VERSACLOCK® LOW POWER CLOCK GENERATOR   IDT5P49EE502

Description

The IDT5P49EE502 is a programmable clock generator intended for low power, battery operated consumer applications. There are four internal PLLs, each individually programmable, allowing for up to five differrent output frequencies. The frequencies are generated from a single reference clock. The reference clock can come from either a TCXO or fundamental mode crystal.

The IDT5P49EE502 can be programmed through the use of the I^{2}C interfaces. The programming interface enables the device to be programmed when it is in normal operation or what is commonly known as in system programmable. An internal EEPROM allows the user to save and restore the configuration of the device without having to reprogram it on power-up.

Each of the four PLLs has an 8-bit reference divider and a 11-bit feedback divider. This allows the user to generate four unique non-integer-related frequencies. The PLL loop bandwidth is programmable to allow the user to tailor the PLL response to the application. For instance, the user can tune the PLL parameters to minimize jitter generation or to maximize jitter attenuation. Spread spectrum generation is supported on one of the PLLs.

Spread spectrum generation is supported on one of the PLLs. The device is specifically designed to work with display applications to ensure that the spread profile remains consistent for each HSYNC in order to reduce ROW noise. It also may operate in standard spread sepctrum mode.

There are total four 8-bit output dividers. The outputs are connected to the PLLs via the switch matrix. The switch matrix allows the user to route the PLL outputs to any output bank. This feature can be used to simplify and optimize the board layout. In addition, each output's slew rate and enable/disable function can be programmed.

Target Applications

- Smart Mobile Handset
- Personal Navigation Device (PND)
- Camcorder
- DSC
- Portable Game Console
- Personal Media Player

Features

- Four internal PLLs
- Internal non-volatile EEPROM
  - Internal I^{2}C EEPROM master interface
- FAST (400kHz) mode I^{2}C serial interfaces
- Input Frequencies
  - TCXO: 10 MHz to 40 MHz
  - Crystal: 8 MHz to 30 MHz
- Output Frequency Ranges: kHz to 120 MHz
- Each PLL has an 8-bit reference divider and a 11-bit feedback-divider
- 8-bit output-divider blocks
- One of the PLLs support Spread Spectrum generation capable of configuration to pixel rate, with adjustable modulation rate and amplitude to support video clock with no visible artifacts
- I/O Standards:
  - Outputs - 1.8V/2.5V/3.3 V LVTTL/ LVCMOS
- 2 independent adjustable VDDO groups.
- Programmable Slew Rate Control
- Programmable Loop Bandwidth Settings
- Programmable output inversion to reduce bimodal jitter
- Individual output enable/disable
- Power-down/Sleep mode
  - 10μA max in power down mode
- 1.8V VDD Core Voltage
- Available in 20pin 3x3mm QFN packages
- -40 to +85 C Industrial Temp operation

# Functional Block Diagram

Pin Assignment

20-pin QFN
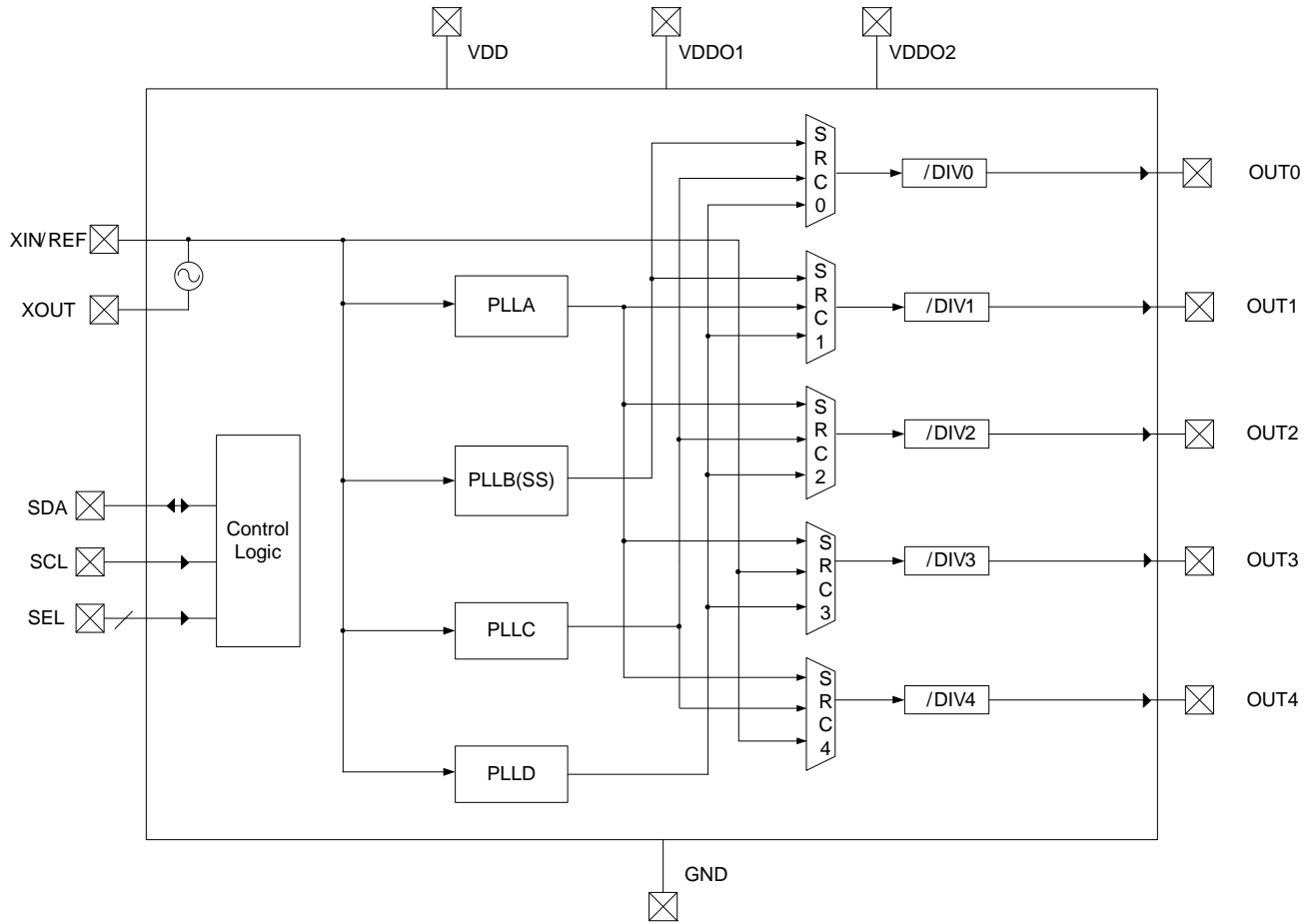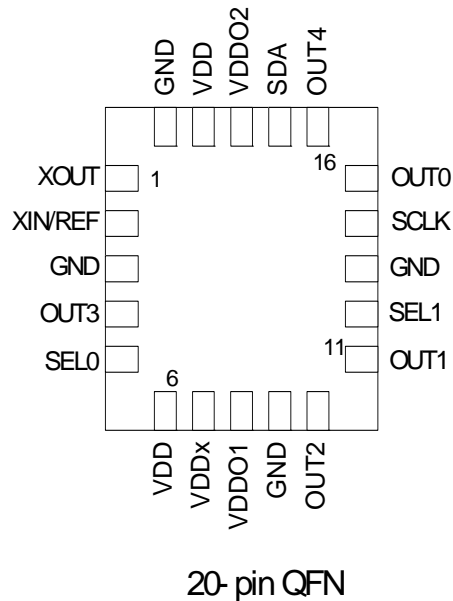
Pin Descriptions

Pin Name	Pin #	I/O	Pin Type	Pin Description
XOUT	1	O	LVTTL	MHz CRYSTAL_OUT -- Reference crystal feedback.
XIN/ REF	2	I	LVTTL	MHz CRYSTAL_IN -- Reference crystal input or external reference clock input.
GND	3		Power	Connect to Ground.
OUT3	4	O	OUTPUT	Buffered reference clock output. Single-ended output voltage levels are register controlled by either VDDO1 or VDDO2.
SEL0	5	I	LVTTL	Configuration select pin. Weak internal pull down resistor.
VDD	6		Power	Device power supply. Connect to 1.8V.
VDDx	7		Power	Device power supply. Connect to 1.8V.
VDDO1	8		Power	Device power supply. Connect to 1.8 to 3.3V. VDDO1 must be the highest voltage on the device. Using register settings, select output voltage levels for OUT0-OUT3.
GND	9		Power	Connect to Ground.
OUT2	10	O	Adjustable	Configurable clock output 2. Single-ended output voltage levels are register controlled by either VDDO1 or VDDO2.
OUT1	11	O	Adjustable	Configurable clock output 1. Single-ended output voltage levels are register controlled by either VDDO1 or VDDO2.
SEL1	12	I	LVTTL	Configuration select pin. Weak internal pull down resistor.
GND	13		Power	Connect to Ground.
SCLK	14	I	LVTTL	I²C clock.

OUT0	15	O	Adjustable	Configurable clock output 0. Single-ended output voltage levels are register controlled by either VDDO1 or VDDO2.
OUT4	16	O	Adjustable	Configurable clock output 8. Single-ended output voltage levels controlled by VDDO2.
SDA	17	I/O	Open Drain	Bidirectional I^{2}C data.
VDDO2	18		Power	Device power supply. Connect to 1.8 to 3.3V. Using register settings, select output voltage levels for OUT0-OUT4.
VDD	19		Power	Device power supply. Connect to 1.8V.
GND	20		Power	Connect to Ground.

Note 1: Outputs are user programmable to drive single-ended 1.8V/2.5V/3.3V LVTTL as indicated above. Alway completely power up VDD and VDDx prior to applying VDDO power.

Note 2: Default configuration CLK3=Buffered Reference output. All other outputs are off.

PLL Features and Descriptions
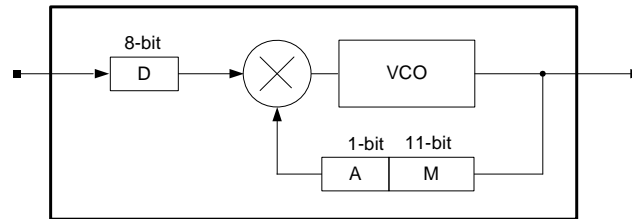
PLL Block Diagram

	Ref-Divider (D) Values	Feedback Pre-Divider (XDIV) Values	Feedback (M) Values	Programmable Loop Bandwidth	Spread Spectrum Generation Capability
PLLA	1 - 255	1 or 4[1]	6 - 2047	Yes	No
PLLB	1 - 255	4	6 - 2047	Yes	Yes
PLLC	1 - 255	1 or 8 bit divide[2]	6 - 2047	Yes	No
PLLD	1 - 255	1 or 4[1]	6 - 2047	Yes	No

1.XDIVA or XDIVD=0, A=1. XDIVA or XDIVD=1, A=4.
2.XDIVC =0, A=1. XDIVC=1 turns on 8 bit predivide multiplier, A=FBC2[7:0]. Total feedback divide equals FBC[10:0] *FBC2[7:0].

Crystal Input (XIN/REF)

The crystal oscillators should be fundamental mode quartz crystals; overtone crystals are not suitable. Crystal frequency should be specified for parallel resonance with 50Ω maximum equivalent series resonance.

Crystal Load Capacitors

The device crystal connections should include pads for small capacitors from X1 to ground and from X2 to ground. These capacitors are used to adjust the stray capacitance of the board to match the nominally required crystal load capacitance. Because load capacitance can only be increased in this trimming process, it is important to keep stray capacitance to a minimum by using very short PCB traces (and no vias) between the crystal and device. Crystal capacitors must be connected from each of the pins X1 and X2 to ground.

The crystal cpacitors are internal to the device and have an effective value of 8pF.

Reference Pre-Divider, Reference Divider, Feedback-Divider and Post-Divider

Each PLL incorporates an 8-bit reference-scaler and a 11-bit feedback divider which allows the user to generate four unique non-integer-related frequencies. PLLA and PLLD each have a feedback pre-divider that provides additional multiplication for kHz reference clock applications. Each output divider supports 8-bit post-divider. The following equation governs how the output frequency is calculated.

$$F_{OUT} = \frac{F_{IN} * \left(\frac{A*M}{D} \right)}{ODIV} \quad \text{(Eq. 2)}$$

Where F_{IN} is the reference frequency, A is the feedback pre-divider value, M is the feedback-divider value, D is the reference divider value, ODIV is the total post-divider value, and F_{OUT} is the resulting output frequency. Programming any of the dividers may cause glitches on the outputs.

SPREAD SPECTRUM GENERATION (PLLB)

PLLB has spread spectrum generation capability, which users have the option of turning on and off. Spread spectrum profile, frequency, and spread are fully programmable (within limits). The programmable spread spectrum generation parameters are NC[10:0], MOD[12:0], and NSS[10:0] bits. To enable spread spectrum, set SSENB_B=0.

The spread spectrum circuitry was specifically developed to accommodate video display applications. The spread modulation frequency can be defined to exactly equal the horizontal line frequency (HSYNC)

NC[10:0]

These bits are used to determine the number of pulses per spread spectrum cycle. For video applications, NC is the number of pixels on the horizontal display row (or integer multiple of displayed pixels in a row). By matching the spread period to the screen, no tearing or "shimmer" will be apparent.

NC must be an even number to insure that the upward spread transition has the same number of steps as the downward spread transition.

For non-video applications, this can also be seen as the number of clock cycles for a complete spread spectrum period.

MOD[12:0]

These bits relate the VCO frequency to the target average spread output frequency (F_{MID}).

$$F_{MID} = (F_{VCO}) / 8$$

$$F_{MAX} = F_{MID} + (SS\% * F_{MID})$$

$$F_{MIN} = F_{MID} - (SS\% * F_{MID})$$

$$MOD = (F_{REF} * NC) / (2 * F_{MID})$$

NSS[10:0]

These bits control the amplitude of the spread modulation.

$$NSS = (NC / 2) + (NC / 8) * (F_{MAX} - F_{MIN}) / F_{MID}$$

Modulation frequency:

$$F_{MOD} = F_{MID} / NC \text{ (Eq. 11)}$$

Video Example

F_{REF} = 27 MHz, F_{OUT} = 27 MHz, 640 pixels per line, center spread of ±1%. Using F_{VCO}=432MHz, find the necessary spread spectrum register settings.

$$F_{MID} = F_{VCO}/8$$

NC = 640 (integer number of spread periods/screen)

MOD = (25MHz * 640)/(2 * 54MHz) = 160

NSS = (640/2)+(640/8)*(27.27MHz-26.73MHz)/27MHz = 321.

$$F_{MOD} = 27MHz/640 = 11.8kHz.$$

Non-Video Example

F_{REF} = 25MHz, F_{OUT} = 27 MHz, 31.25kHz modulation rate, center spread of ±1%. Find the necessary spread spectrum register settings.
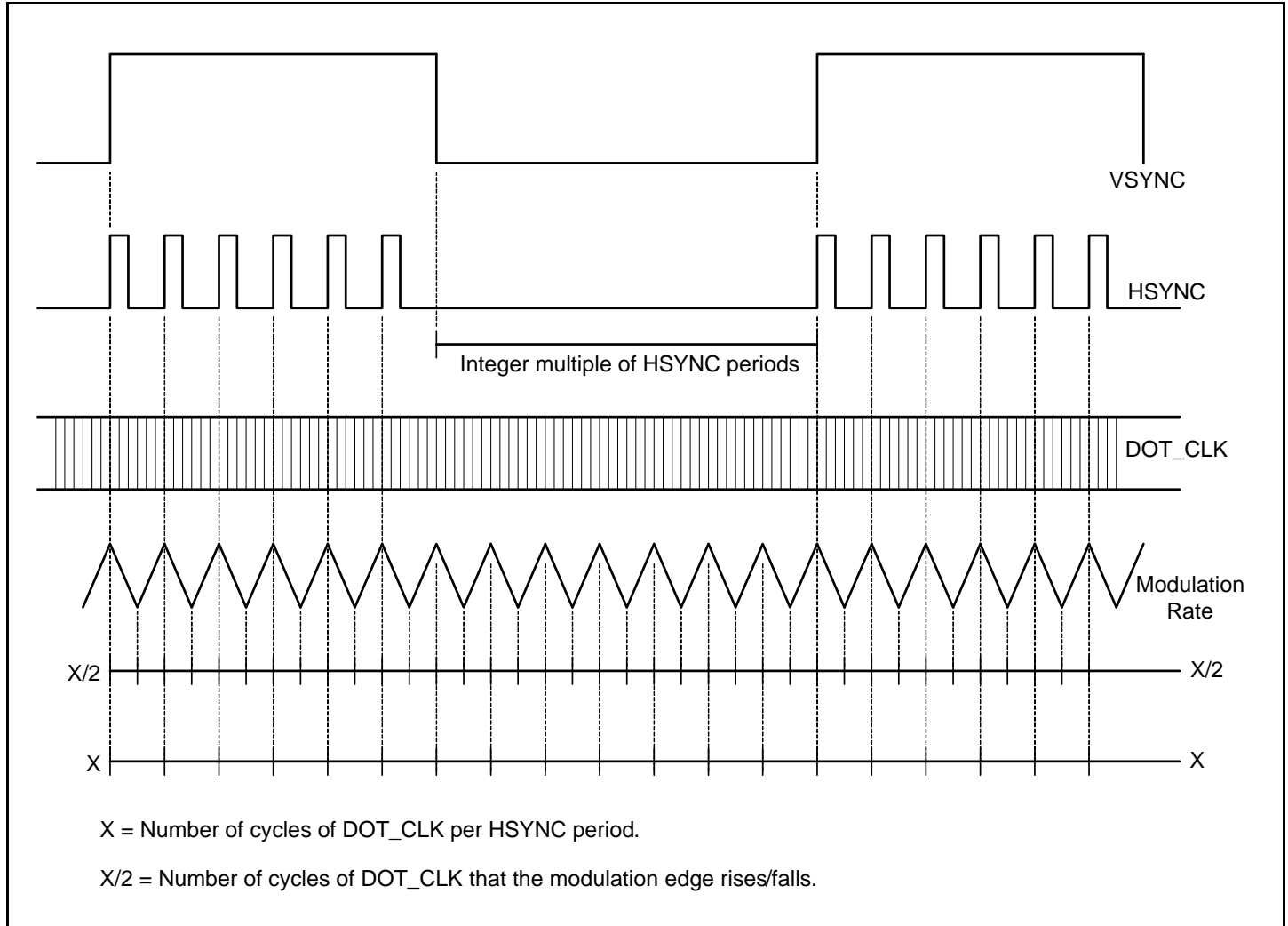
$$F_{MID} = F_{VCO}/ 8$$

$$F_{MOD} = 31.25kHz = 50.625MHz/NC.$$

NC = 1620

MOD = (25MHz * 1620)/(2 * 50.625MHz) = 400

NSS = (1620/2)+(1620/8)*(27.27MHz-26.73MHz)/27MHz = 814.

VSYNC, HSYNC, DOT_CLK – Modulation Rate Relationship

VSYNC

HSYNC

Integer multiple of HSYNC periods

DOT_CLK

Modulation
Rate

X/2   X/2

X   X

X = Number of cycles of DOT_CLK per HSYNC period.

X/2 = Number of cycles of DOT_CLK that the modulation edge rises/falls.

Zero capacitor (Cz) = 280pF

LOOP FILTER

The loop filter for each PLL can be programmed to optimize
the jitter performance. The low-pass frequency response of
the PLL is the mechanism that dictates the jitter transfer
characteristics. The loop bandwidth can be extracted from
the jitter transfer. A narrow loop bandwidth is good for jitter
attenuation while a wide loop bandwidth is best for low jitter
generation. The specific loop filter components that can be
programmed are the resistor via the RZ[4:0] bits, zero
capacitor via the CZ[2:0] bits, pole capacitor via the CP[1:0]
bits, and the charge pump current via the IP#[2:0] bits.

The following equations govern how the loop filter is set:

Pole capacitor (Cp) = 30pF

Charge pump (Ip) = IP#[2:0] uA

VCO gain (Kvco) = 350MHz/V * 2π
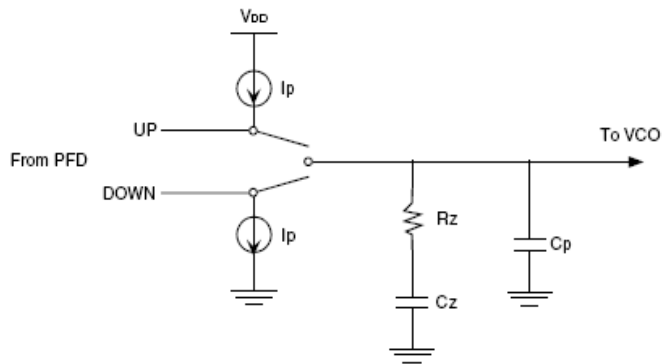
PLL Loop Bandwidth:

Charge pump gain (Kϕ) = Ip / 2π

VCO gain (K$_{VCO}$) = 950MHz/V * 2π

M = Total multiplier value (See the PRE-SCALERS, FEEDBACK-DIVIDERS, POST-DIVIDERS section for more detail)

ωc = (Rz * Kϕ * K$_{VCO}$ * Cz)/(M * (Cz + Cp))

Fc = ωc / 2π

Note, the phase/frequency detector frequency (F$_{PFD}$) is typically seven times the PLL closed-loop bandwidth (Fc) but too high of a ratio will reduce your phase margin thus compromising loop stability.

To determine if the loop is stable, the phase margin (ϕm) would need to be calculated as follows.

Phase Margin:

ωz = 1 / (Rz * Cz)

ωp = (Cz + Cp)/(Rz * Cz * Cp)

ϕm = (360 / 2π) * [tan^{-1}(ωc/ ωz) - tan^{-1}(ωc/ ωp)]

To ensure stability in the loop, the phase margin is recommended to be > 60° but too high will result in the lock time being excessively long. Certain loop filter parameters would need to be compromised to not only meet a required loop bandwidth but to also maintain loop stability.

Example

Fc = 150KHz is the desired loop bandwidth. The total A*M value is 160. The ζ(damping factor) target should be 0.7, meaning the loop is critically damped. Given Fc and A*M, an optimal loop filter setting needs to be solved for that will meet both the PLL loop bandwidth and maintain loop stability.

Choose a mid-range charge pump from register table

Icp= 11.9uA.

Kϕ * K$_{VCO}$ = 300MHz/V * 40uA = 12000A/Vs

ωc = 2π * Fc = 9.42x10^5 s^{-1}

ωp = (Cz + Cp)/(Rz * Cz * Cp) = ωz (1 + Cz / Cp)

Solving for Rz, the best possible value Rz=30kOhms (RZ[1:0]=10) gives

ζ= 1.2

Solving back for the PLL loop bandwidth, Fc=149kHz.

The phase margin must be checked for loop stability.

ϕm = (360 / 2π) * [tan$_{-1}$ (9.42x10^5 s^{-1} / 1.19x10^{5}s^{-1}) - tan^{-1}(9.42x10^5 s^{-1}/ 1.23x10^6 s^{-1})] = 45°

The phase margin would be acceptable with a fairly stable loop.

SEL[1:0] Function

The IDT5P49EE502 can support up to three unique configurations. Users may pre-programmed all these configurations, and select the configurations using SEL[1:0] pins. Alternatively, users may use I2C interface to configure these registers on- the-fly.

Always power with SEL1=1 and/or SEL0=1.

Power Down/Sleep Mode is selected by the No_PD bit. No_PD=0 enables Power Down mode with no outputs. No_PD=1 enables sleep mode with 32kHz output on OUT4.

SEL1	SEL0	Configuration Selections
0	0	Power Down/Sleep Mode
0	1	Select CONFIG0
1	0	Select CONFIG1
1	1	Select CONFIG2

Configuration OUTx IO Standard

Users can configure the individual output IO standard from a single 1.8V power supply. Each output can support 1.8V/

2.5V or 3.3V LVCMOS. VDDO1 must have the highest voltage of any pin on the device. VDDO2 may have any value between 1.8V and VDDO1.

Programming the Device

I^{2}C may be used to program the IDT5P49EE502.

– Device (slave) address = 7'b1101010

I^{2}C Programming

The IDT5P49EE502 is programmed through an I2C-Bus serial interface, and is an I2C slave device. The read and write transfer formats are supported. The first byte of data after a write frame to the correct slave address is interpreted as the register address; this address auto-increments after each byte written or read.

The frame formats are shown in the following illustration.

Framing

R/W
0 - Slave will be written by master
1 - Slave will be read by master
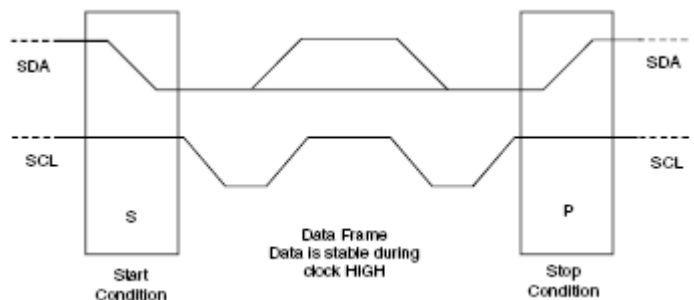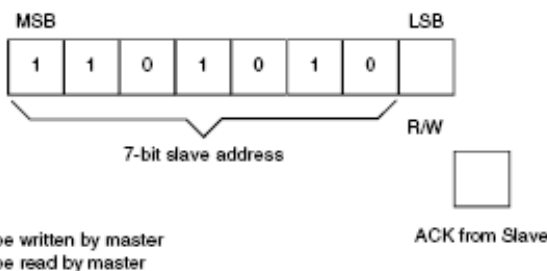
The first byte transmitted by the Master is the Slave Address followed by the R/W bit.
The Slave acknowledges by sending a "1" bit.

First Byte Transmitted on I^{2}C Bus

External I^{2}C Interface Condition

KEY:

▢ From Master to Slave

▨ From Master to Slave, but can be omitted if followed by the correct sequence
Normally data transfer is terminated by a STOP condition generated by the Master. However, if the Master still wishes to communicate on the bus, it can generate a repeated START condition, and address another Slave address without first generating a STOP condition.

▤ From Slave to Master

SYMBOLS:
ACK - Acknowledge (SDA LOW)
NACK - Not Acknowledge (SDA HIGH)
Sr - Repeated Start Condition
S - START Condition
P - STOP Condition

EEPROM Interface

The IDT5P49EE502 can store its configuration in an internal EEPROM. The contents of the device's internal programming registers can be saved to the EEPROM by issuing a save instruction (ProgSave) and can be loaded back to the internal programming registers by issuing a restore instruction (ProgRestore).

To initiate a save or restore using I^{2}C, only two bytes are transferred. The Device Address is issued with the read/write bit set to "0", followed by the appropriate command code. The save or restore instruction executes after the STOP condition is issued by the Master, during which time the IDT5P49EE502 will not generate Acknowledge bits. The IDT5P49EE502 will acknowledge the instructions after it has completed execution of them. During that time, the I^{2}C bus should be interpreted as busy by all other users of the bus.

On power-up of the IDT5P49EE502, an automatic restore is performed to load the EEPROM contents into the internal programming registers. The IDT5P49EE502 will be ready to accept a programming instruction once it acknowledges its 7-bit I^{2}C address.

Progwrite

S	Address	R/W	ACK	Command Code	ACK	Register	ACK	Data	ACK	P
	7-bits	0	1-bit	8-bits: xxxxxx00	1-bit	8-bits	1-bit	8-bits	1-bit	

Progwrite Command Frame

Writes can continue as long as a Stop condition is not sent and each byte will increment the register address.

Progread

Note: If the expected read command is not from the next higher register to the previous read or write command, then set a known "read" register address prior to a read operation by issuing the following command:

S	Address	R/W	ACK	Command Code	ACK	Register	ACK	P
	7-bits	0	1-bit	8-bits: xxxxxx00	1-bit	8-bits	1-bit	

Prior to Progread Command Set Register Address

The user can ignore the STOP condition above and use a repeated START condition instead, straight after the slave acknowledgement bit (i.e., followed by the Progread command):

Sr	Address	R/W	ACK	ID Byte	ACK	Data_1	ACK	Data_2	ACK	Data_last	NACK	P
	7-bits	1	1-bit	8 bits	1-bit	8-bits	1-bit	8-bits	1-bit	8-bits	1-bit	

Progread Command Frame

Progsave

S	Address	R/W	ACK	Command Code	ACK	P
	7-bits	0	1-bit	8-bits:xxxxxx01	1-bit	

Note:
PROGWRITE is for writing to the IDT5P49EE502 registers.
PROGREAD is for reading the IDT5P49EE502 registers.
PROGSAVE is for saving all the contents of the
IDT5P49EE502 registers to the EEPROM.
PROGRESTORE is for loading the entire EEPROM
contents to the IDT5P49EE502 registers.

Progrestore

S	Address	R/W	ACK	Command Code	ACK	P
	7-bits	0	1-bit	8-bits:xxxxxx10	1-bit	

During PROGRESTORE, outputs will be turned off to
ensure that no improper voltage levels are experienced
before initialization.

I^{2}C Bus DC Characteristics

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
V_{IH}	Input HIGH Level		0.7xVDD		5.5	V
V_{IL}	Input LOW Level				0.3xVDD	V
V_{HYS}	Hysteresis of Inputs		0.05xVDD			V
I_{IN}	Input Leakage Current	V_{DD} = 0V			±1.0	µA
V_{OL}	Output LOW Voltage	I_{OL} = 3 mA			0.4	V

I^{2}C Bus AC Characteristics for Standard Mode[1]

Symbol	Parameter	Min	Typ	Max	Unit
F_{SCLK}	Serial Clock Frequency (SCL)	0		100	kHz
t_{BUF}	Bus free time between STOP and START	4.7			µs
$t_{SU:START}$	Setup Time, START	4.7			µs
$t_{HD:START}$	Hold Time, START	4			µs
$t_{SU:DATA}$	Setup Time, data input (SDA)	250			ns
$t_{HD:DATA}$	Hold Time, data input (SDA)[2]	0			µs
t_{OVD}	Output data valid from clock			3.45	µs
C_B	Capacitive Load for Each Bus Line			400	pF
t_R	Rise Time, data and clock (SDA, SCLK)			1000	ns
t_F	Fall Time, data and clock (SDA, SCLK)			300	ns
t_{HIGH}	HIGH Time, clock (SCLK)	4			µs
t_{LOW}	LOW Time, clock (SCLK)	4.7			µs
$t_{SU:STOP}$	Setup Time, STOP	4			µs

1) No activity is allowed on I^{2}C lines until VDD>1.62V.
2) A device must internally provide a hold time of at least 300 ns for the SDA signal (referred to the V_{IH}MIN of the SCLK signal) to bridge the undefined region of the falling edge of SCLK.

I²C Bus AC Characteristics for Fast Mode[1]

Symbol	Parameter	Min	Typ	Max	Unit
F_{SCLK}	Serial Clock Frequency (SCL)	0		400	kHz
t_{BUF}	Bus free time between STOP and START	1.3			µs
$t_{SU:START}$	Setup Time, START	0.6			µs
$t_{HD:START}$	Hold Time, START	0.6			µs
$t_{SU:DATA}$	Setup Time, data input (SDA)	100			ns
$t_{HD:DATA}$	Hold Time, data input (SDA) [1]	0			µs
t_{OVD}	Output data valid from clock			0.9	µs
C_B	Capacitive Load for Each Bus Line			400	pF
t_R	Rise Time, data and clock (SDA, SCL)	$20 + 0.1 \times C_B$		300	ns
t_F	Fall Time, data and clock (SDA, SCL)	$20 + 0.1 \times C_B$		300	ns
t_{HIGH}	HIGH Time, clock (SCL)	0.6			µs
t_{LOW}	LOW Time, clock (SCL)	1.3			µs
$t_{SU:STOP}$	Setup Time, STOP	0.6			µs

1) No activity is allowed on I²C lines until VDD>1.62V.
2) A device must internally provide a hold time of at least 300 ns for the SDA signal (referred to the V_{IH}MIN of the SCL signal) to bridge the undefined region of the falling edge of SCL.

Absolute Maximum Ratings

Stresses above the ratings listed below can cause permanent damage to the IDT5P49EE502. These ratings, which are standard values for IDT commercially rated parts, are stress ratings only. Functional operation of the device at these or any other conditions above those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods can affect product reliability. Electrical parameters are guaranteed only over the recommended operating temperature range.

Symbol	Description	Max	Unit
V_{DD}	Internal Power Supply Voltage	-0.5 to +4.6	V
V_I	Input Voltage	-0.5 to +4.6	V
V_O	Output Voltage (not to exceed 4.6 V)	-0.5 to V_{DD}+0.5	V
T_J	Junction Temperature	150	°C
T_{STG}	Storage Temperature	-65 to +150	°C

Recommended Operation Conditions

Symbol	Parameter	Min	Typ	Max	Unit
V_{DD}	Power supply voltage for VDD	1.62	1.8	1.98	V
T_A	Operating temperature, ambient	-40		+85	°C
C_{LOAD_OUT}	Maximum load capacitance (3.3V LVTTL only)			15	pF
C_{LOAD_OUT}	Maximum load capacitance (1.8V or 2.5V LVTTL only)			8	pF
F_{IN}	External reference crystal	8		30	MHz
	External reference clock CLKIN	1		40	
t_{PU}	Power up time for all V_{DD}s to reach minimum specified voltage (power ramps must be monotonic)	0.05		5	ms

Capacitance (T_A = +25 °C, f = 1 MHz, V_{IN} = 0V)

Symbol	Parameter	Min	Typ	Max	Unit
C_{IN}	Input Capacitance		3		pF
Crystal Specifications					
XTAL_FREQ	Crystal frequency	8		30	MHz
XTAL_MIN	Minimum crystal load capacitance		TBD		pF
XTAL_MAX	Maximum crystal load capacitance		35.4		pF
XTAL_V_{PP}	Voltage swing (peak-to-peak, nominal)	1.5	2.3	3.2	V

DC Electrical Characteristics for 3.3 Volt LVTTL [1]

Symbol	Parameter	Test Conditions	Min	Typ	Max	Unit
V_{OH}	Output HIGH Voltage	I_{OH} = 33mA	2.4		VDDO	V
V_{OL}	Output LOW Voltage	I_{OH} = 33mA			0.4	V
V_{IH}	Input HIGH Voltage		2			V
V_{IL}	Input LOW Voltage				0.8	V
I_{OZDD}	Output Leakage Current	3-state outputs			5	µA

DC Electrical Characteristics for 2.5Volt LVTTL [1]

Symbol	Parameter	Test Conditions	Min	Typ	Max	Unit
V_{OH}	Output HIGH Voltage	I_{OH} = 25mA	2.1		VDDO	V
V_{OL}	Output LOW Voltage	I_{OH} = 25mA			0.4	V
I_{OZDD}	Output Leakage Current	3-state outputs			5	µA

DC Electrical Characteristics for 1.8Volt LVTTL [1]

Symbol	Parameter	Test Conditions	Min	Typ	Max	Unit
V_{OH}	Output HIGH Voltage	VDD = 1.71V to 1.89V	0.65*VDDO		VDDO	V
V_{OL}	Output LOW Voltage				0.35*VDDO	V
I_{OZDD}	Output Leakage Current	3-state outputs			5	µA

Power Supply Characteristics for LVTTL Outputs

Symbol	Parameter	Test Conditions	Typ	Max	Unit
I_{TOT}	Total Power VDD Supply Current	$F_{REFERENCE\ CLOCK}$ = 25 MHz, C_L = 7 pF	TBD		mA

1: See "Recommended Operating Conditions" table. Alway completely power up VDD and VDDx prior to applying VDDO power.

AC Timing Electrical Characteristics

(Spread Spectrum Generation = OFF)

Symbol	Parameter	Test Conditions	Min.	Typ.	Max.	Units
f_{IN}	Input Frequency	Input Frequency Limit (CLKIN)	1[1]		40	MHz
1 / t1	Output Frequency	Single Ended Clock output limit (LVTTL) 3.3V	0.001		120	MHz
		Single Ended Clock output limit (LVTTL) 2.5V			110	MHz
		Single Ended Clock output limit (LVTTL) 1.8V			100	MHz
f_{VCO}	VCO Frequency	VCO operating Frequency Range	100		475	MHz
f_{PFD}	PFD Frequency	PFD operating Frequency Range	0.5 [1]		20	MHz
t2	Input Duty Cycle	Duty Cycle for Input	40		60	%
t3	Output Duty Cycle	Measured at VDD/2	45		55	%
t4	Slew Rate, SLEWx(bits) = 00	Single-Ended 3.3V LVCMOS Output clock rise and fall time, 20% to 80% of VDD (Output Load = 7 pF)		3.5		V/ns
	Slew Rate, SLEWx(bits) = 01	Single-Ended 3.3V LVCMOS Output clock rise and fall time, 20% to 80% of VDD (Output Load = 7 pF)		2.75		
	Slew Rate, SLEWx(bits) = 10	Single-Ended 3.3V LVCMOS Output clock rise and fall time, 20% to 80% of VDD (Output Load = 7 pF)		2		
	Slew Rate, SLEWx(bits) = 11	Single-Ended 3.3V LVCMOS Output clock rise and fall time, 20% to 80% of VDD (Output Load = 7 pF)		1.25		
t5	Clock Jitter	Peak-to-peak period jitter, CLK outputs measured at VDD/2; f_{PFD} >= 10 MHz Single output frequency only.			100	ps
		Peak-to-peak period jitter, CLK outputs measured at VDD/2; f_{PFD} >= 10 MHz Multiple output frequencies switching.			200	ps
t6	Output Skew	Skew between any output (Same freq and IO type, FOUT >10MHz)			200	ps
t7	Lock Time	PLL Lock Time from Power-up (using MHz reference clock)[2]		5	20	ms
		PLL Lock time from shutdown mode			5	ms

1. Input clock (square wave) may be used at 1 MHz.
2. Time from supply voltage crosses VDD=1.62V to PLLs are locked.

Spread Spectrum Generation Specifications

Symbol	Parameter	Description	Min	Typ	Max	Unit
f_{IN}	Input Frequency	Input Frequency Limit	1 [1]		40	MHz
f_{MOD}	Mod Frequency	Modulation Frequency	32		120	kHz
f_{SPREAD}	Spread Value	Amount of Spread Value (programmable) - Down Spread	Programmable			$\%f_{OUT}$
		Amount of Spread Value (programmable) - Center Spread	Programmable			

1) Practical lower frequency is determined by loop filter settings.

Test Circuits and Conditions [1]
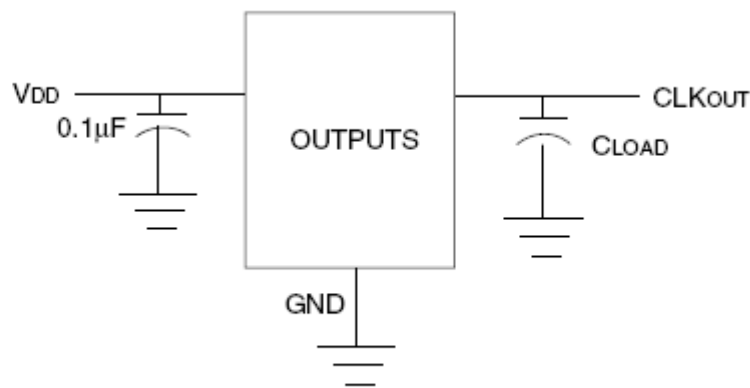
NOTE:
1. All V_DD pins must be tied together.

Test Circuits for DC Outputs

Other Termination Scheme (Block Diagram)

LVTTL Output Load: ~7pF for each output

Programming Registers Table

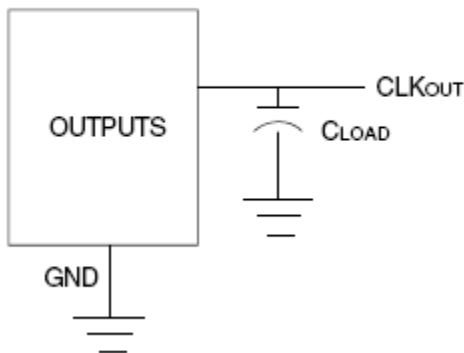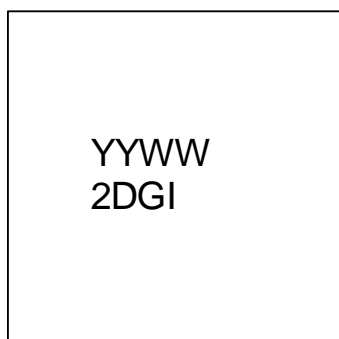| Addr	Default Register Hex Value	Bit # 7	6	5	4	3	2	1	0	Description
0x00	04	ONXTALB	Reserved							ONXTALB - MHz Crystal active low
0x01	00	INV[0]	SLEW1[0:1]		Reserved	PS0[2:1]		Reserved		INV[#] - Invert output#
0x02	00	Reserved								SLEW#[0:1] - output# slew setting
0x03	00	Reserved								0 0 - 5.1V/ns
0x04	00	INV[1]	SLEW1[0:1]		Reserved	PS1[2:1]		Reserved		0 1 - 4.4V/ns
0x05	00	INV[2]	SLEW2[0:1]		Reserved	PS2[2:1]		Reserved		1 0 - 2.8V/ns
0x06	00	INV[3]	SLEW3[0:1]		Reserved	PS3[2:1]		Reserved		1 1 - 1.8V/ns
0x07	00	Reserved								PS#[2:1] -Power Select
0x08	00	Reserved								00 - Reserved
0x09	00	Reserved	INV[4]	SLEW4[0:1]		Reserved				01 - OUT# connects to VDDO1
0x0A	00	Reserved								10 - OUT# connects to VDDO2
0x0B	00	Reserved								11 - Reserved
0x0C	00	Reserved								CLK4 is tied to VDD02
0x0D	00	Reserved								
0x0E	00	REFA[7:0]								**Configuration0** REFA[7:0] - Reference Divide PLLA
0x0F	00	FBA[10:3]								FBA[10:0] - Feedback Divide PLLA
0x10	00	Reserved					FBA[2:0]			
0x11	00	Reserved	XDIVA	RZA[1:0]		IPA[2:0]			Reserved	XDIVA - FB predivide PLLA; 0 - /1; 1 - /4 RZA[1:0] - Zero Resistor PLLA 00 - 5kOhm 01 - 10kOhm 10 - 30kOhm 11 - 80kOhm IPA[2:0] - charge Pump Current PLLA 100 - 6.3uA 101 -11.9 uA 110 - 17.7 uA 111 - 22.7uA
0x12	00	REFB[7:0]								REFB[7:0] - Reference Divide PLLB
0x13	00	FBB[10:3]								FBB[10:0] - Feedback Divide PLLB
0x14	00	MOD[4:0]					FBB[2:0]			PLLB Spread Parameters MOD[12:0] NC[10:0] NSS[12:0]
0x15	00	MOD[12:5]								
0x16	00	NC[10:3]								
0x17	00	NSS[4:0]					NC[2:0]			
0x18	00	NSS[12:5]								
0x19	20	Reserved			IPB[2:0]			RZB[1:0]		RZB[1:0] - Zero Resistor PLLB
0x1A	00	Reserved							SSENB_B	00 - 5kOhm 01 - 10kOhm 10 - 30kOhm 11 - 80kOhm IPB[2:0] - charge Pump Current PLLB 000 - 0.37uA, 100 - 6.3uA 001 - 1.1uA, 101 - 11.9uA 010 - 1.8 uA, 110 - 17.7uA 011 - 3.4uA, 111 - 22.7uA
0x1B	00	REFC[7:0]								REFC[7:0] - Reference Divide PLLC
0x1C	00	FBC[10:3]								FBC[10:0] - Feedback Divide PLLC
0x1D	00	Reserved					FBC[2:0]			
0x1E	00	FBC2[7:0]								FBC2 - Feedback Predivide PLLC Turn on using XDIVC=1

Addr	Default Register Hex Value	Bit #								Description
		7	6	5	4	3	2	1	0	
0x1F	00	IPC[2:0]			RZC[1:0]		Reserved	XDIVC	Reserved	RZC[1:0] - Zero Resistor PLLC 00 - 5kOhm 01 - 10kOhm 10 - 30kOhm 11 - 80kOhm IPC[2:0] - charge Pump Current PLLC 100 - 6.3uA 101 -11.9 uA 110 - 17.7 uA 111 - 22.7uA
0x20	00	REFD[7:0]								REFD[7:0] - Reference Divide PLLD
0x21	00	FBD[10:3]								FBD[10:0] - Feedback Divide PLLD
0x22	00	Reserved					FBD[2:0]			
0x23	00	XDIVD	RZD[1:0]		IPD[2:0]			Reserved		XDIVD - FB predivide PLLD; 0 - /1; 1 - /4 RZD[1:0] - Zero Resistor PLLD 00 - 5kOhm 01 - 10kOhm 10 - 30kOhm 11 - 80kOhm IPD[2:0] - charge Pump Current PLLD 100 - 6.3uA 101 -11.9 uA 110 - 17.7 uA 111 - 22.7uA
0x24	00	OD0[7:0]								
0x25	00	Reserved								
0x26	00	Reserved								
0x27	00	OD1[7:0]								
0x28	00	OD2[7:0]								
0x29	00	OD3[7:0]								
0x2A	00	Reserved								
0x2B	00	Reserved								
0x2C	00	OD4[7:0]								
0x2D	00	SCR4[1:0]		Reserved				SCR3[1:0]		SRC3[1:0] - OD4 source 00 - off; 10 - PLLA 01 - Reference; 11 - PLLD SRC4[1:0] - OD4 source 00 - off; 10 - PLLC 01 - PLLA; 11 - Reference
0x2E	00	SCR2[1:0]		SCR1[1:0]		Reserved				SRC1[1:0] - OD1 source 00 - off; 10 - PLLB 01 - PLLA; 11 - PLLD SRC2[1:0] - OD2 source 00 - off; 10 - PLLC 01 - PLLA; 11 - PLLD
0x2F	00	SCR0[1:0]		Reserved						SRC0[1:0] - OD0 source 00 - off; 10 - PLLC 01 - PLLB; 11 - PLLD
0x30	00	Reserved								
0x31	00	PDB[4]	Reserved		OE[4]	Reserved				PDB[4:0] - Powerdown OUT#. PDB#=0, OUT# driven low OE[4:0] - Output enable OUT#. OE#=0, OUT# tri-stated. If PDB#=OE#=0, OUT# driven low
0x32	00	Reserved		OE[3]	OE[2]	OE[1]	Reserved		OE[0]	
0x33	00	Reserved		PDB[3]	PDB[2]	PDB[1]	Reserved		PDB[0]	

Addr	Default Register Hex Value	7	6	5	4	3	2	1	0	Description
0x34	00	REFA[7:0]								**Configuration1** (See definitions from Configuration0 above)
0x35	00	FBA[10:3]								
0x36	00	Reserved					FBA[2:0]			
0x37	00	Reserved	XDIVA	RZA[1:0]		IPA[2:0]			Reserved	
0x38	00	REFB[7:0]								
0x39	00	FBB[10:3]								
0x3A	00	MOD[4:0]					FBB[2:0]			
0x3B	00	MOD[12:5]								
0x3C	00	NC[10:3]								
0x3D	00	NSS[4:0]					NC[2:0]			
0x3E	00	NSS[12:5]								
0x3F	40	Reserved				IPB[2:0]			RZB[1:0]	
0x40	00	Reserved						Reserved	SSENB_B	
0x41	00	REFC[7:0]								
0x42	00	FBC[10:3]								
0x43	00	Reserved					FBC[2:0]			
0x44	00	FBC2[7:0]								
0x45	00	IPC[2:0]			RZC[1:0]		Reserved	XDIV	Reserved	
0x46	00	REFD[7:0]								
0x47	00	FBD[10:3]								
0x48	00	Reserved					FBD[2:0]			
0x49	00	XDIVD	RZD[1:0]		IPD[2:0]			Reserved		
0x4A	00	OD0[7:0]								
0x4B	00	Reserved								
0x4C	00	Reserved								
0x4D	00	OD1[7:0]								
0x4E	00	OD2[7:0]								
0x4F	00	OD3[7:0]								
0x50	00	Reserved								
0x51	00	Reserved								
0x52	00	OD4[7:0]								
0x53	00	SCR4[1:0]		Reserved				SCR3[1:0]		
0x54	00	SCR2[1:0]		SCR1[1:0]		Reserved				
0x55	00	SCR0[1:0]		Reserved						
0x56	00	Reserved								
0x57	00	PDB[4]	Reserved		OE[4]	Reserved				
0x58	00	Reserved		OE[3]	OE[2]	OE[1]	Reserved		OE[0]	
0x59	00	Reserved		PDB[3]	PDB[2]	PDB[1]	Reserved		PDB[0]	

Addr	Default Register Hex Value	Bit #								Description
		7	**6**	**5**	**4**	**3**	**2**	**1**	**0**	
0x5A	00	REFA[7:0]								**Configuration2** (See definitions from Configuration0 above)
0x5B	00	FBA[10:3]								
0x5C	00	Reserved					FBA[2:0]			
0x5D	00	Reserved	XDIVA	RZA[1:0]		IPA[2:0]			Reserved	
0x5E	00	REFB[7:0]								
0x5F	00	FBB[10:3]								
0x60	00	MOD[4:0]					FBB[2:0]			
0x61	00	MOD[12:5]								
0x62	00	NC[10:3]								
0x63	00	NSS[4:0]					NC[2:0]			
0x64	00	NSS[12:5]								
0x65	40	Reserved			IPB[2:0]			RZB[1:0]		
0x66	00	Reserved					Reserved		SSENB_B	
0x67	00	REFC[7:0]								
0x68	00	FBC[10:3]								
0x69	00	Reserved					FBC[2:0]			
0x6A	00	FBC2[7:0]								
0x6B	00	IPC[2:0]			RZC[1:0]		Reserved	XDIV	Reserved	
0x6C	00	REFD[7:0]								
0x6D	00	FBD[10:3]								
0x6E	00	Reserved					FBD[2:0]			
0x6F	00	XDIVD	RZD[1:0]		IPD[2:0]			Reserved		
0x70	00	OD0[7:0]								
0x71	00	Reserved								
0x72	00	Reserved								
0x73	00	OD17:0]								
0x74	00	OD2[7:0]								
0x75	00	OD3[7:0]								
0x76	00	Reserved								
0x77	00	Reserved								
0x78	00	OD4[7:0]								
0x79	00	SCR4[1:0]		Reserved				SCR3[1:0]		
0x7A	00	SCR2[1:0]		SCR1[1:0]		Reserved				
0x7B	00	SCR0[1:0]		Reserved						
0x7C	00	Reserved								
0x7D	00	PDB[4]	Reserved		OE[4]	Reserved				
0x7E	00	Reserved		OE[3]	OE[2]	OE[1]	Reserved		OE[0]	
0x7F	00	Reserved		PDB[3]	PDB[2]	PDB[1]	Reserved		PDB[0]	

Marking Diagram (ND20)

```
┌─────────────────────┐
│                     │
│                     │
│      YYWW           │
│      2DGI           │
│                     │
│                     │
│                     │
└─────────────────────┘
```
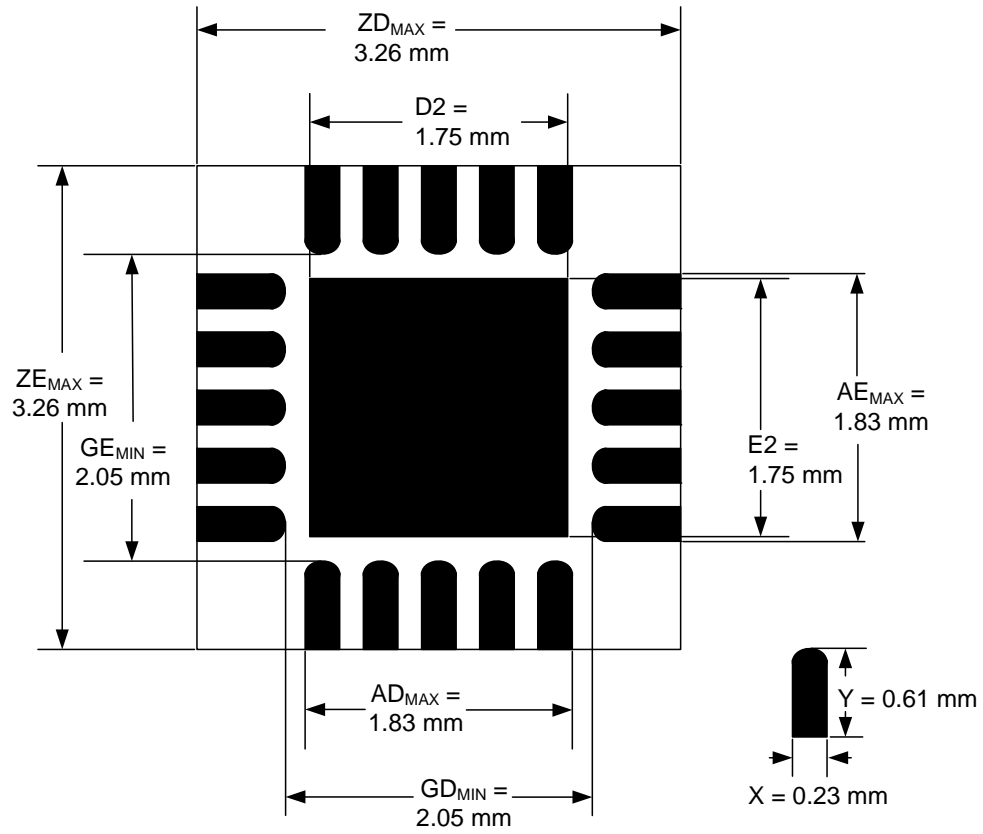
Notes:

1. "Z" is the device step (1 to 2 characters).

2. YYWW is the last two digits of the year and week that the part was assembled.

3. "$" is the assembly mark code.

4. "G" after the two-letter package code designates RoHS compliant package.

5. "I" at the end of part number indicates industrial temperature range.

6. Bottom marking: country of origin if not USA.

Thermal Characteristics 20-pin VFQFPN

Parameter	Symbol	Conditions	Min.	Typ.	Max.	Units
Thermal Resistance Junction to Ambient	θ_{JA}	Still air				°C/W
	θ_{JA}	1 m/s air flow				°C/W
	θ_{JA}	3 m/s air flow				°C/W
Thermal Resistance Junction to Case	θ_{JC}					°C/W

20-pin QFN Solder Mask

$ZD_{MAX} =$
3.26 mm

$D2 =$
1.75 mm

$ZE_{MAX} =$
3.26 mm

$GE_{MIN} =$
2.05 mm

$AE_{MAX} =$
1.83 mm

$E2 =$
1.75 mm

$AD_{MAX} =$
1.83 mm

$GD_{MIN} =$
2.05 mm

$Y = 0.61$ mm

$X = 0.23$ mm

Package Outline and Package Dimensions (20-pin QFN)

Package dimensions are kept current with JEDEC Publication No. 95
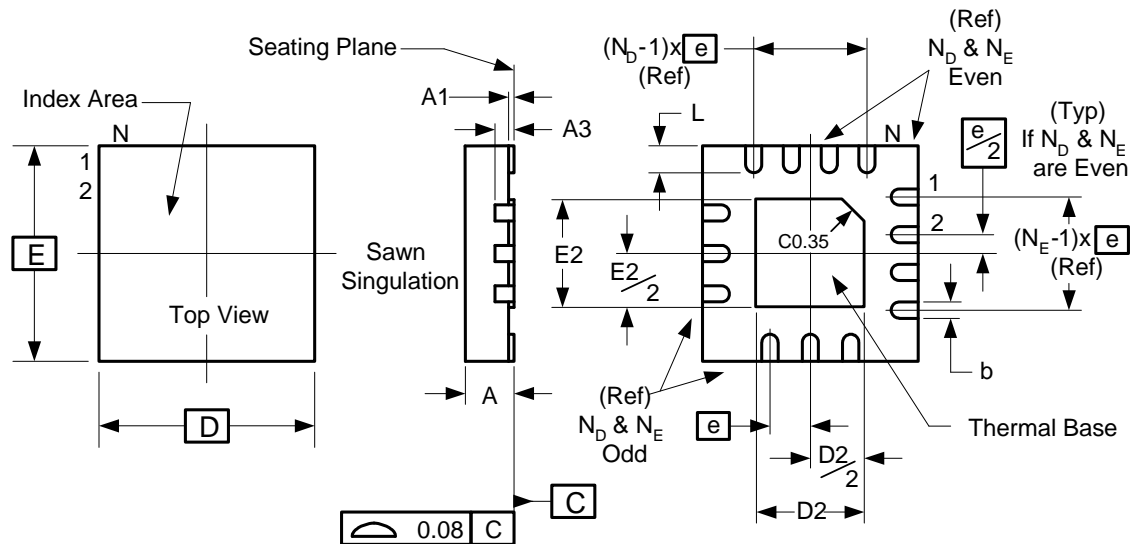
Symbol	Millimeters	
	Min	**Max**
A	0.80	1.00
A1	0	0.05
A3	0.25 Reference	
b	0.15	0.23
e	0.40 BASIC	
N	20	
N_D	5	
N_E	5	
D x E BASIC	3.00 x 3.00	
D2	1.55	1.75
E2	1.55	1.75
L	0.30	0.50

Ordering Information

Part / Order Number	Marking	Shipping Packaging	Package	Temperature
5P49EE502NDGI	See page 22	Tubes	20pin VFQFPN	-40 to +85° C
5P49EE502NDGI8	See page 22	Tape and Reel	20pin VFQFPN	-40 to +85° C

Parts that are ordered with a "G" after the two-letter pacakage code are the Pb-Free configuration and are RoHS compliant.

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Revision History

Rev.	Originator	Date	Description of Change
--	R.Willner	10/27/09	Initial Preliminary Datasheet
A	R.Willner	11/10/09	Revised pinout.
B	R.Willner	3/25/10	Typographical changes. Register corrections. Correct spread spectrum calculations.
C	R. Willner	6/11/10	Default configuration. Clarification of OUT4 is tied to VDDO2.
D		07/26/10	Updated thermal pad and dimensions on package drawing.

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