

**TPS77501-EP, TPS77515-EP, TPS77516-EP, TPS77518-EP
TPS77525-EP, TPS77533-EP WITH RESET OUTPUT
FAST-TRANSIENT-RESPONSE 500-mA LOW-DROPOUT VOLTAGE REGULATORS**

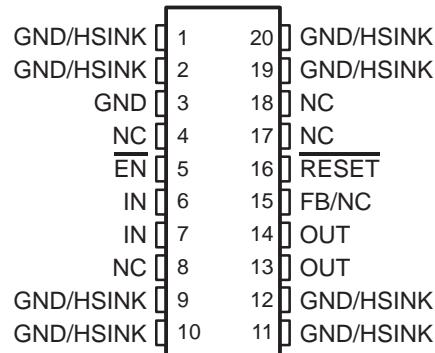
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- Controlled Baseline
 - One Assembly/Test Site, One Fabrication Site
- Extended Temperature Performance of -55°C to 125°C
- Enhanced Diminishing Manufacturing Sources (DMS) Support
- Enhanced Product Change Notification
- Qualification Pedigree[†]
- Open Drain Power-On Reset With 200-ms Delay
- 500-mA Low-Dropout Voltage Regulator
- Available in 1.5-V, 1.6-V, 1.8-V, 2.5-V, 3.3-V Fixed Output and Adjustable Versions
- Dropout Voltage to 169 mV (Typ) at 500 mA (TPS77533)

[†] Component qualification in accordance with JEDEC and industry standards to ensure reliable operation over an extended temperature range. This includes, but is not limited to, Highly Accelerated Stress Test (HAST) or biased 85/85, temperature cycle, autoclave or unbiased HAST, electromigration, bond intermetallic life, and mold compound life. Such qualification testing should not be viewed as justifying use of this component beyond specified performance and environmental limits.

- Ultralow 85 μA Typical Quiescent Current
- Fast Transient Response
- 2% Tolerance Over Specified Conditions for Fixed-Output Versions
- 20-Pin TSSOP PowerPAD™ (PWP) Package
- Thermal Shutdown Protection

**PWP PACKAGE
(TOP VIEW)**



NC – No internal connection

description

The TPS775xx devices are designed to have a fast transient response and be stable with a 10- μF low ESR capacitors. This combination provides high performance at a reasonable cost.

Because the PMOS device behaves as a low-value resistor, the dropout voltage is very low (typically 169 mV at an output current of 500 mA for the TPS77533) and is directly proportional to the output current. Additionally, since the PMOS pass element is a voltage-driven device, the quiescent current is very low and independent of output loading (typically 85 μA over the full range of output current, 0 mA to 500 mA). These two key specifications yield a significant improvement in operating life for battery-powered systems. This LDO family also features a sleep mode; applying a TTL high signal to $\overline{\text{EN}}$ (enable) shuts down the regulator, reducing the quiescent current to 1 μA at $T_J = 25^{\circ}\text{C}$.

The $\overline{\text{RESET}}$ output of the TPS775xx initiates a reset in microcomputer and microprocessor systems in the event of an undervoltage condition. An internal comparator in the TPS775xx monitors the output voltage of the regulator to detect an undervoltage condition on the regulated output voltage.

The TPS775xx is offered in 1.5-V, 1.6-V, 1.8-V, 2.5-V, and 3.3-V fixed-voltage versions and in an adjustable version (programmable over the range of 1.5 V to 5.5 V for TPS77501 option. Output voltage tolerance is specified as a maximum of 2% over line, load, and temperature ranges. The TPS775xx family is available in 20 pin TSSOP package.



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.

PowerPAD is a trademark of Texas Instruments.

PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

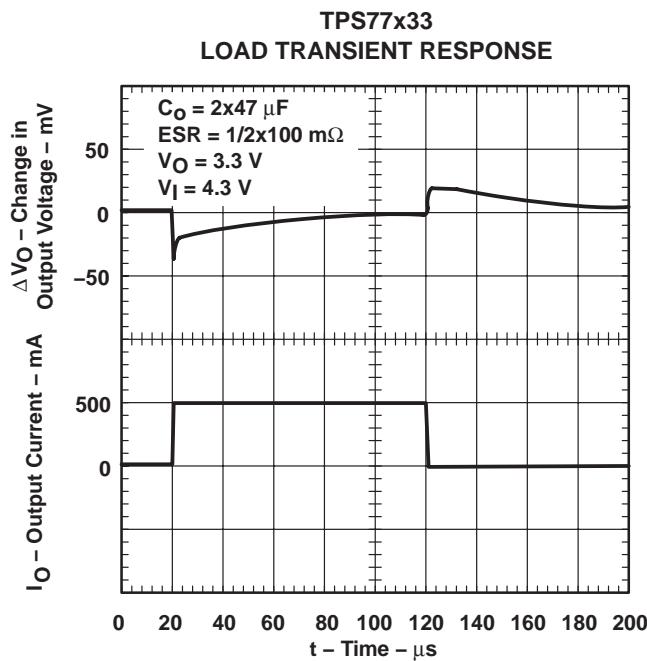
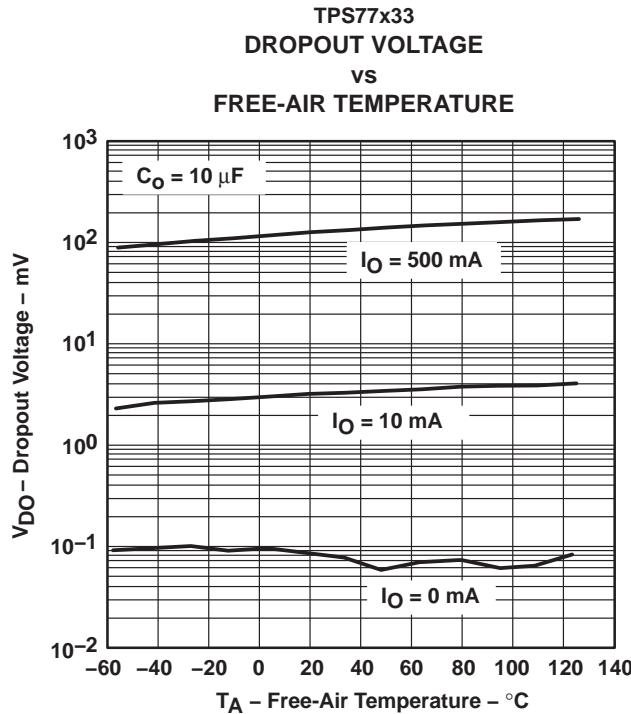
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ORDERING INFORMATION

T_A	OUTPUT VOLTAGE (V TYP)	PACKAGE		ORDERABLE PART NUMBER	TOP-SIDE MARKING
–55°C to 125°C	3.3	TSSOP – PW	Tape and reel	TPS77533MPWPREP	77533ME
	2.5	TSSOP – PW	Tape and reel	TPS77525MPWPREP	77525ME
	1.8	TSSOP – PW	Tape and reel	TPS77518MPWPREP	77518ME
	1.6	TSSOP – PW	Tape and reel	TPS77516MPWPREP\$	77516ME
	1.5	TSSOP – PW	Tape and reel	TPS77515MPWPREP	77515ME
	Adjustable† 1.5 V to 5.5 V	TSSOP – PW	Tape and reel	TPS77501MPWPREP	77501ME

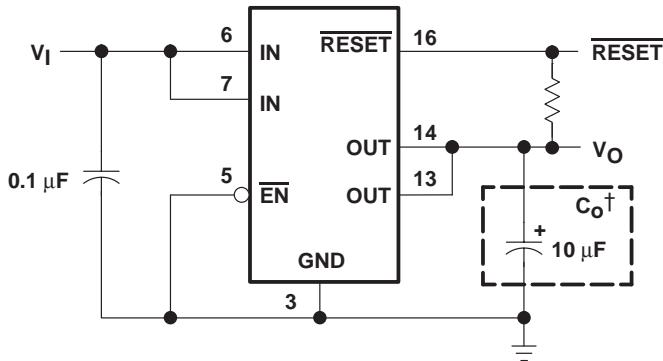
† Package drawings, standard packing quantities, thermal data, symbolization, and PCB design guidelines are available at www.ti.com/sc/package.

‡ The TPS77501 is programmable using an external resistor divider (see application information).

\$ TPS77516 is Product Preview.

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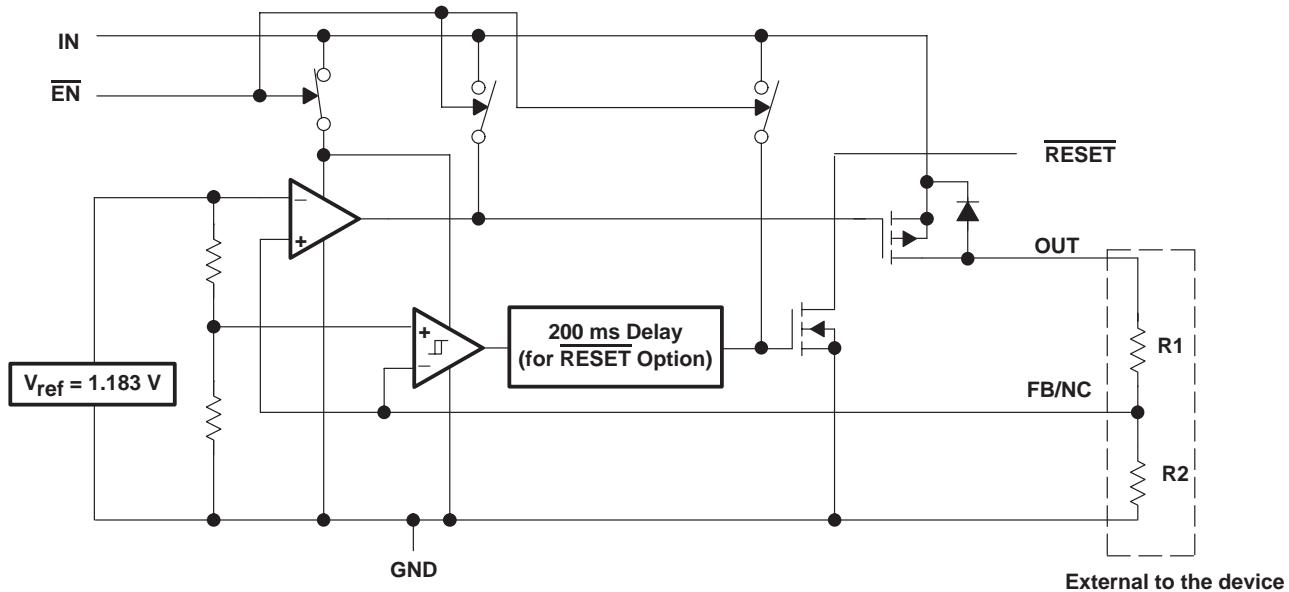
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† See application information section for capacitor selection details.

Figure 1. Typical Application Configuration for Fixed Output Options

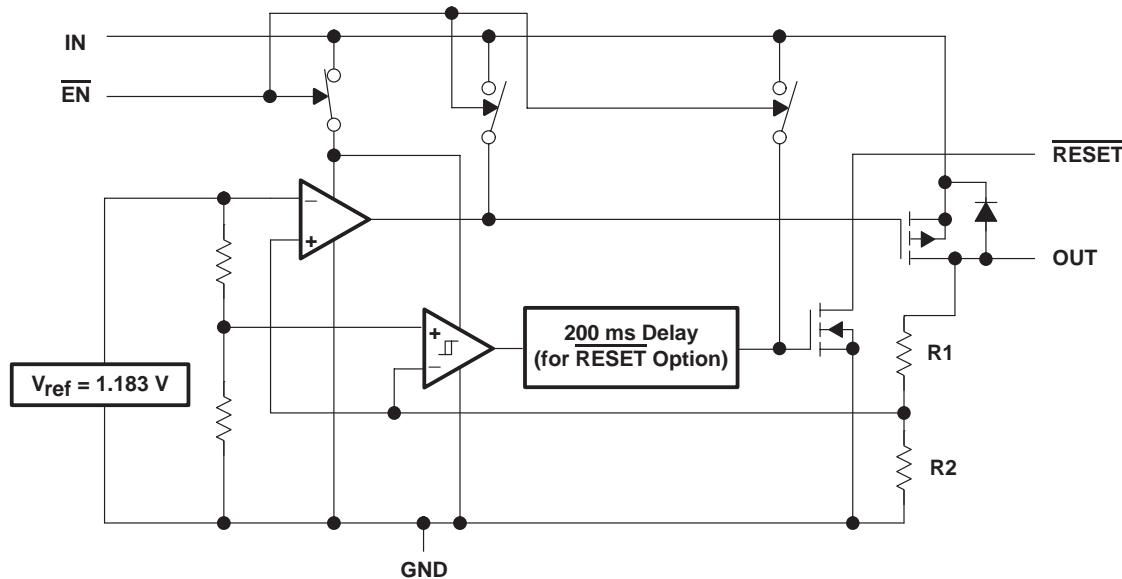
functional block diagram—adjustable version



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functional block diagram—fixed-voltage version



Terminal Functions

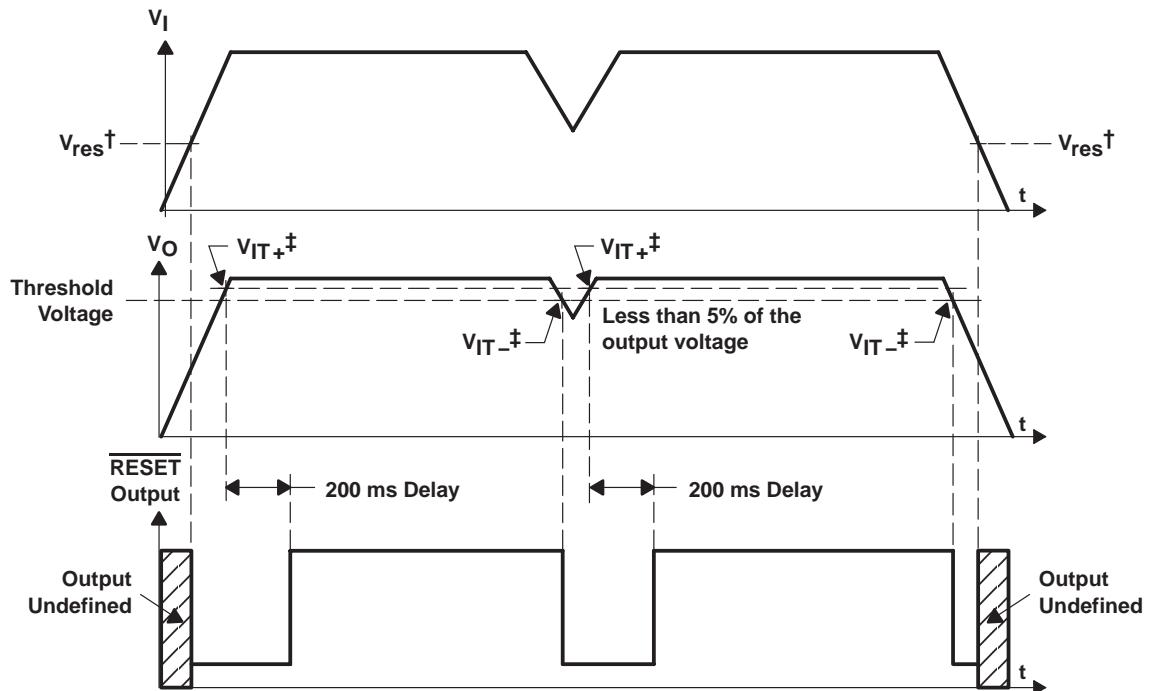
TSSOP Package

TERMINAL NAME	NO.	I/O	DESCRIPTION
EN	5	I	Enable input
FB/NC	15	I	Feedback input voltage for adjustable device (no connect for fixed options)
GND	3		Regulator ground
GND/HSINK	1, 2, 9, 10, 11, 12, 19, 20		Ground/heatsink
IN	6, 7	I	Input voltage
NC	4, 8, 17, 18		No connect
OUT	13, 14	O	Regulated output voltage
RESET	16	O	RESET output

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RESET timing diagram



† V_{res} is the minimum input voltage for a valid RESET. The symbol V_{res} is not currently listed within EIA or JEDEC standards for semiconductor symbology.

‡ V_{IT} –Trip voltage is typically 5% lower than the output voltage (95% V_O) V_{IT-} to V_{IT+} is the hysteresis voltage.

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absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†

Input voltage range‡, V_I	–0.3 V to 13.5 V
Voltage range at \bar{E}_N	–0.3 V to 16.5 V
Maximum $\bar{R}\text{ESE}\bar{T}$ voltage	16.5 V
Peak output current	Internally limited
Output voltage, V_O (OUT, FB)	7 V
Continuous total power dissipation	See dissipation rating tables
Operating virtual junction temperature range, T_J	–55°C to 125°C
Storage temperature range, T_{stg}	–65°C to 150°C
ESD rating, HBM	2 kV

† Stresses beyond those listed under “absolute maximum ratings” may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under “recommended operating conditions” is not implied.

Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

‡ All voltage values are with respect to network terminal ground.

DISSIPATION RATING TABLE – FREE-AIR TEMPERATURES

PACKAGE	AIR FLOW (CFM)	$T_A < 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 85^\circ\text{C}$ POWER RATING
PWP§	0	2.9 W	23.5 mW/°C	1.9 W	1.5 W
	300	4.3 W	34.6 mW/°C	2.8 W	2.2 W
PWP¶	0	3 W	23.8 mW/°C	1.9 W	1.5 W
	300	7.2 W	57.9 mW/°C	4.6 W	3.8 W

§ This parameter is measured with the recommended copper heat sink pattern on a 1-layer PCB, 5-in × 5-in PCB, 1 oz. copper, 2-in × 2-in coverage (4 in²).

¶ This parameter is measured with the recommended copper heat sink pattern on a 8-layer PCB, 1.5-in × 2-in PCB, 1 oz. copper with layers 1, 2, 4, 5, 7, and 8 at 5% coverage (0.9 in²) and layers 3 and 6 at 100% coverage (6 in²). For more information, refer to TI technical brief SLMA002.

recommended operating conditions

	MIN	MAX	UNIT
Input voltage, V_I #	2.7	10	V
Output voltage range, V_O	1.5	5.5	V
Output current, I_O (see Note 1)	0	500	mA
Operating virtual junction temperature, T_J (see Note 1)	–55	125	°C

To calculate the minimum input voltage for your maximum output current, use the following equation: $V_I(\text{min}) = V_O(\text{max}) + V_{DO}(\text{max load})$.

NOTE 1: Continuous current and operating junction temperature are limited by internal protection circuitry, but it is not recommended that the device operate under conditions beyond those specified in this table for extended periods of time.

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**electrical characteristics over recommended operating free-air temperature range,
 $V_I = V_O(\text{typ}) + 1 \text{ V}$, $I_O = 1 \text{ mA}$, $\text{EN} = 0 \text{ V}$, $C_O = 10 \mu\text{F}$ (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT		
Output voltage (10 μA to 500 mA load) (see Note 2)	$V_O = 1.5 \text{ V} \leq V_O \leq 5.5 \text{ V}$, $T_J = 25^\circ\text{C}$		V_O		V		
	$V_O = 1.5 \text{ V} \leq V_O \leq 5.5 \text{ V}$, $T_J = -55^\circ\text{C} \text{ to } 125^\circ\text{C}$	0.98 V_O		1.02 V_O			
	$T_J = 25^\circ\text{C}$, $2.7 \text{ V} < V_{IN} < 10 \text{ V}$		1.5		V		
		1.470		1.530			
	$T_J = 25^\circ\text{C}$, $2.7 \text{ V} < V_{IN} < 10 \text{ V}$		1.6		V		
		1.568		1.632			
	$T_J = 25^\circ\text{C}$, $2.8 \text{ V} < V_{IN} < 10 \text{ V}$		1.8		V		
		1.764		1.836			
	$T_J = 25^\circ\text{C}$, $3.5 \text{ V} < V_{IN} < 10 \text{ V}$		2.5		V		
		2.450		2.550			
	$T_J = 25^\circ\text{C}$, $4.3 \text{ V} < V_{IN} < 10 \text{ V}$		3.3		V		
		3.234		3.366			
Quiescent current (GND current) $\text{EN} = 0 \text{ V}$, (see Note 2)		$10 \mu\text{A} < I_O < 500 \text{ mA}$, $T_J = 25^\circ\text{C}$		85	μA		
		$I_O = 500 \text{ mA}$, $T_J = -55^\circ\text{C} \text{ to } 125^\circ\text{C}$		125			
Output voltage line regulation ($\Delta V_O / V_O$) (see Notes 2 and 3)		$V_O + 1 \text{ V} < V_I \leq 10 \text{ V}$, $T_J = 25^\circ\text{C}$		0.01	%/V		
Load regulation				3	mV		
Output noise voltage (TPS77518)		$BW = 200 \text{ Hz to } 100 \text{ kHz}$, $I_C = 500 \text{ mA}$ $C_O = 10 \mu\text{F}$, $T_J = 25^\circ\text{C}$		53	μVRms		
Output current limit		$V_O = 0 \text{ V}$		1.7	2.4	A	
Thermal shutdown junction temperature				150	$^\circ\text{C}$		
Standby current		$\overline{\text{EN}} = V_I$, $T_J = 25^\circ\text{C}$, $2.7 \text{ V} < V_I < 10 \text{ V}$		1	μA		
		$\overline{\text{EN}} = V_I$, $T_J = -55^\circ\text{C} \text{ to } 125^\circ\text{C}$ $2.7 \text{ V} < V_I < 10 \text{ V}$		10	μA		
FB input current	TPS77501	$FB = 1.5 \text{ V}$		2	nA		
High level enable input voltage				1.7	V		
Low level enable input voltage				0.9	V		
Power supply ripple rejection (see Note 2)		$f = 1 \text{ KHz}$, $C_O = 10 \mu\text{F}$, $T_J = 25^\circ\text{C}$		60	dB		

NOTES: 2. Minimum IN operating voltage is 2.7 V or $V_O(\text{typ}) + 1 \text{ V}$, whichever is greater. Maximum IN voltage 10V.
3. If $V_O \leq 1.8 \text{ V}$ then $V_{I\min} = 2.7 \text{ V}$, $V_{I\max} = 10 \text{ V}$:

$$\text{Line Reg. (mV)} = (\%/\text{V}) \times \frac{V_O(V_{I\max} - 2.7 \text{ V})}{100} \times 1000$$

If $V_O \geq 2.5 \text{ V}$ then $V_{I\min} = V_O + 1 \text{ V}$, $V_{I\max} = 10 \text{ V}$:

$$\text{Line Reg. (mV)} = (\%/\text{V}) \times \frac{V_O(V_{I\max} - (V_O + 1 \text{ V}))}{100} \times 1000$$

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**electrical characteristics over recommended operating free-air temperature range,
 $V_I = V_O(\text{typ}) + 1 \text{ V}$, $I_O = 1 \text{ mA}$, $\text{EN} = 0 \text{ V}$, $C_o = 10 \mu\text{F}$ (unless otherwise noted) (continued)**

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
Reset	Minimum input voltage for valid RESET	$I_O(\text{RESET}) = 300 \mu\text{A}$		1.1		V
	Trip threshold voltage	V_O decreasing	92	98		% V_O
	Hysteresis voltage	Measured at V_O		0.5		% V_O
	Output low voltage	$V_I = 2.7 \text{ V}$, $I_O(\text{RESET}) = 1 \text{ mA}$	0.15	0.4		V
	Leakage current	$V(\text{RESET}) = 5 \text{ V}$		1		μA
	RESET time-out delay		200			ms
Input current ($\overline{\text{EN}}$)		$\overline{\text{EN}} = 0 \text{ V}$	-1	0	1	μA
		$\overline{\text{EN}} = V_I$	-1		1	
Dropout voltage (see Note 4)	TPS77533	$I_O = 500 \text{ mA}$, $T_J = 25^\circ\text{C}$		169		mV
		$I_O = 500 \text{ mA}$, $T_J = -55^\circ\text{C}$ to 125°C			287	

NOTE 4: IN voltage equals $V_O(\text{typ}) - 100 \text{ mV}$; TPS77515, TPS77516, TPS77518, and TPS77525 dropout voltage limited by input voltage range limitations (i.e., TPS77533 input voltage needs to drop to 3.2 V for purpose of this test).

ELECTRICAL CHARACTERISTICS

Table of Graphs

		FIGURE
V_O	Output voltage	vs Output current
		vs Free-air temperature
Ground current		vs Free-air temperature
Power supply ripple rejection		vs Frequency
Output spectral noise density		vs Frequency
Z_O	Output impedance	vs Frequency
V_{DO}	Dropout voltage	vs Input voltage
		vs Free-air temperature
Input voltage (min)		vs Output voltage
Line transient response		
Load transient response		
V_O	Output voltage	vs Time
	Equivalent series resistance (ESR)	vs Output current

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TYPICAL CHARACTERISTICS

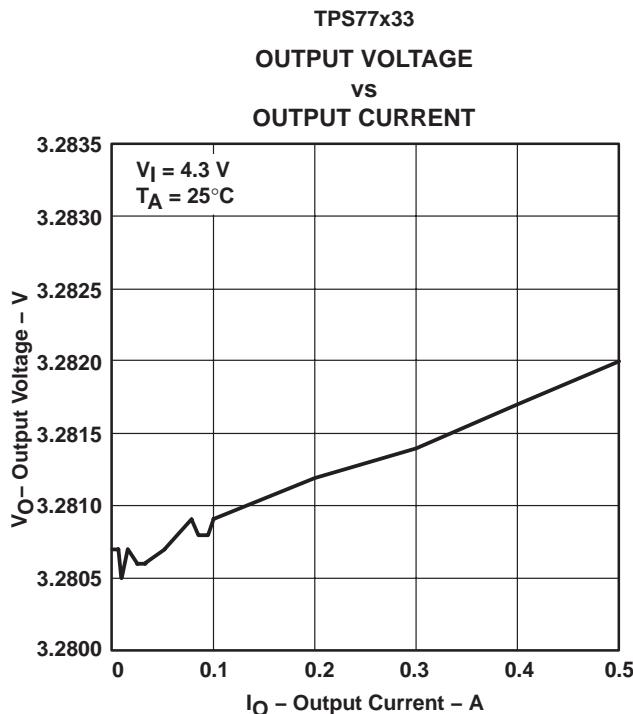


Figure 2

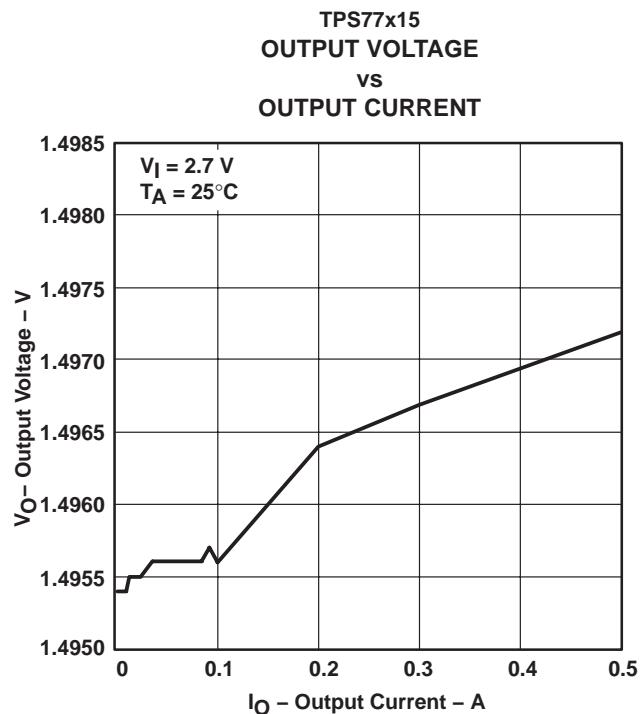


Figure 3

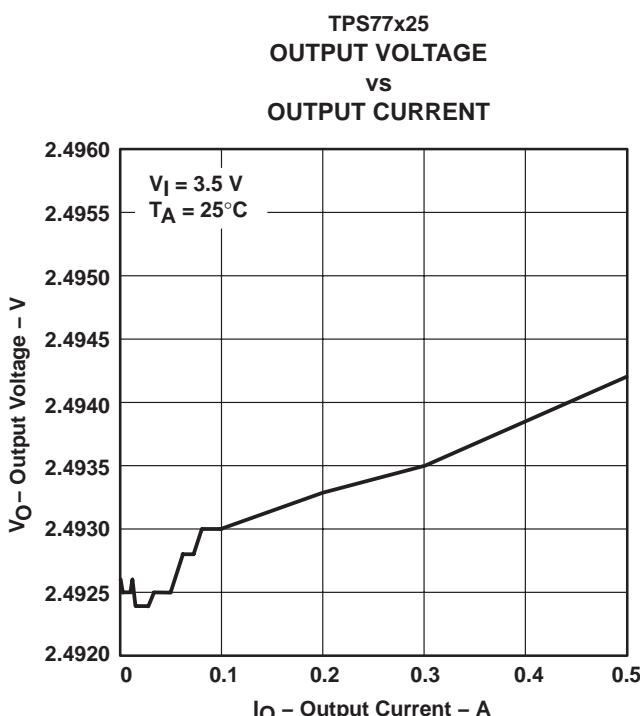


Figure 4

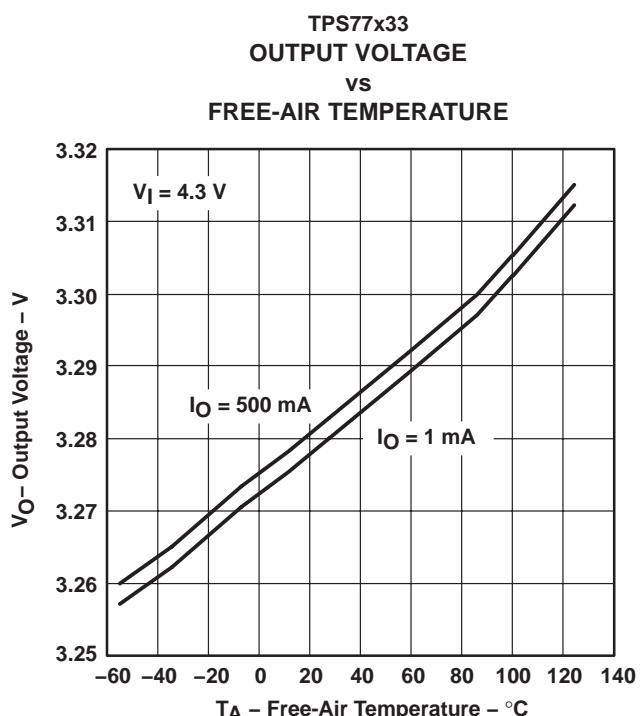


Figure 5

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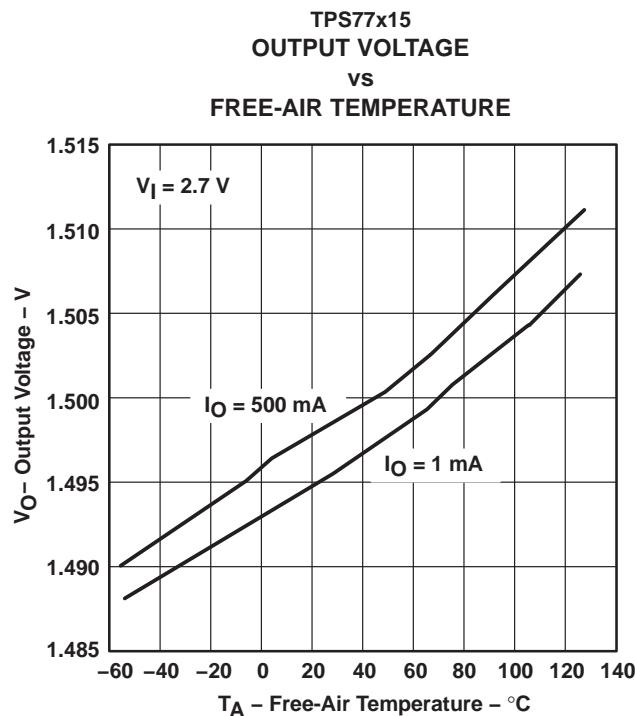


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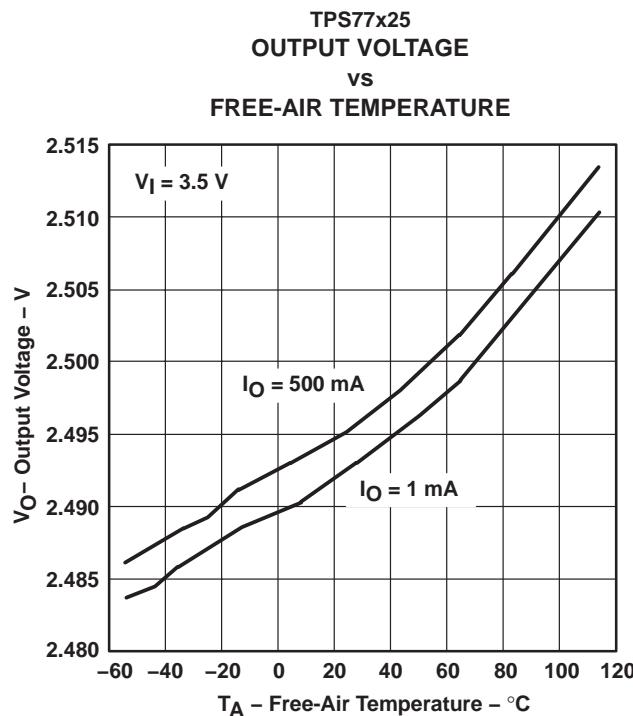


Figure 7

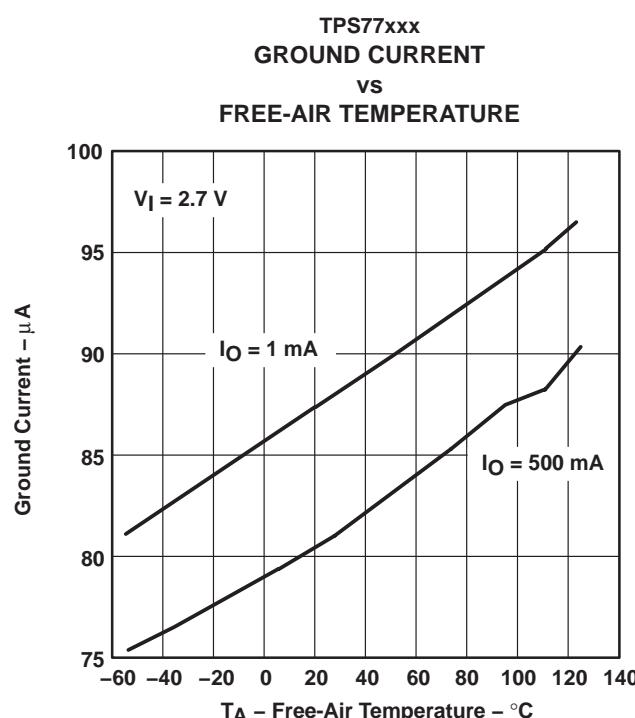


Figure 8

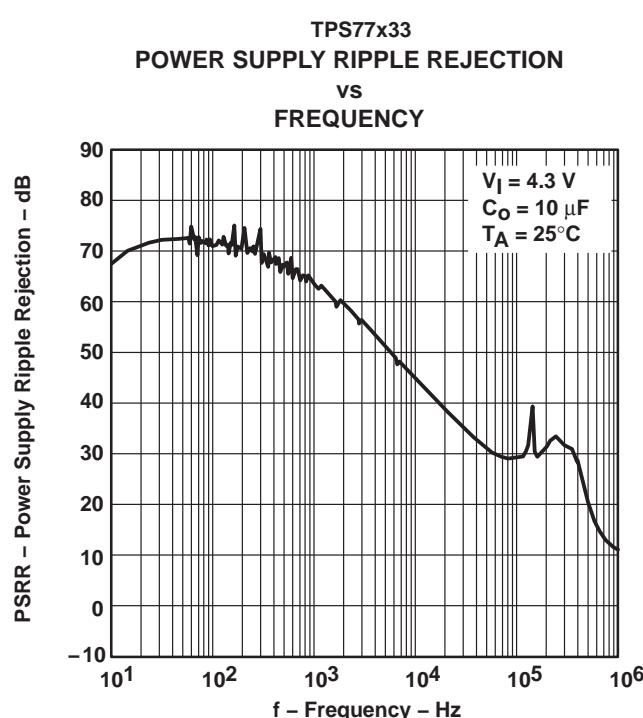
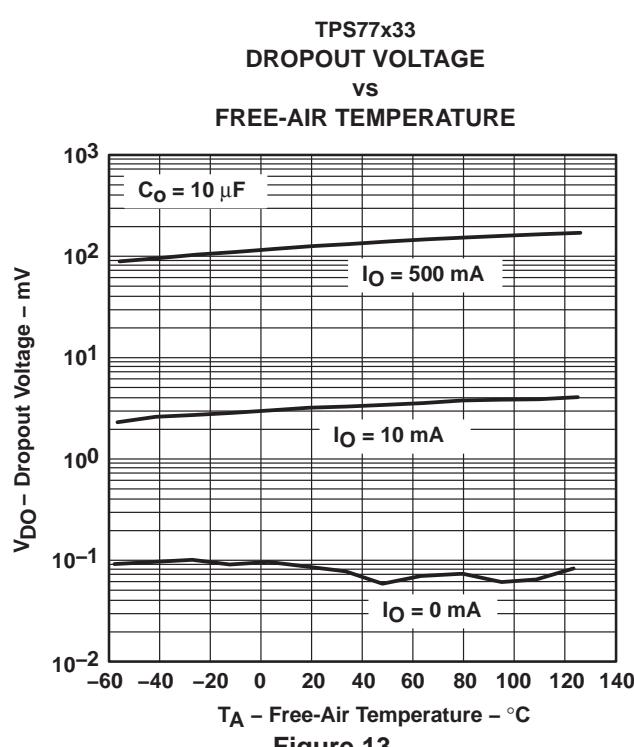
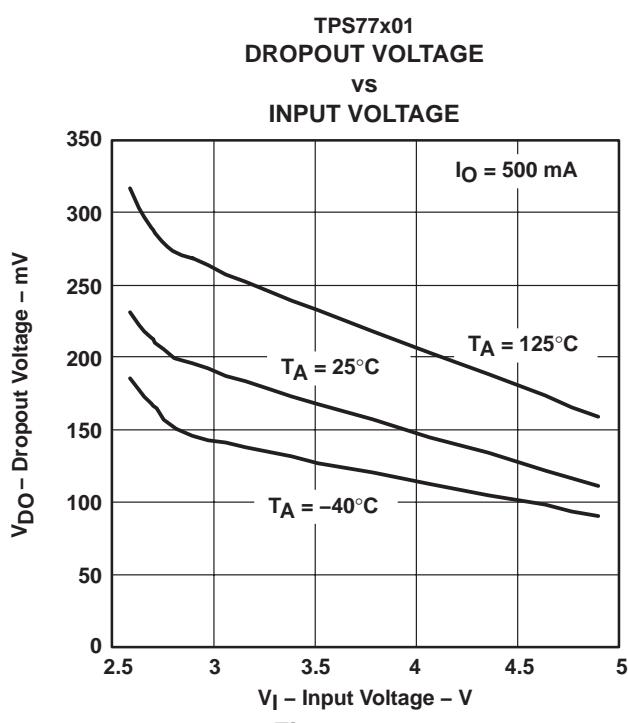
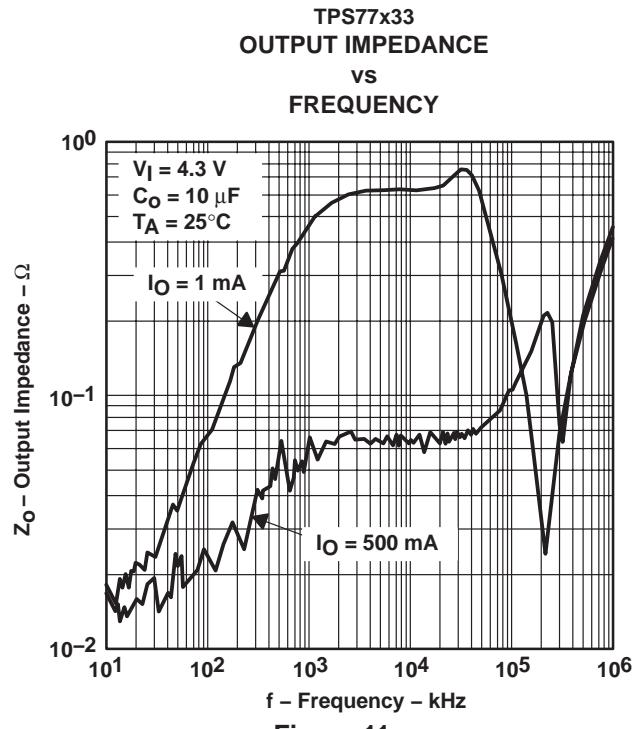
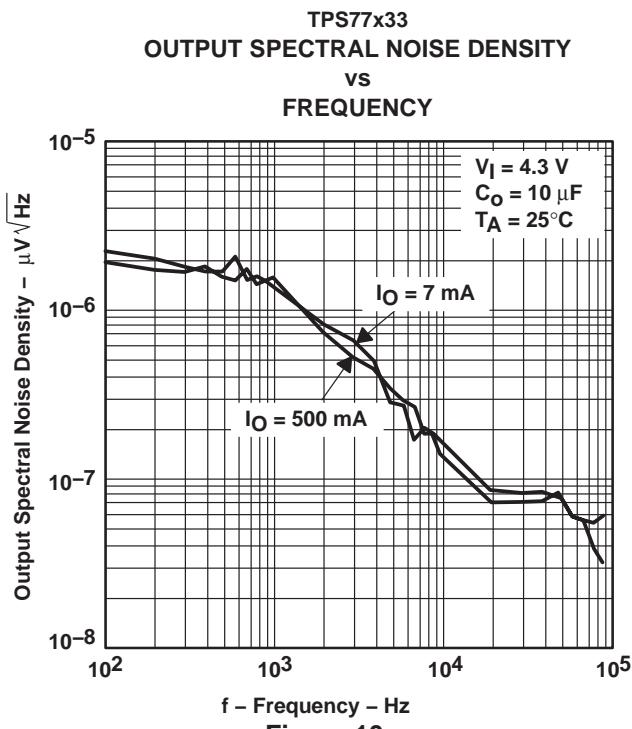


Figure 9

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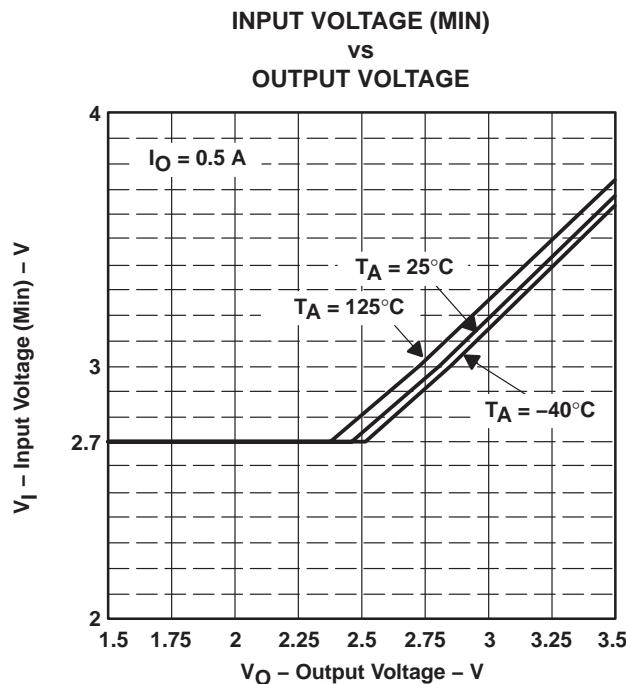


Figure 14

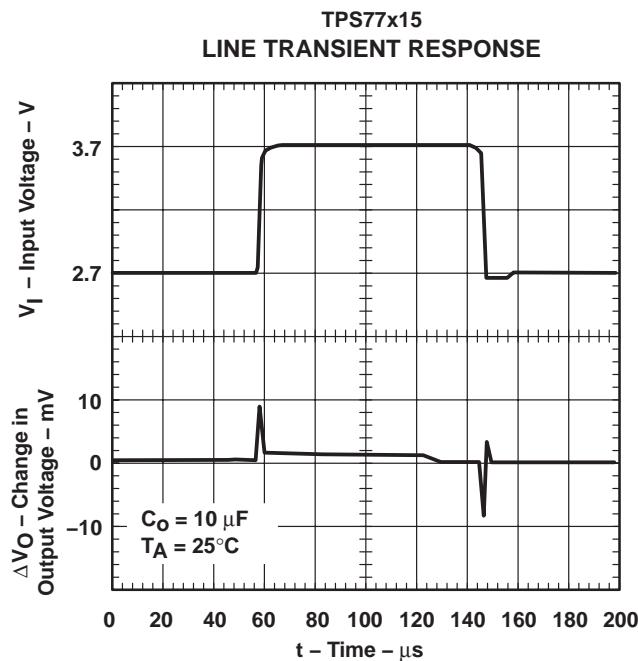


Figure 15

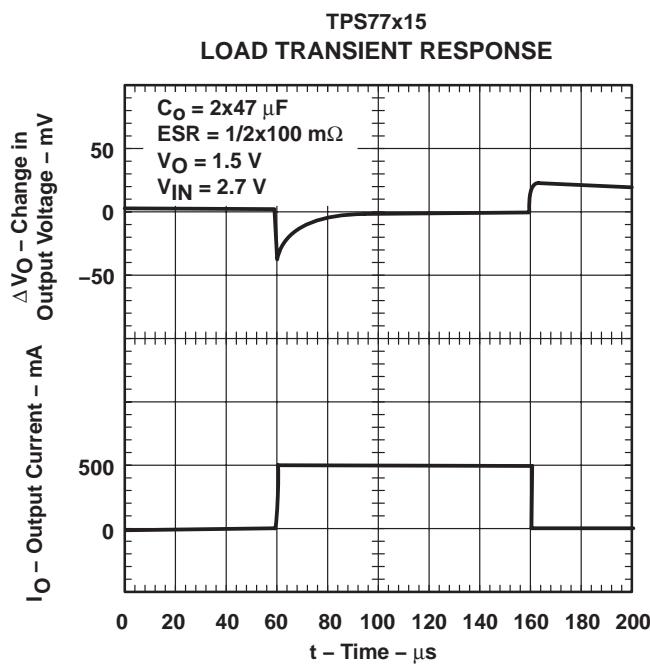


Figure 16

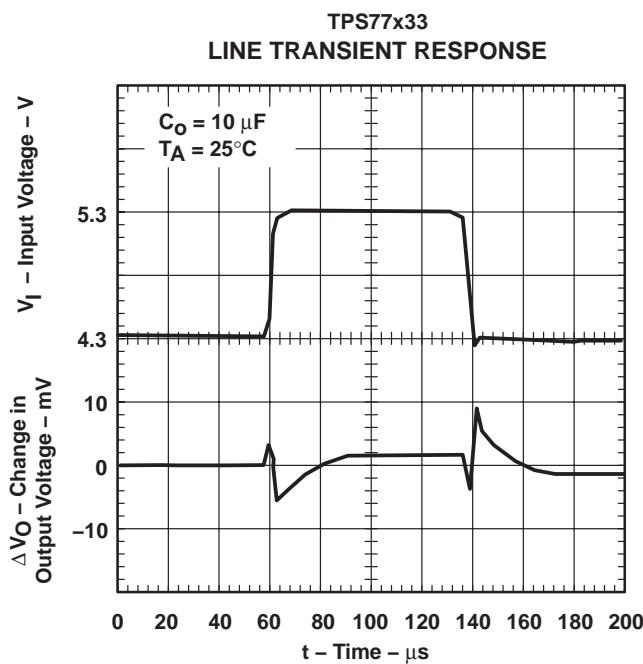


Figure 17

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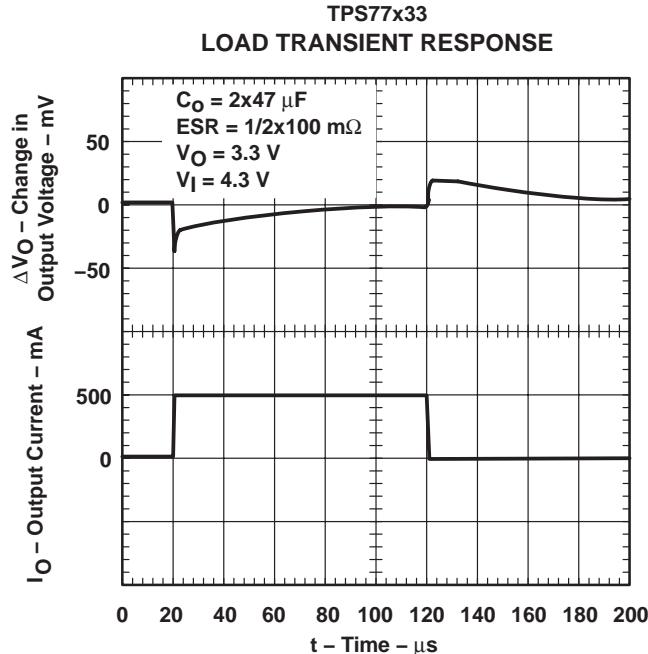


Figure 18

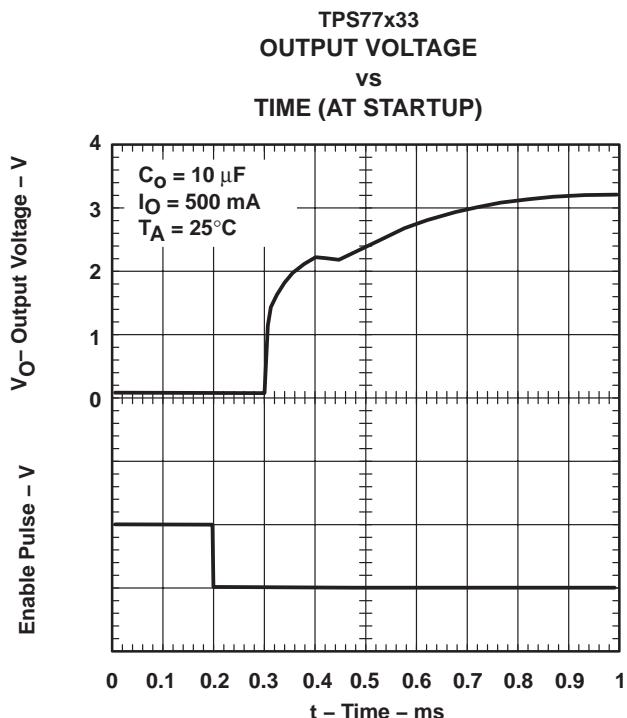


Figure 19

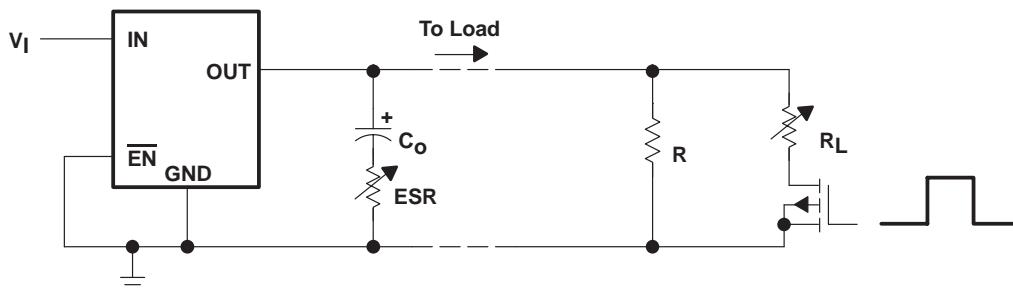


Figure 20. Test Circuit for Typical Regions of Stability (Figures 21 through 24) (Fixed Output Options)

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TYPICAL CHARACTERISTICS

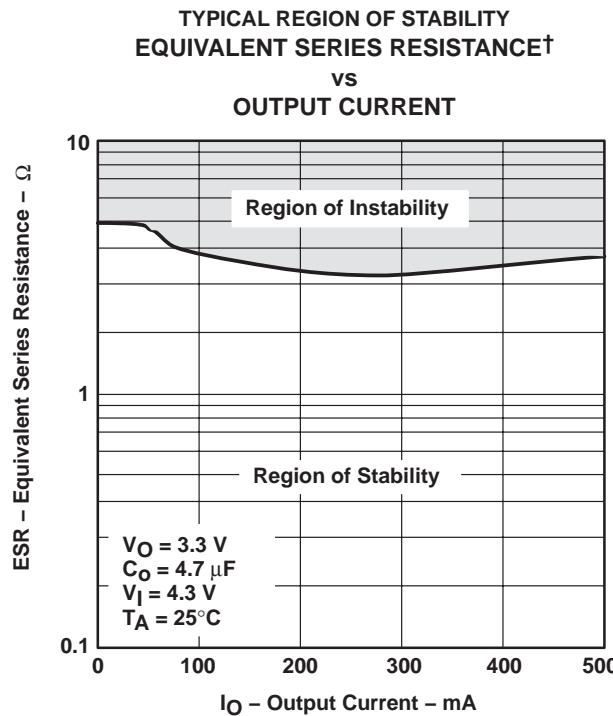


Figure 21

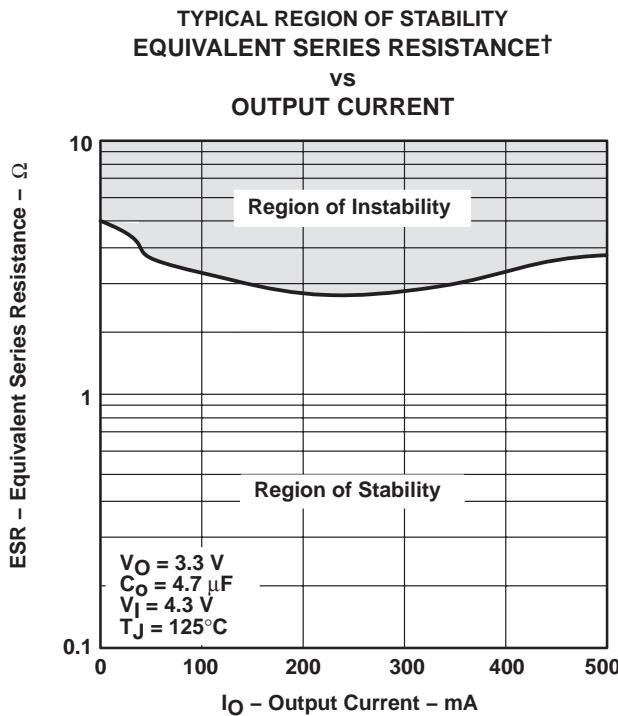


Figure 22

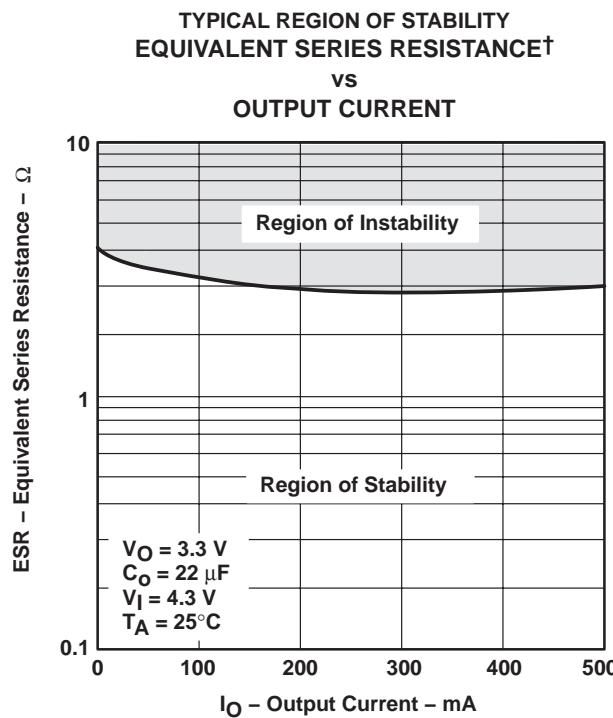


Figure 23

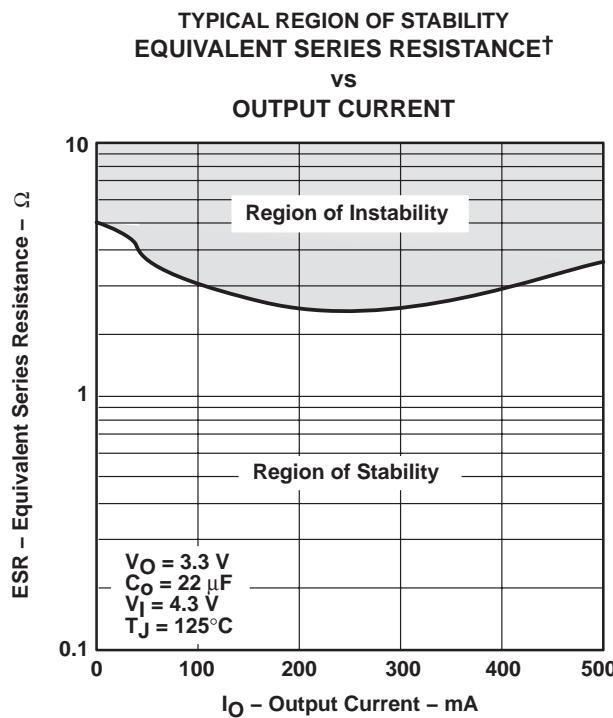


Figure 24

[†] Equivalent series resistance (ESR) refers to the total series resistance, including the ESR of the capacitor, any series resistance added externally, and PWB trace resistance to C_O.

**TPS77501-EP, TPS77515-EP, TPS77516-EP, TPS77518-EP
TPS77525-EP, TPS77533-EP WITH RESET OUTPUT
FAST-TRANSIENT-RESPONSE 500-mA LOW-DROPOUT VOLTAGE REGULATORS**

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APPLICATION INFORMATION

The TPS775xx family includes five fixed-output voltage regulators (1.5 V, 1.6 V, 1.8 V, 2.5 V, and 3.3 V), and an adjustable regulator, the TPS77501 (adjustable from 1.5 V to 5.5 V).

device operation

The TPS775xx feature very low quiescent current, which remains virtually constant even with varying loads. Conventional LDO regulators use a pnp pass element, the base current of which is directly proportional to the load current through the regulator ($I_B = I_C/\beta$). The TPS775xx use a PMOS transistor to pass current; because the gate of the PMOS is voltage driven, operating current is low and invariable over the full load range.

Another pitfall associated with the pnp-pass element is its tendency to saturate when the device goes into dropout. The resulting drop in β forces an increase in I_B to maintain the load. During power up, this translates to large start-up currents. Systems with limited supply current may fail to start up. In battery-powered systems, it means rapid battery discharge when the voltage decays below the minimum required for regulation. The TPS775xx quiescent currents remain low even when the regulator drops out, eliminating both problems.

The TPS775xx family also features a shutdown mode that places the output in the high-impedance state (essentially equal to the feedback-divider resistance) and reduces quiescent current to 2 μ A. If the shutdown feature is not used, EN should be tied to ground.

minimum load requirements

The TPS775xx family is stable even at zero load; no minimum load is required for operation.

FB—pin connection (adjustable version only)

The FB pin is an input pin to sense the output voltage and close the loop for the adjustable option . The output voltage is sensed through a resistor divider network to close the loop as it is shown in Figure 26. Normally, this connection should be as short as possible; however, the connection can be made near a critical circuit to improve performance at that point. Internally, FB connects to a high-impedance wide-bandwidth amplifier and noise pickup feeds through to the regulator output. Routing the FB connection to minimize/avoid noise pickup is essential.

external capacitor requirements

An input capacitor is not usually required; however, a ceramic bypass capacitor (0.047 μ F or larger) improves load transient response and noise rejection if the TPS775xx is located more than a few inches from the power supply. A higher-capacitance electrolytic capacitor may be necessary if large (hundreds of millamps) load transients with fast rise times are anticipated.

Like all low dropout regulators, the TPS775xx require an output capacitor connected between OUT and GND to stabilize the internal control loop. The minimum recommended capacitance value is 10 μ F and the ESR (equivalent series resistance) must be between 50 m Ω and 1.5 Ω . Capacitor values 10 μ F or larger are acceptable, provided the ESR is less than 1.5 Ω . Solid tantalum electrolytic, aluminum electrolytic, and multilayer ceramic capacitors are all suitable, provided they meet the requirements described previously.

**TPS77501-EP, TPS77515-EP, TPS77516-EP, TPS77518-EP
TPS77525-EP, TPS77533-EP WITH RESET OUTPUT
FAST-TRANSIENT-RESPONSE 500-mA LOW-DROPOUT VOLTAGE REGULATORS**

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APPLICATION INFORMATION

external capacitor requirements (continued)

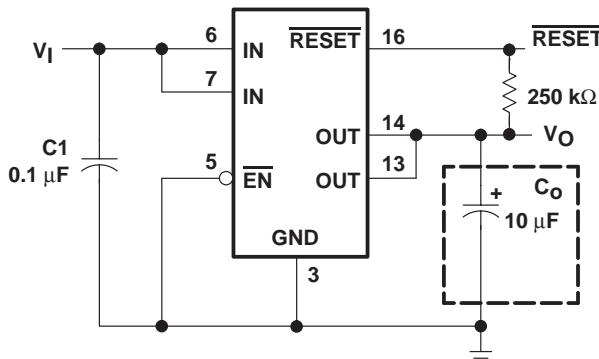


Figure 25. Typical Application Circuit (Fixed Versions)

programming the TPS77501 adjustable LDO regulator

The output voltage of the TPS77501 adjustable regulator is programmed using an external resistor divider as shown in Figure 26. The output voltage is calculated using:

$$V_O = V_{\text{ref}} \times \left(1 + \frac{R_1}{R_2} \right) \quad (1)$$

Where:

$V_{\text{ref}} = 1.1834 \text{ V typ}$ (the internal reference voltage)

Resistors R1 and R2 should be chosen for approximately 10-μA divider current. Lower value resistors can be used but offer no inherent advantage and waste more power. Higher values should be avoided as leakage currents at FB increase the output voltage error. The recommended design procedure is to choose $R_2 = 110 \text{ k}\Omega$ to set the divider current at approximately 10 μA and then calculate R1 using:

$$R_1 = \left(\frac{V_O}{V_{\text{ref}}} - 1 \right) \times R_2 \quad (2)$$

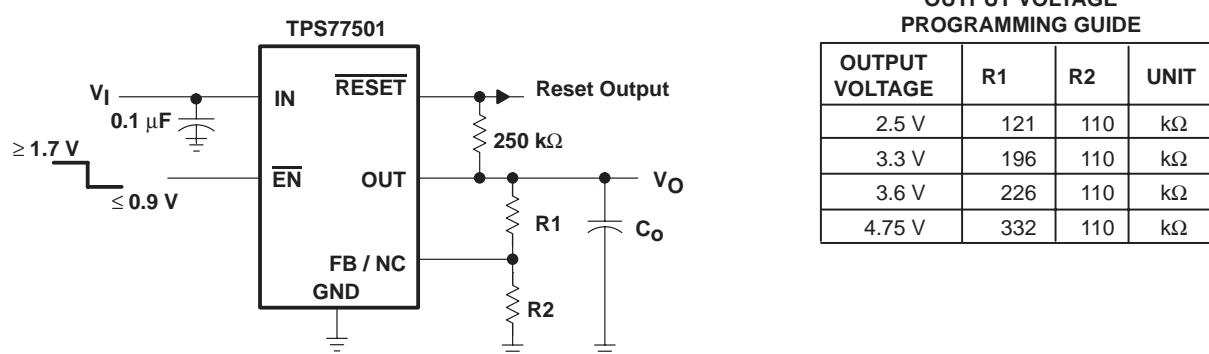


Figure 26. TPS77501 Adjustable LDO Regulator Programming

**TPS77501-EP, TPS77515-EP, TPS77516-EP, TPS77518-EP
TPS77525-EP, TPS77533-EP WITH RESET OUTPUT
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APPLICATION INFORMATION

reset indicator

The TPS775xx features a RESET output that can be used to monitor the status of the regulator. The internal comparator monitors the output voltage: when the output drops to between 92% and 98% of its nominal regulated value, the RESET output transistor turns on, taking the signal low. The open-drain output requires a pullup resistor. If not used, it can be left floating. RESET can be used to drive power-on reset circuitry or as a low-battery indicator. RESET does not assert itself when the regulated output voltage falls outside the specified 2% tolerance, but instead reports an output voltage low relative to its nominal regulated value (refer to timing diagram for start-up sequence).

regulator protection

The TPS775xx PMOS-pass transistors have a built-in back diode that conducts reverse currents when the input voltage drops below the output voltage (e.g., during power down). Current is conducted from the output to the input and is not internally limited. When extended reverse voltage is anticipated, external limiting may be appropriate.

The TPS775xx also feature internal current limiting and thermal protection. During normal operation, the TPS775xx limit output current to approximately 1.7 A. When current limiting engages, the output voltage scales back linearly until the overcurrent condition ends. While current limiting is designed to prevent gross device failure, care should be taken not to exceed the power dissipation ratings of the package. If the temperature of the device exceeds 150°C(typ), thermal-protection circuitry shuts it down. Once the device has cooled below 130°C(typ), regulator operation resumes.

power dissipation and junction temperature

Specified regulator operation is assured to a junction temperature of 125°C; the maximum junction temperature should be restricted to 125°C under normal operating conditions. This restriction limits the power dissipation the regulator can handle in any given application. To ensure the junction temperature is within acceptable limits, calculate the maximum allowable dissipation, $P_{D(\max)}$, and the actual dissipation, P_D , which must be less than or equal to $P_{D(\max)}$.

The maximum-power-dissipation limit is determined using the following equation:

$$P_{D(\max)} = \frac{T_{J\max} - T_A}{R_{\theta JA}}$$

Where:

$T_{J\max}$ is the maximum allowable junction temperature.

$R_{\theta JA}$ is the thermal resistance junction-to-ambient for the package, i.e., 32.6°C/W for the 20-terminal PWP with no airflow.

T_A is the ambient temperature.

The regulator dissipation is calculated using:

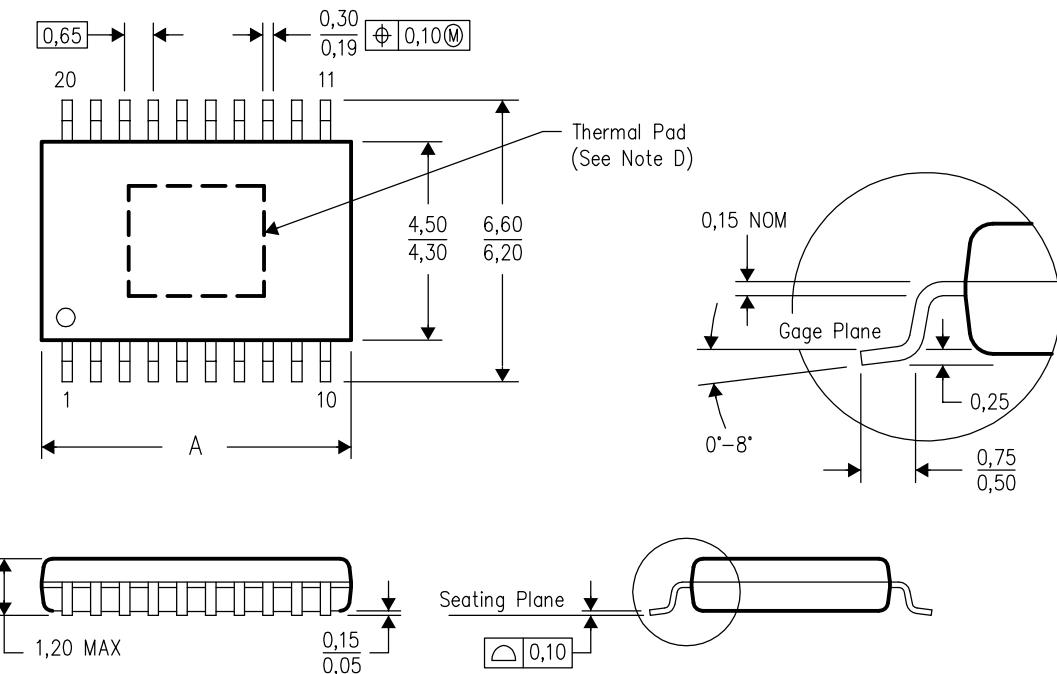
$$P_D = (V_I - V_O) \times I_O$$

Power dissipation resulting from quiescent current is negligible. Excessive power dissipation will trigger the thermal protection circuit.

PWP (R-PDSO-G**)

20 PIN SHOWN

PowerPAD™ PLASTIC SMALL-OUTLINE PACKAGE



DIM \ PINS **	14	16	20	24	28
A MAX	5,10	5,10	6,60	7,90	9,80
A MIN	4,90	4,90	6,40	7,70	9,60

4073225/G 08/03

- NOTES:
- A. All linear dimensions are in millimeters.
 - B. This drawing is subject to change without notice.
 - C. Body dimensions do not include mold flash or protrusions.
 - D. This package is designed to be soldered to a thermal pad on the board. Refer to Technical Brief, PowerPad Thermally Enhanced Package, Texas Instruments Literature No. SLMA002 for information regarding recommended board layout. This document is available at www.ti.com <<http://www.ti.com>>.
 - E. Falls within JEDEC MO-153

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Mailing Address: Texas Instruments
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