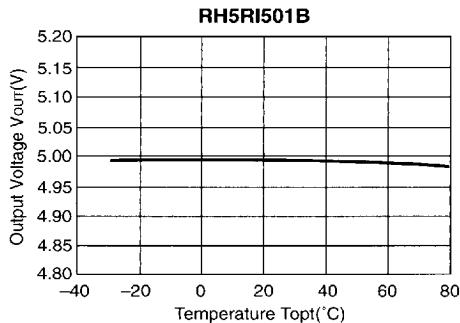
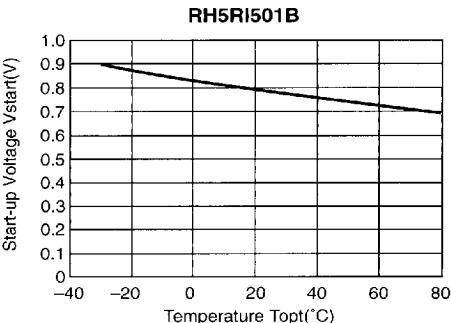
7744690 0004099 7T3

**RICOH**

**3) Output Current vs.Ripple Voltage**

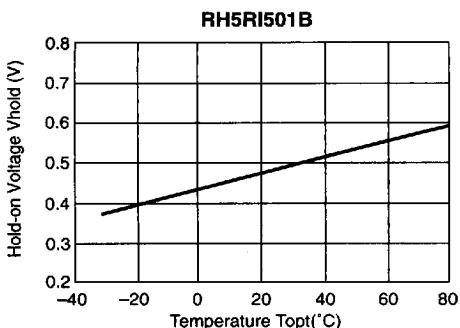
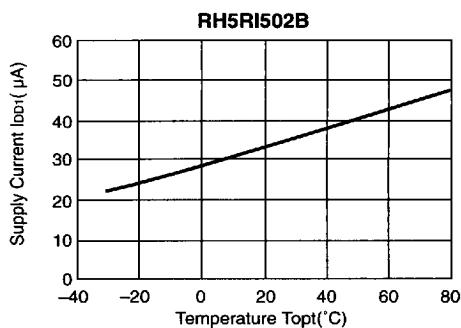
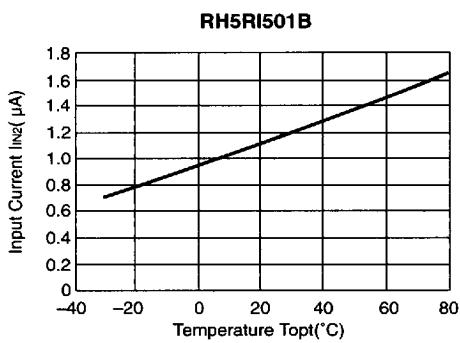
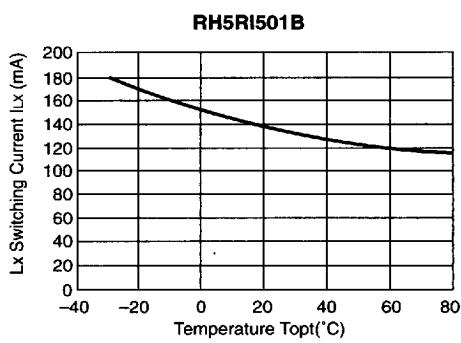
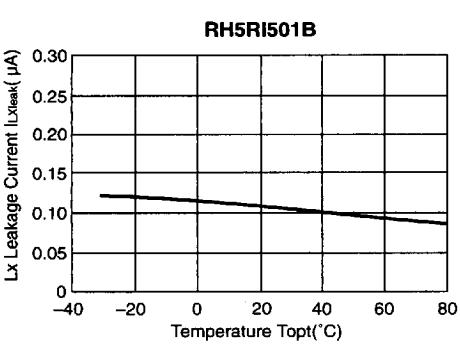
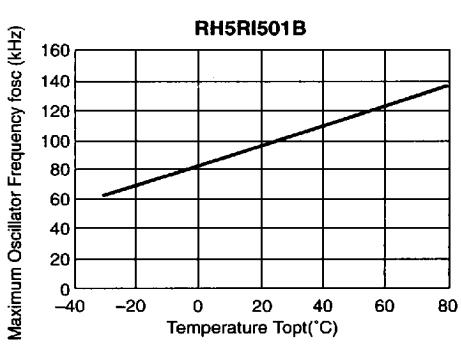
7744690 0004100 245

**RICOH**

**4) Start-up/Hold-on Voltage vs. Output Current****5) Output Voltage vs. Temperature****6) Start-up Voltage vs. Temperature**

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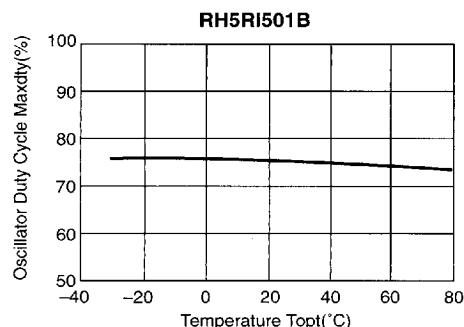
RICOH

**7) Hold-on Voltage vs. Temperature****8) Supply Current 1 vs. Temperature****9) Input Current 2 vs. Temperature****10) Lx Switching Current vs. Temperature****11) Lx Leakage Current vs. Temperature****12) Maximum Oscillator Frequency vs. Temperature**

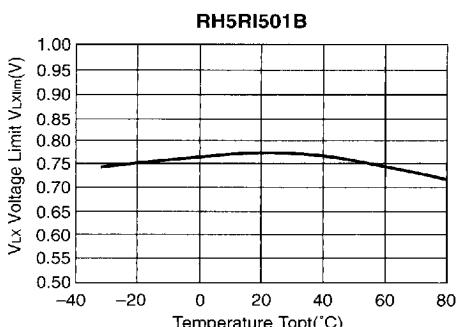
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**RICOH**

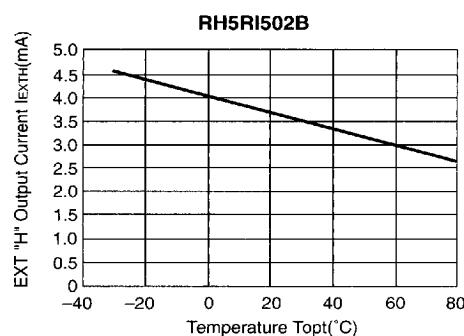
13) Oscillatar Duty Cycle vs. Temperature



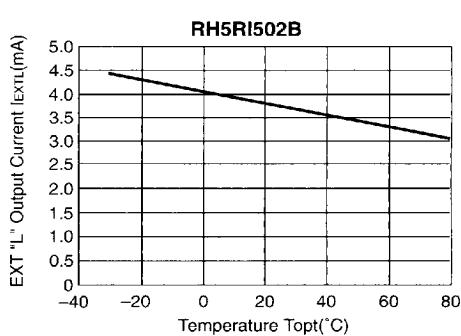
14) Vlx Voltage Limit vs. Temperature



15) Output Current vs. Temperature



16) Output Current vs. Temperature

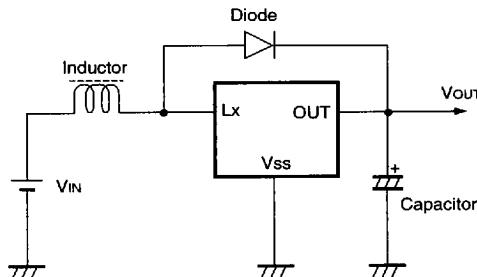


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**RICOH**

## TYPICAL APPLICATIONS

- RH5RI××1B



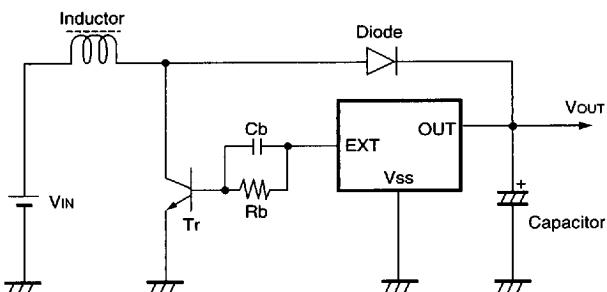
Components Inductor (L) : 82 $\mu$ H (Sumida Electric Co., Ltd.)

Diode (D) : MA721 (Matsushita Electronics Corporation, Schottky Type)

Capacitor (CL) : 22 $\mu$ F (Tantalum Type)

FIG. 1

- RH5RI××2B



Components Inductor (L) : 28 $\mu$ H (Troidal Core)

Diode (D) : HRP22 (Hitachi, Schottky Type)

Capacitor (CL) : 100 $\mu$ F (Tantalum Type)

Transistor (Tr) : 2SD1628G

Base Resistor (Rb) : 300 $\Omega$

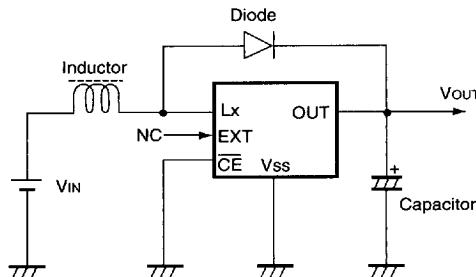
Base Capacitor (Cb) : 0.01 $\mu$ F

FIG. 2

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**RICOH**

## • RH5RI××3B

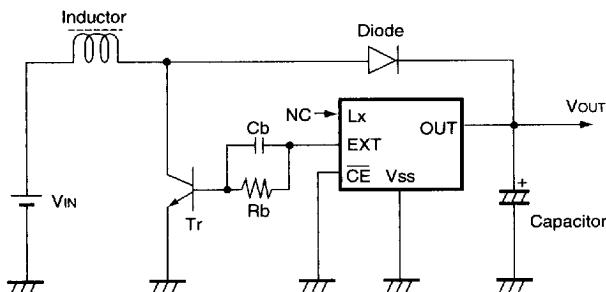


Components Inductor (L) : 82 $\mu$ H (Sumida Electric Co., Ltd.)

Diode (D) : MA721 (Matsushita Electronics Corporation, Schottky Type)

Capacitor (CL) : 22 $\mu$ F (Tantalum Type)

FIG. 3



Components Inductor (L) : 28 $\mu$ H (Troidal Core)

Diode (D) : HRP22 (Hitachi, Schottky Type)

Capacitor (CL) : 100 $\mu$ F (Tantalum Type)

Transistor (Tr) : 2SD1628G

Base Resistor (Rb) : 300 $\Omega$

Base Capacitor (Cb) : 0.01 $\mu$ F

FIG. 4

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• **CE pin Drive Circuit**

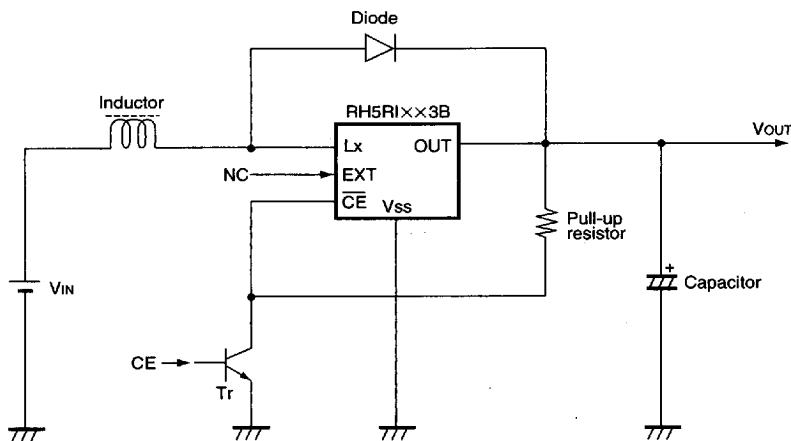


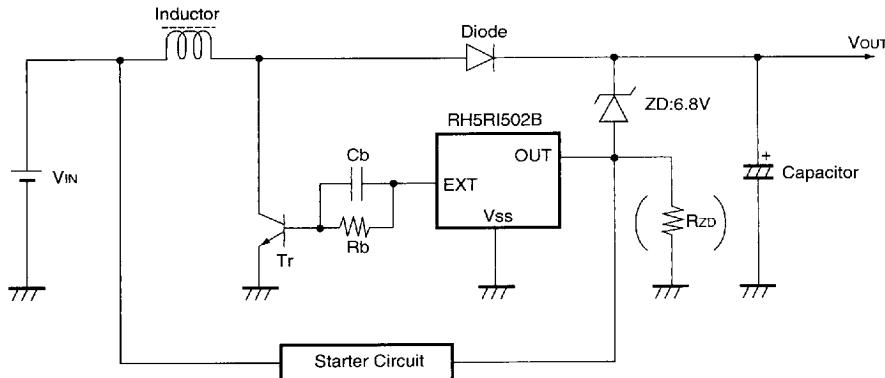
FIG. 5

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**RICOH**

## APPLICATION CIRCUITS

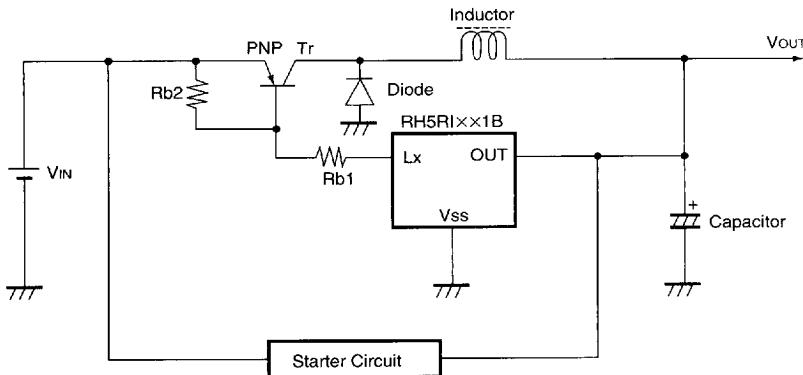
- 12V Step-up Circuit



(Note) When the Output Current is small or the Output Voltage is unstable,use the  $R_{ZD}$  for flowing the bias current through the Zener diode  $ZD$ .

FIG. 6

- Step-down Circuit



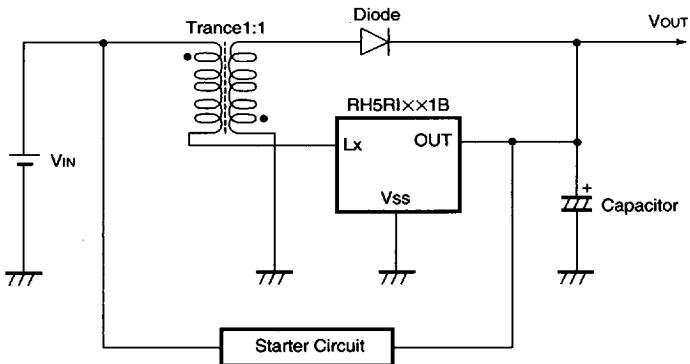
(Note) When the  $Lx$  pin Voltage is over the rating at the time PNP  $Tr$  is OFF,use a RH5RIxx2B and drive the PNP  $Tr$  by the external NPN  $Tr$ .

FIG. 7

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**RICOH**

• Step-up/Step-down Circuit with Flyback

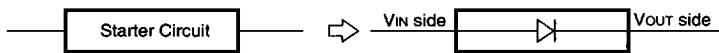


(Note) Use a RH5RIxx2B, depend on the Output Current.

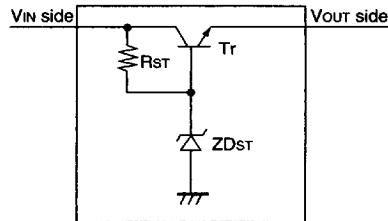
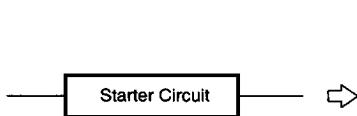
**FIG. 8**

\*The Starter Circuit is necessary for all above circuits.

1. For Step-up Circuit.



1. For Step-down and Step-up/Step-down Circuit.



$Z_{DST} \leq 2.5V \leq$  Designation of Output Voltage  
 $R_{ST}$  Input Bias Current of  $Z_{DST}$  and  $Tr$ .  
 (several kΩ to several hundreds kΩ)

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**RICOH**

**APPLICATION HINTS**

When using these ICs, be sure to take care of the following points :

- Set external components as close as possible to the IC and minimize the connection between the components and the IC. In particular, when an external component is connected to OUT Pin, make minimum connection with the capacitor.
- Make sufficient grounding. A large current flows through Vss Pin by switching. When the impedance of the Vss connection is high, the potential within the IC is varied by the switching current. This may result in unstable operation of the IC.
- Use capacitor with a capacity of 10 $\mu$ F or more, and with good high frequency characteristics such as tantalum capacitor. We recommend the use of a capacitor with an allowable voltage which is at least three times the output set voltage. This is because there may be the case where a spike-shaped high voltage is generated by the inductor when Lx transistor is turned OFF.
- Take the utmost care when choosing an inductor. Namely, choose such an inductor that has sufficiently small d.c. resistance and large allowable current, and hardly reaches magnetic saturation. When the inductance value of the inductor is small, there may be the case where ILX exceeds the absolute maximum ratings at the maximum load. Use an inductor with an appropriate inductance (refer to Selection of peripheral components).
- Use a diode of a Schottky type with high switching speed, and also take care of the rated current (refer to Selection of peripheral components).

The performance of power source circuits using these ICs largely depends upon the peripheral circuits. Take the utmost care in the selection of the peripheral circuits. In particular, design the peripheral circuits in such a manner that the values such as voltage, current and power of each component, PCB patterns and the IC do not exceed their respective rated values.

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**RICOH**