

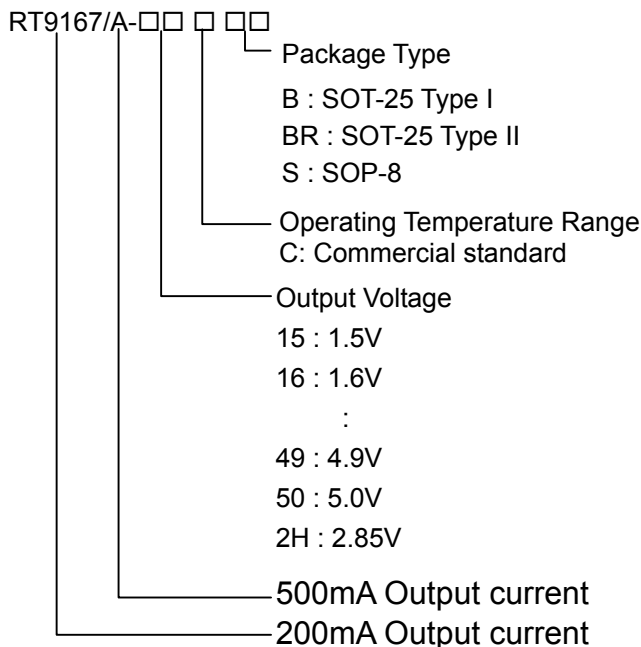
Low-Noise, Fixed Output Voltage, 200mA/500mA LDO Regulator

General Description

The RT9167/A is a 200mA/500mA low dropout and low noise micropower regulator suitable for portable applications. The output voltages range from 1.5V to 5.0V in 100mV increments and 2% accuracy. The RT9167/A is designed for use with very low ESR capacitors. The output remains stable even with 1 μ F ceramic output capacitor.

The RT9167/A uses an internal PMOS as the pass device, which does not cause extra GND current in heavy load and dropout conditions. The shutdown mode of nearly zero operation current makes the IC suitable for battery-power devices. Other features include a reference bypass pin to improve low noise performance, current limiting, and over temperature protection.

Ordering Information



Marking Information

For marking information, contact our sales representative directly or through a RichTek distributor located in your area, otherwise visit our website for detail.

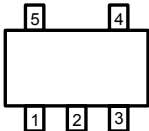
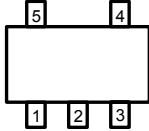
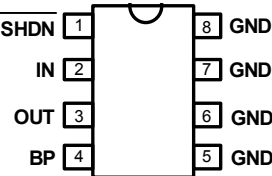
Features

- Stable with Low-ESR Output Capacitor
- Low Dropout Voltage (220mV and 200mA)
- Low Operation Current - 80 μ A Typical
- Shutdown Function
- Low Noise Output
- Low Temperature Coefficient
- Current and Thermal Limiting
- Custom Voltage Available
- SOT-25 and SOP-8 Packages

Applications

- Cellular Telephones
- Laptop, Notebook, and Palmtop Computers
- Battery-powered Equipment
- Hand-held Equipment

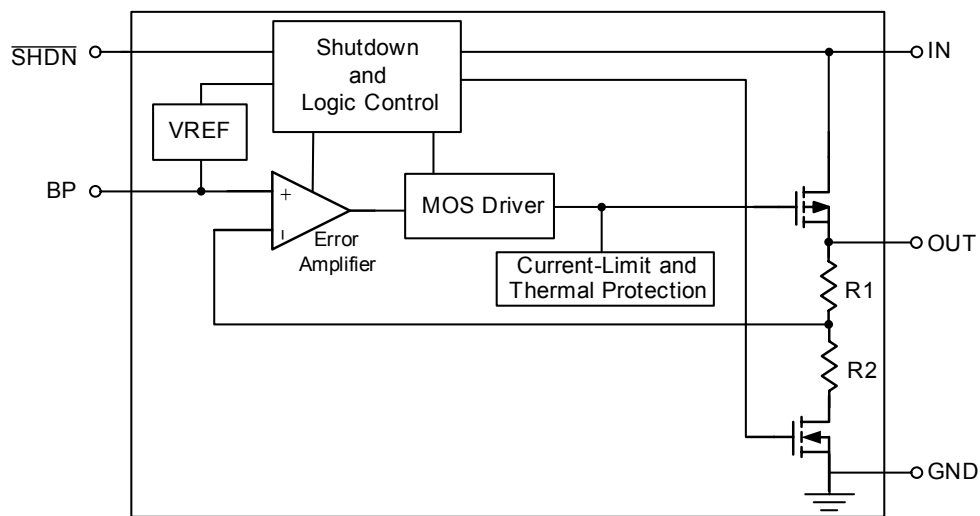
Pin Configurations

Part Number	Pin Configurations
RT9167/A-□□CB (Plastic SOT-25)	<p>TOP VIEW</p>  <ol style="list-style-type: none"> 1. IN 2. <u>GND</u> 3. SHDN 4. BP 5. OUT
RT9167/A-□□CBR (Plastic SOT-25)	<p>TOP VIEW</p>  <ol style="list-style-type: none"> 1. OUT 2. GND 3. <u>IN</u> 4. SHDN 5. BP
RT9167/A-□□CS (Plastic SOP-8)	<p>TOP VIEW</p>  <ol style="list-style-type: none"> SHDN 1 IN 2 OUT 3 BP 4 8 GND 7 GND 6 GND 5 GND

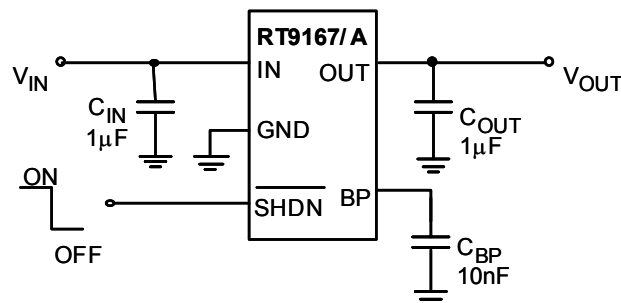
Pin Description

Pin Name	Pin Function
IN	Input
GND	Ground
$\overline{\text{SHDN}}$	Active Low Shutdown Input
BP	Reference Noise Bypass
OUT	Output

Function Block Diagram



Typical Application Circuit



Absolute Maximum Ratings

- Input Voltage 8V
- Power Dissipation, P_D @ $T_A = 25^\circ\text{C}$
 - SOT-25 0.25W
 - SOP-8 0.625W
- Operating Junction Temperature Range -40°C to 125°C
- Storage Temperature Range -65°C to 150°C
- Package Thermal Resistance
 - SOT-25, θ_{JA} 250°C/W
 - SOP-8, θ_{JA} 160°C/W
- Lead Temperature (Soldering, 5 sec.) 260°C

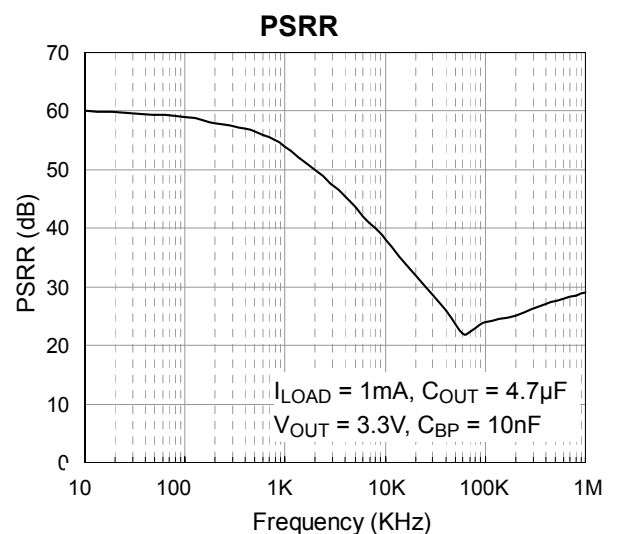
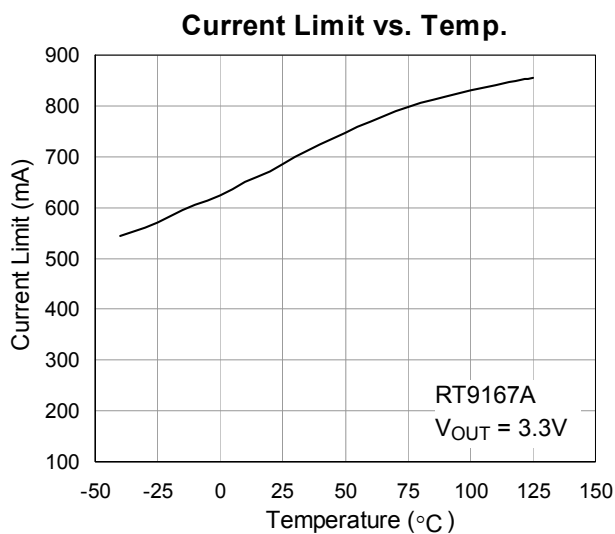
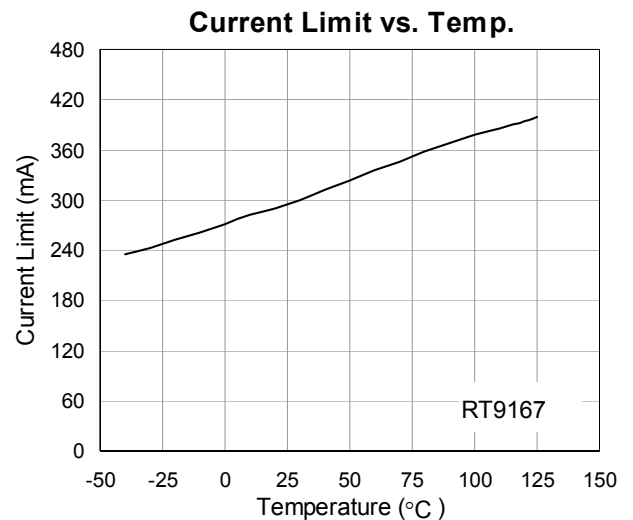
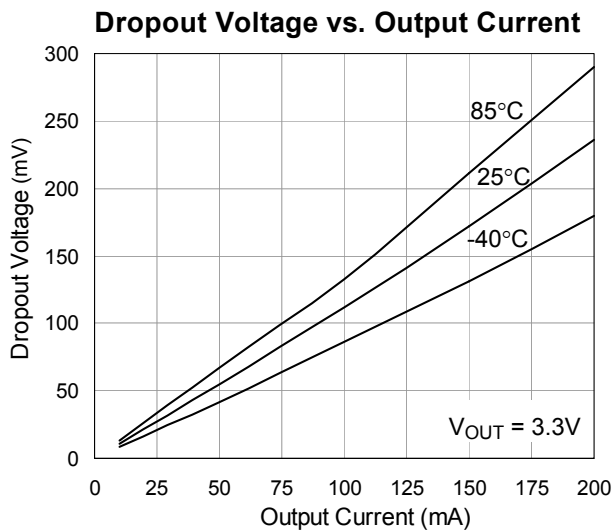
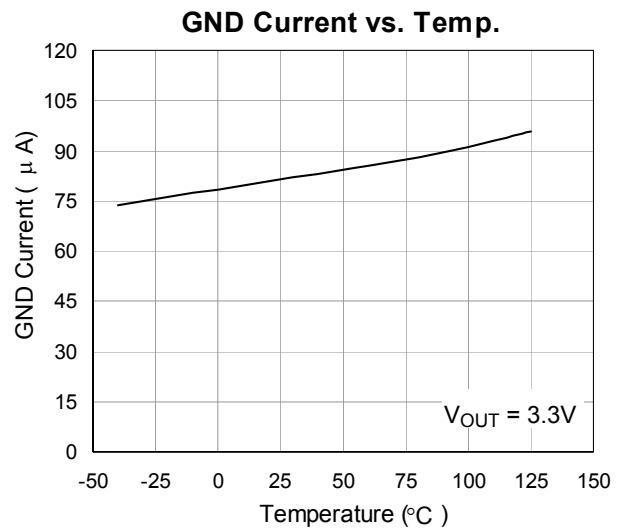
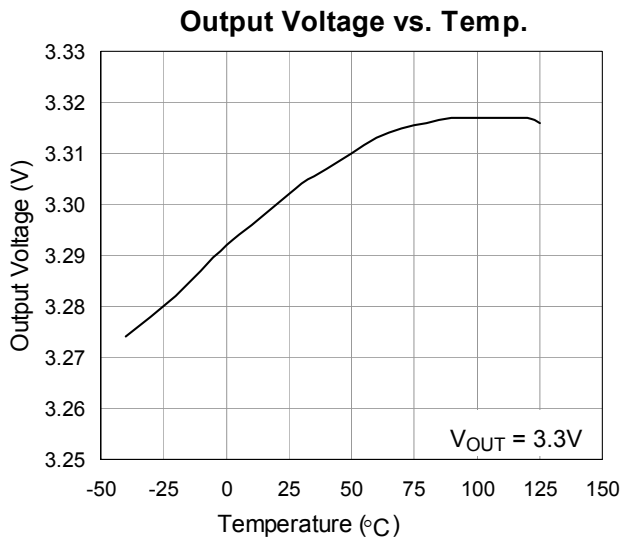
Electrical Characteristics

($V_{IN} = 5.0\text{V}$, $C_{IN} = 1\mu\text{F}$, $C_{OUT} = 1\mu\text{F}$, $T_A = 25^\circ\text{C}$, unless otherwise specified)

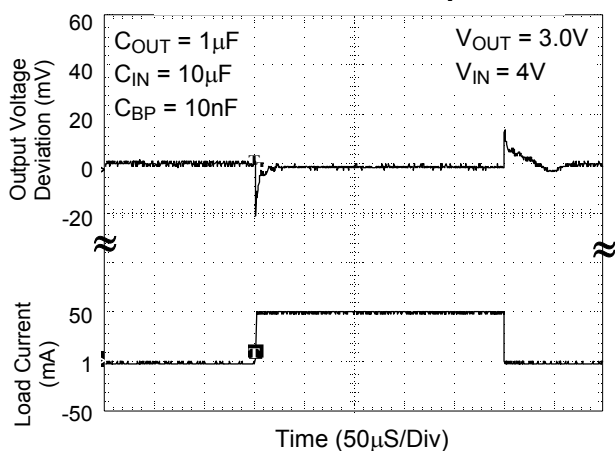
Parameter		Symbol	Test Conditions	Min	Typ	Max	Units
Input Voltage Range		V_{IN}		2.9	--	7	V
			$I_L = 50\text{mA}$	2.7	--	7	
Output Voltage Accuracy		ΔV_{OUT}	$I_L = 1\text{mA}$	-2	--	+2	%
Maximum Output Current	RT9167	I_{MAX}		200	--	--	mA
	RT9167A			500	--	--	
Current Limit	RT9167	I_{LIMIT}	$R_{LOAD} = 1\Omega$	--	300	--	mA
	RT9167A			--	700	--	
GND Pin Current	RT9167/A	I_G	No Load	--	80	150	μA
	RT9167		$I_{OUT} = 200\text{mA}$	--	90	150	
	RT9167A		$I_{OUT} = 500\text{mA}$	--	90	150	
Dropout Voltage ^(Note) ($V_{OUT}(\text{Nominal}) \geq 3.0\text{V}$ Version)	RT9167/A	V_{DROP}	$I_{OUT} = 1\text{mA}$	--	1.1	5	mV
	RT9167/A		$I_{OUT} = 50\text{mA}$	--	55	100	
	RT9167/A		$I_{OUT} = 200\text{mA}$	--	220	300	
	RT9167A		$I_{OUT} = 500\text{mA}$	--	600	750	
Line Regulation		ΔV_{LINE}	$V_{IN} = (V_{OUT} + 0.15)$ to 7V , $I_{OUT} = 1\text{mA}$	--	--	6	mV/V
Load Regulation	RT9167	ΔV_{LOAD}	$I_{OUT} = 0\text{mA}$ to 200mA	--	--	25	mV
	RT9167A		$I_{OUT} = 0\text{mA}$ to 500mA	--	--	35	
SHDN Input High Threshold		V_{IH}	$V_{IN} = 3\text{V}$ to 5.5V	1.6	--	--	V
SHDN Input Low Threshold		V_{IL}	$V_{IN} = 3\text{V}$ to 5.5V	--	--	0.4	V
SHDN Bias Current		I_{SD}		--	--	100	nA
Shutdown Supply Current		I_{GSD}	$V_{OUT} = 0\text{V}$	--	0.01	1	μA
Thermal Shutdown Temperature		T_{SD}		--	155	--	$^\circ\text{C}$
Output Noise		e_{NO}	$C_{BP} = 10\text{nF}$, $C_{OUT} = 10\mu\text{F}$	--	350	--	$\text{nV} \sqrt{\text{Hz}}$
Ripple Rejection		PSRR	$F = 100\text{Hz}$, $C_{BP} = 10\text{nF}$, $C_{OUT} = 10\mu\text{F}$	--	58	--	dB

Notes: Dropout voltage definition: $V_{IN} - V_{OUT}$ when V_{OUT} is 50 mV below the value of V_{OUT} at $V_{IN} = V_{OUT} + 0.5\text{V}$

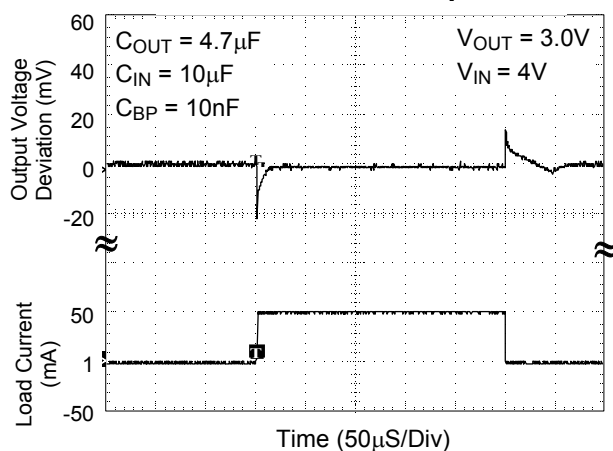
Typical Operating Characteristics



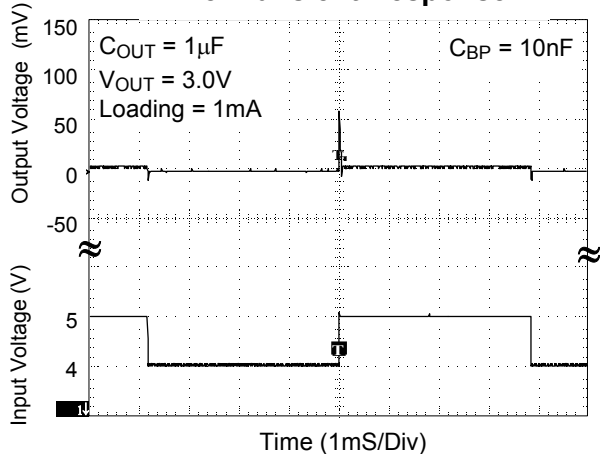
Load Transient Response



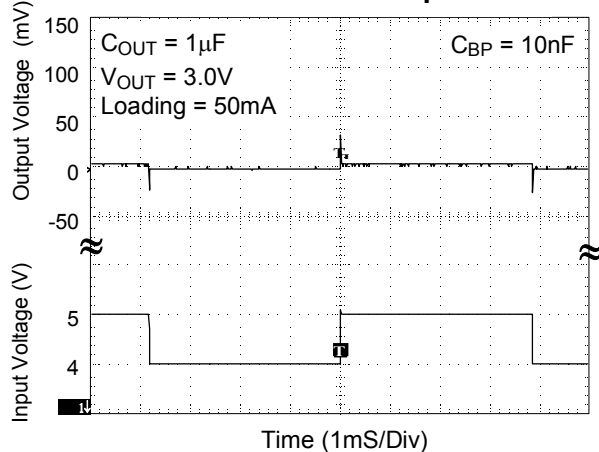
Load Transient Response



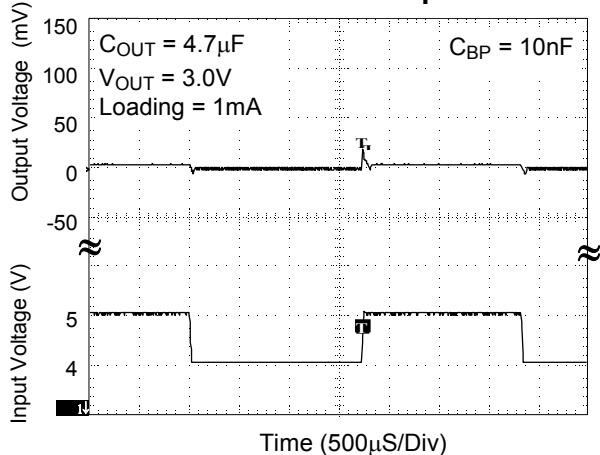
Line Transient Response



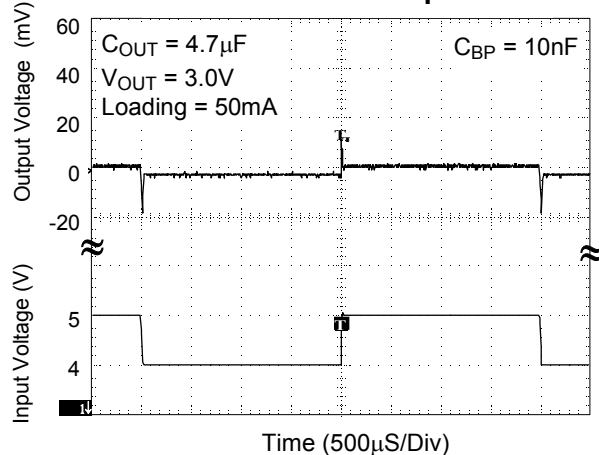
Line Transient Response



Line Transient Response



Line Transient Response



Applications Guides

Capacitor Selection and Regulator Stability

Like any low-dropout regulator, the external capacitors used with the RT9167/A must be carefully selected for regulator stability and performance.

Using a capacitor whose value is $> 1\mu\text{F}$ on the RT9167/A input and the amount of capacitance can be increased without limit. The input capacitor must be located a distance of not more than 0.5" from the input pin of the IC and returned to a clean analog ground. Any good quality ceramic or tantalum can be used for this capacitor. The capacitor with larger value and lower ESR (equivalent series resistance) provides better PSRR and line-transient response.

The output capacitor must meet both requirements for minimum amount of capacitance and ESR in all LDOs application. The RT9167/A is designed specifically to work with low ESR ceramic output capacitor in space-saving and performance consideration. Using a ceramic capacitor whose value is at least $1\mu\text{F}$ with ESR is $> 5\text{m}\Omega$ on the RT9167/A output ensures stability. The RT9167/A still works well with output capacitor of other types due to the wide stable ESR range. Fig.1 shows the curves of allowable ESR range as a function of load current for various output voltages and capacitor values. Output capacitor of larger capacitance can reduce noise and improve load-transient response, stability, and PSRR. The output capacitor should be located not more than 0.5" from the V_{OUT} pin of the RT9167/A and returned to a clean analog ground.

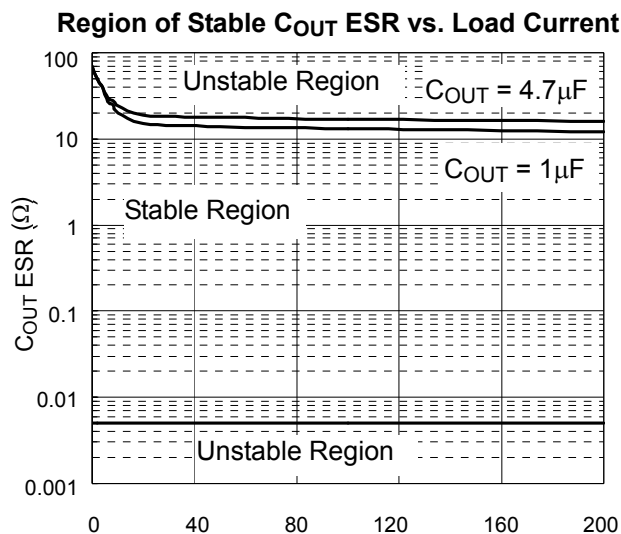


Fig. 1

Note that some ceramic dielectrics exhibit large capacitance and ESR variation with temperature. It may be necessary to use $2.2\mu\text{F}$ or more to ensure stability at temperatures below -10°C in this case. Also, tantalum capacitors, $2.2\mu\text{F}$ or more may be needed to maintain capacitance and ESR in the stable region for strict application environment.

Tantalum capacitors maybe suffer failure due to surge current when it is connected to a low-impedance source of power (like a battery or very large capacitor). If a tantalum capacitor is used at the input, it must be guaranteed to have a surge current rating sufficient for the application by the manufacture.

Use a 10nF bypass capacitor at BP for low output voltage noise. The capacitor, in conjunction with an internal $200\text{K}\Omega$ resistor, which connects bypass pin and the band-gap reference, creates an 80Hz low-pass filter for noise reduction. Increasing the capacitance will slightly decrease the output noise, but increase the start-up time. The capacitor connected to the bypass pin for noise reduction must have very low leakage. This capacitor leakage current causes the output voltage to decline by a proportional amount to the current due to the voltage drop on the internal $200\text{K}\Omega$ resistor. Fig. 2 shows the power on response.

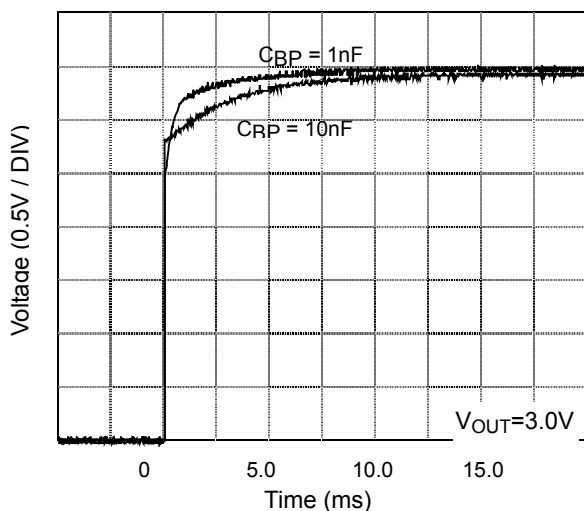


Fig. 2

Load-Transient Considerations

The RT9167/A load-transient response graphs (see Typical Operating Characteristics) show two components of the output response: a DC shift from the output impedance due to the load current change, and the transient response. The DC shift is quite small due to the excellent load regulation of the IC. Typical output voltage transient spike for a step change in the load current from 0mA to 50mA is tens mV, depending on the ESR of the output capacitor. Increasing the output capacitor's value and decreasing the ESR attenuates the overshoot.

Shutdown Input Operation

The RT9167/A is shutdown by pulling the SHDN input low, and turned on by driving the input high. If this feature is not to be used, the SHDN input should be tied to V_{IN} to keep the regulator on at all times (the SHDN input must **not** be left floating).

To ensure proper operation, the signal source used to drive the SHDN input must be able to swing above and below the specified turn-on/turn-off voltage thresholds which guarantee an ON or OFF state (see Electrical Characteristics). The ON/OFF signal may come from either CMOS output, or an open-collector output with pull-up resistor to the RT9167/A input voltage or another logic supply. The high-level voltage may exceed the RT9167/A input voltage, but must remain within the absolute maximum ratings for the SHDN pin.

Internal P-Channel Pass Transistor

The RT9167/A features a typical 1.1Ω P-channel MOSFET pass transistor. It provides several advantages over similar designs using PNP pass transistors, including longer battery life. The P-channel MOSFET requires no base drive, which reduces quiescent current considerably. PNP-based regulators waste considerable current in dropout when the pass transistor saturates. They also use high base-drive currents under large loads. The RT9167/A does not suffer from these problems and consume only $80\mu\text{A}$ of quiescent current whether in dropout, light-load, or heavy-load applications.

Input-Output (Dropout) Voltage

A regulator's minimum input-output voltage differential (or dropout voltage) determines the lowest usable supply voltage. In battery-powered systems, this will determine the useful end-of-life battery voltage. Because the RT9167/A uses a P-channel MOSFET pass transistor, the dropout voltage is a function of drain-to-source on-resistance $[R_{DS(ON)}]$ multiplied by the load current.

Reverse Current Path

The power transistor used in the RT9167/A has an inherent diode connected between the regulator input and output (see Fig.3). If the output is forced above the input by more than a diode-drop, this diode will become forward biased and current will flow from the V_{OUT} terminal to V_{IN} . This diode will also be turned on by abruptly stepping the input voltage to a value below the output voltage. To prevent regulator mis-operation, a Schottky diode should be used in any applications where input/output voltage conditions can cause the internal diode to be turned on (see Fig.4). As shown, the Schottky diode is connected in parallel with the internal parasitic diode and prevents it from being turned on by limiting the voltage drop across it to about 0.3V. $< 100\text{ mA}$ to prevent damage to the part.

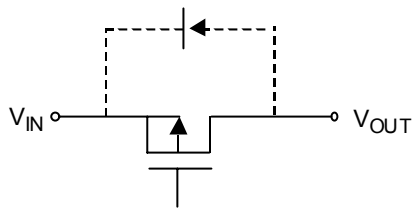


Fig. 3

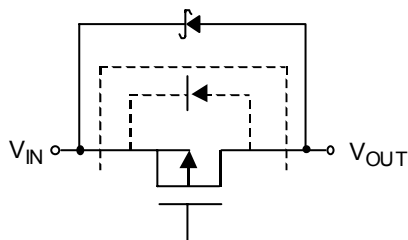


Fig. 4

Current Limit and Thermal Protection

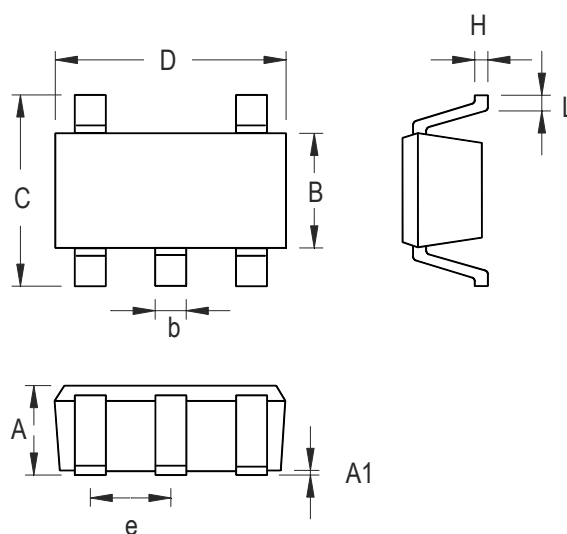
The RT9167 includes a current limit which monitors and controls the pass transistor's gate voltage limiting the output current to 300mA Typ. (700mA Typ. for RT9167A). Thermal-overload protection limits total power dissipation in the RT9167/A. When the junction temperature exceeds $T_J = +155^{\circ}\text{C}$, the thermal sensor signals the shutdown logic turning off the pass transistor and allowing the IC to cool. The thermal sensor will turn the pass transistor on again after the IC's junction temperature cools by 10°C , resulting in a pulsed output during continuous thermal-overload conditions. Thermal-overload protection is designed to protect the RT9167/A in the event of fault conditions. Do not exceed the absolute maximum junction-temperature rating of $T_J = +150^{\circ}\text{C}$ for continuous operation. The output can be shorted to ground for an indefinite amount of time without damaging the part by cooperation of current limit and thermal protection.

Operating Region and Power Dissipation

The maximum power dissipation of RT9167/A depends on the thermal resistance of the case and circuit board, the temperature difference between the die junction and ambient air, and the rate of airflow. The power dissipation across the device is $P = I_{OUT} (V_{IN} - V_{OUT})$. The maximum power dissipation is: $P_{MAX} = (T_J - T_A) / \theta_{JA}$

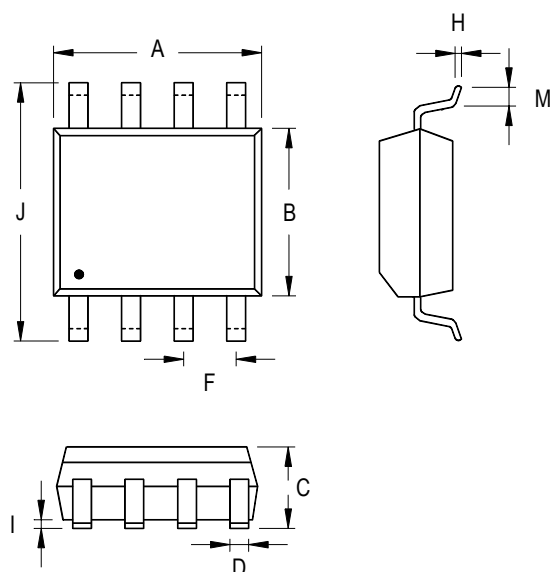
where $T_J - T_A$ is the temperature difference between the RT9167/A die junction and the surrounding environment, θ_{JA} is the thermal resistance from the junction to the surrounding environment. The GND pin of the RT9167/A performs the dual function of providing an electrical connection to ground and channeling heat away. Connect the GND pin to ground using a large pad or ground plane.

Package Information



Symbol	Dimensions In Millimeters		Dimensions In Inches	
	Min	Max	Min	Max
A	0.889	1.295	0.035	0.051
A1	0.000	0.152	0.000	0.006
B	1.397	1.803	0.055	0.071
b	0.356	0.559	0.014	0.022
C	2.591	2.997	0.102	0.118
D	2.692	3.099	0.106	0.122
e	0.838	1.041	0.033	0.041
H	0.102	0.254	0.004	0.010
L	0.356	0.610	0.014	0.024

SOT- 25 Surface Mount Package



Symbol	Dimensions In Millimeters		Dimensions In Inches	
	Min	Max	Min	Max
A	4.801	5.004	0.189	0.197
B	3.810	3.988	0.150	0.157
C	1.346	1.753	0.053	0.069
D	0.330	0.508	0.013	0.020
F	1.194	1.346	0.047	0.053
H	0.178	0.254	0.007	0.010
I	0.102	0.254	0.004	0.010
J	5.791	6.198	0.228	0.244
M	0.406	1.270	0.016	0.050

8-Lead SOP Plastic Package

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