

μ A102 • μ A302 • μ A110 • μ A310

VOLTAGE FOLLOWER OPERATIONAL AMPLIFIER

FAIRCHILD LINEAR INTEGRATED CIRCUITS

GENERAL DESCRIPTION – The μ A102/302 and μ A110/310 are monolithic Operational Amplifiers internally connected as unity gain non-inverting amplifiers. They are constructed using the Fairchild Planar* epitaxial process. These circuits are ideal for such applications as fast sample and hold circuits, active filters, or as general purpose buffers. Super-beta transistors are used allowing the devices to operate at very low input currents without sacrificing speed. They may be used interchangeably with the μ A101 and the μ A741 in voltage follower applications. The μ A110/310 are suggested for new designs and are direct replacements for the μ A102/302. They feature lower offset voltage, drift, bias current, noise, plus higher speed and a wider operating voltage range.

- HIGH SLEW RATE – 30 V/ μ s
- LOW INPUT CURRENT
- INTERNALLY COMPENSATED
- PLUG-IN REPLACEMENT FOR BOTH THE μ A101 AND μ A741 VOLTAGE FOLLOWER APPLICATIONS
- WIDE RANGE OF SUPPLY VOLTAGES

ABSOLUTE MAXIMUM RATINGS

Supply Voltage	± 18 V
Internal Power Dissipation (Note 1)	500 mW
Input Voltage (Note 2)	± 15 V
Output Short Circuit Duration (Note 3)	Indefinite
Storage Temperature Range	-65° C to $+150^{\circ}$ C
Operating Temperature Range	
Military (μ A102, μ A110)	-55° C to $+125^{\circ}$ C
Commercial (μ A302, μ A310)	0° C to $+70^{\circ}$ C
Pin Temperature (Soldering, 60 s)	300° C

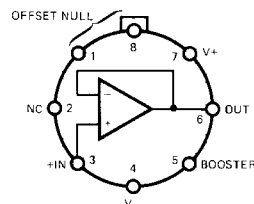
CONNECTION DIAGRAM

8-PIN METAL CAN

(TOP VIEW)

PACKAGE OUTLINE SS

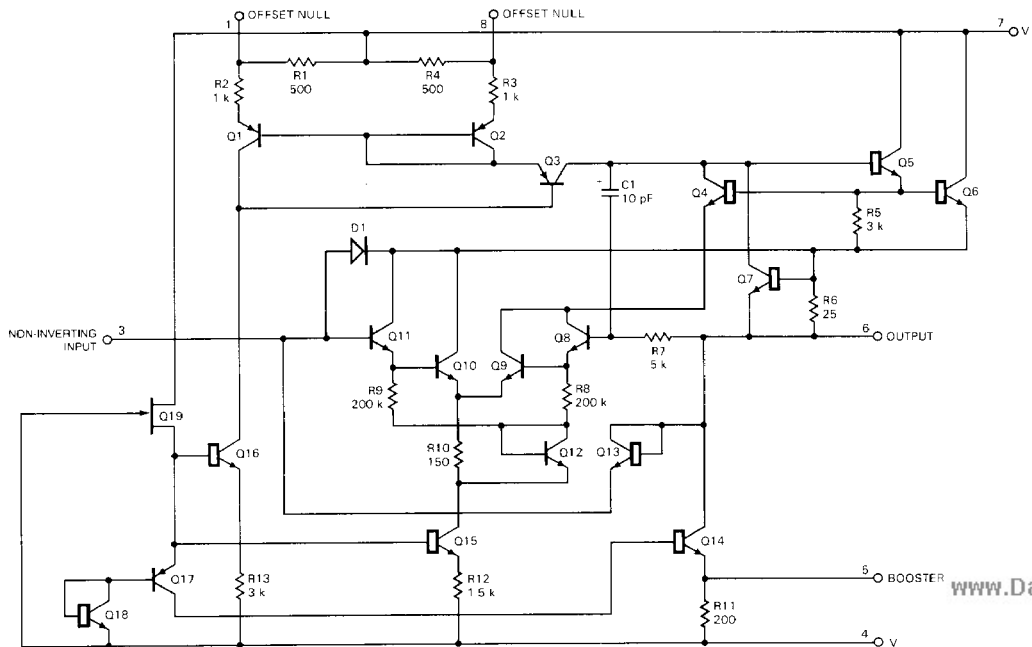
PACKAGE CODE H



ORDER INFORMATION

TYPE	PART NO.
μ A102	μ A102HM
μ A302	μ A302HC
μ A110	μ A110HM ✓
μ A310	μ A310HC ✓

EQUIVALENT CIRCUIT



See notes on following pages

* Planar is a patented Fairchild process.

FAIRCHILD • μ A102 • μ A110 • μ A302 • μ A310 μ A102**ELECTRICAL CHARACTERISTICS:** $V_S = \pm 15$ V, $T_A = 25^\circ\text{C}$, $C_L \leq 100$ pF, unless otherwise specified.

CHARACTERISTICS	CONDITIONS	MIN	TYP	MAX	UNITS
Offset Voltage			2.0	5.0	mV
Average Temperature Coefficient of Offset Voltage			6.0		$\mu\text{V}/^\circ\text{C}$
Input Current			3.0	10	nA
Input Resistance		10^{10}	10^{12}		Ω
Voltage Gain	$R_L \geq 10$ k Ω	0.999	0.9996		
Output Resistance			0.8	2.5	Ω
Output Voltage Swing (Note 4)	$R_L \geq 8$ k Ω	± 10	± 13		V
Supply Current			3.5	5.5	mA
Positive Supply Rejection		60			dB
Negative Supply Rejection		70			dB
Input Capacitance				3.0	pF
Offset Voltage	$-55^\circ\text{C} \leq T_A \leq 125^\circ\text{C}$			7.5	mV
Input Current	$T_A = 125^\circ\text{C}$		3.0	10	nA
	$T_A = -55^\circ\text{C}$		30	100	nA
Voltage Gain	$-55^\circ\text{C} \leq T_A \leq 125^\circ\text{C}$	0.999			
	$R_L \geq 10$ k Ω				
Output Voltage Swing	$R_L \geq 10$ k Ω	± 10			V
Supply Current	$T_A = 125^\circ\text{C}$		2.6	4.0	mA

 μ A302**ELECTRICAL CHARACTERISTICS:** $V_S = \pm 15$ V, $T_A = 25^\circ\text{C}$, $C_L \leq 100$ pF, unless otherwise specified.

CHARACTERISTICS	CONDITIONS	MIN	TYP	MAX	UNITS
Offset Voltage			5.0	15	mV
Average Temperature Coefficient of Offset Voltage			20		$\mu\text{V}/^\circ\text{C}$
Input Current			10	30	nA
Input Resistance		10^9	10^{12}		Ω
Voltage Gain	$R_L \geq 8$ k Ω	0.9985	0.9995	1.000	
Output Resistance			0.8	2.5	Ω
Output Voltage Swing (Note 4)	$R_L \geq 8$ k Ω	± 10			V
Supply Current			3.5	5.5	mA
Positive Supply Rejection		60			dB
Negative Supply Rejection		70			dB
Input Capacitance			3.0		pF
Offset Voltage	$0^\circ\text{C} \leq T_A \leq 70^\circ\text{C}$			20	mV
Input Current	$T_A = 70^\circ\text{C}$		3.0	15	nA
	$T_A = 0^\circ\text{C}$		20	50	nA

FAIRCHILD • μ A102 • μ A110 μ A302 • μ A310 μ A110**ELECTRICAL CHARACTERISTICS:** $\pm 5.0 \text{ V} \leq V_S \leq \pm 18 \text{ V}$, $-55^\circ\text{C} < T_A \leq +125^\circ\text{C}$, unless otherwise specified.

CHARACTERISTICS	CONDITIONS	MIN	TYP	MAX	UNITS
Input Offset Voltage	$T_A = 25^\circ\text{C}$		1.5	4.0	mV
Input Bias Current	$T_A = 25^\circ\text{C}$		1.0	3.0	nA
Input Resistance	$T_A = 25^\circ\text{C}$	10^{10}	10^{12}		Ω
Input Capacitance			1.5		pF
Large Signal Voltage Gain	$T_A = 25^\circ\text{C}$, $V_S = \pm 15 \text{ V}$ $V_{OUT} = \pm 10 \text{ V}$, $R_L = 8 \text{ k}\Omega$	0.999	0.9999		
Output Resistance	$T_A = 25^\circ\text{C}$		0.75	2.5	Ω
Supply Current	$T_A = 25^\circ\text{C}$		3.9	5.5	mA
Input Offset Voltage				6.0	mV
Offset Voltage Temperature Drift	$-55^\circ\text{C} \leq T_A \leq 85^\circ\text{C}$		6.0		$\mu\text{V}/^\circ\text{C}$
	$T_A = 125^\circ\text{C}$		12		$\mu\text{V}/^\circ\text{C}$
Input Bias Current				.10	nA
Large Signal Voltage Gain	$V_S = \pm 15 \text{ V}$, $V_{OUT} = \pm 10 \text{ V}$ $R_L = 10 \text{ k}\Omega$	0.999			
Output Voltage Swing (Note 4)	$V_S = \pm 15 \text{ V}$, $R_L = 10 \text{ k}\Omega$	± 10			V
Supply Current	$T_A = 125^\circ\text{C}$		2.0	4.0	mA
Supply Voltage Rejection Ratio	$\pm 5 \text{ V} \leq V_S \leq \pm 18 \text{ V}$	70	80		dB

 μ A310**ELECTRICAL CHARACTERISTICS:** $\pm 5.0 \text{ V} \leq V_S \leq \pm 18 \text{ V}$, $0^\circ\text{C} \leq T_A \leq +70^\circ\text{C}$, unless otherwise specified.

CHARACTERISTICS	CONDITIONS	MIN	TYP	MAX	UNITS
Input Offset Voltage	$T_A = 25^\circ\text{C}$		2.5	7.5	mV
Input Bias Current	$T_A = 25^\circ\text{C}$		2.0	7.0	nA
Input Resistance	$T_A = 25^\circ\text{C}$	10^{10}	10^{12}		Ω
Input Capacitance			1.5		pF
Large Signal Voltage Gain	$T_A = 25^\circ\text{C}$, $V_S = \pm 15 \text{ V}$ $V_{OUT} = \pm 10 \text{ V}$, $R_L = 8 \text{ k}\Omega$	0.999	0.9999		
Output Resistance	$T_A = 25^\circ\text{C}$		0.75	2.5	Ω
Supply Current	$T_A = 25^\circ\text{C}$		3.9	5.5	mA
Input Offset Voltage				10	mV
Offset Voltage Temperature Drift			10		$\mu\text{V}/^\circ\text{C}$
Input Bias Current				10	nA
Large Signal Voltage Gain	$V_S = \pm 15 \text{ V}$, $V_{OUT} = \pm 10 \text{ V}$ $R_L = 10 \text{ k}\Omega$	0.999			
Output Voltage Swing (Note 4)	$V_S = \pm 15 \text{ V}$, $R_L = 10 \text{ k}\Omega$	± 10			V
Supply Voltage Rejection Ratio	$\pm 5 \text{ V} \leq V_S \leq \pm 18 \text{ V}$	70	80		dB

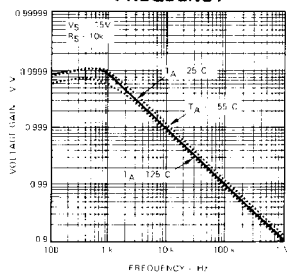
NOTES:

- Rating applies to ambient temperatures up to $+70^\circ\text{C}$. Above $+70^\circ\text{C}$ ambient, derate linearly at $6.3 \text{ mW}/^\circ\text{C}$.
- For supply voltages less than $\pm 15 \text{ V}$, the absolute maximum input voltage is equal to the supply voltage.
- For 102 and 110 continuous short circuit is allowed for case temperature of $+125^\circ\text{C}$ and ambient temperature to $+70^\circ\text{C}$. For 302 and 310 continuous short circuit is allowed for case temperature to $+70^\circ\text{C}$ and ambient temperature to $+55^\circ\text{C}$. It is necessary to insert a resistor greater than $2 \text{ k}\Omega$ in series with the input when the amplifier is driven from low impedance sources to prevent damage when the input is shorted.
- Increased output swing under load can be obtained by connecting an external resistor between the booster and V_- terminals (see curve).

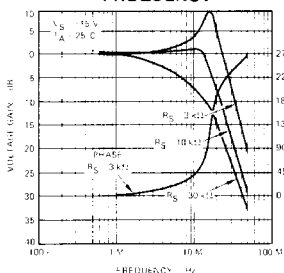
FAIRCHILD • $\mu A102$ • $\mu A110$ $\mu A302$ • $\mu A310$

TYPICAL PERFORMANCE CURVES FOR $\mu A102$ • $\mu A302$ • $\mu A110$ • $\mu A310$

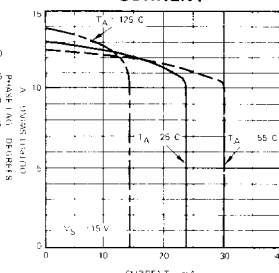
VOLTAGE GAIN AS A FUNCTION OF FREQUENCY



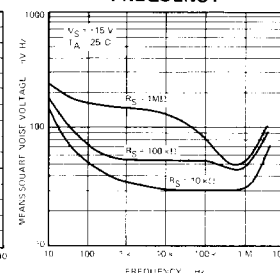
VOLTAGE GAIN AS A FUNCTION OF FREQUENCY



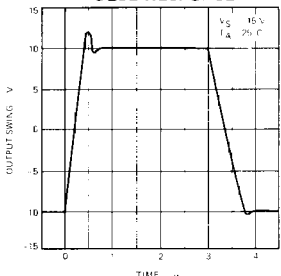
POSITIVE OUTPUT SWING AS A FUNCTION OF CURRENT



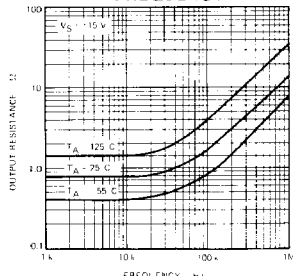
OUTPUT NOISE VOLTAGE AS A FUNCTION OF FREQUENCY



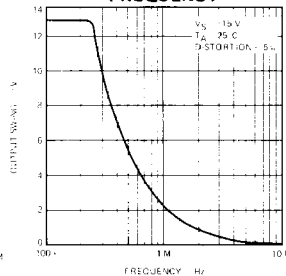
LARGE SIGNAL PULSE RESPONSE



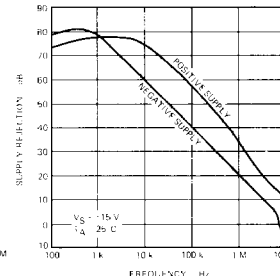
OUTPUT RESISTANCE AS A FUNCTION OF FREQUENCY



LARGE SIGNAL FREQUENCY RESPONSE AS A FUNCTION OF FREQUENCY

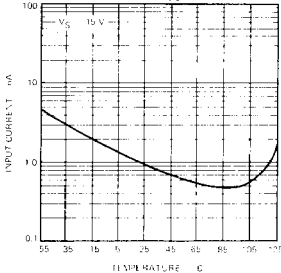


POWER SUPPLY REJECTION AS A FUNCTION OF FREQUENCY

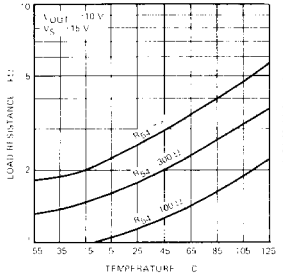


TYPICAL PERFORMANCE CURVES FOR $\mu A102$ • $\mu A110$

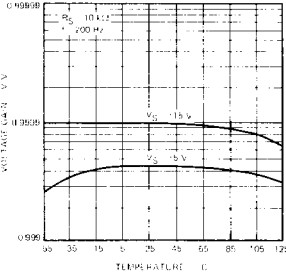
INPUT CURRENT AS A FUNCTION OF TEMPERATURE



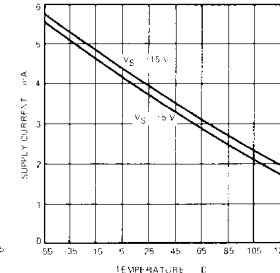
SYMMETRICAL OUTPUT SWING



VOLTAGE GAIN AS A FUNCTION OF TEMPERATURE

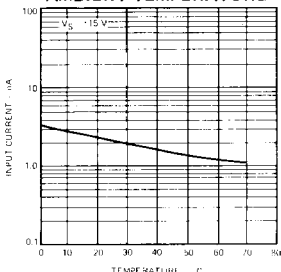


SUPPLY CURRENT AS A FUNCTION OF AMBIENT TEMPERATURE

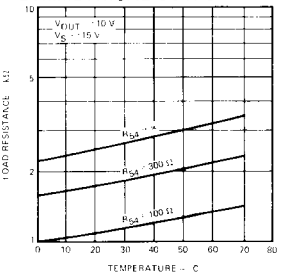


TYPICAL PERFORMANCE CURVES FOR $\mu A302$ • $\mu A310$

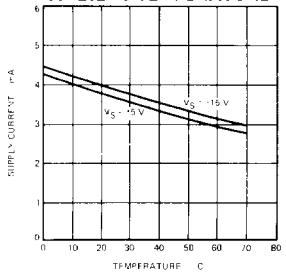
INPUT CURRENT AS A FUNCTION OF AMBIENT TEMPERATURE



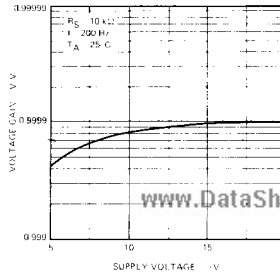
SYMMETRICAL OUTPUT SWING



SUPPLY CURRENT AS A FUNCTION OF AMBIENT TEMPERATURE



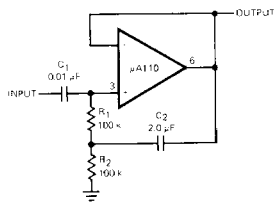
VOLTAGE GAIN AS A FUNCTION OF SUPPLY VOLTAGE



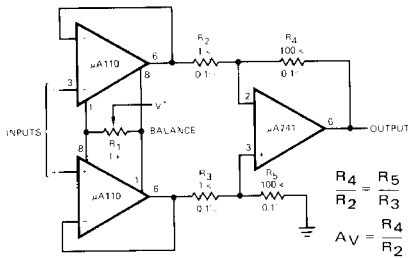
FAIRCHILD • μ A102 • μ A110 μ A302 • μ A310

TYPICAL APPLICATIONS

HIGH INPUT IMPEDANCE AC AMPLIFIER



DIFFERENTIAL INPUT INSTRUMENTATION AMPLIFIER

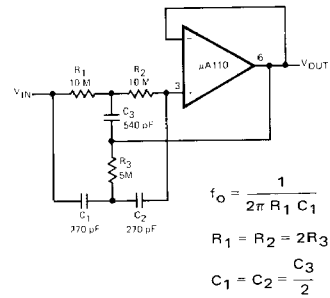


$$R_4 = \frac{R_5}{R_3}$$

$$R_2 = \frac{R_4}{R_2}$$

$$A_V = \frac{R_4}{R_2}$$

HIGH Q NOTCH FILTER

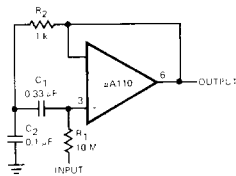


$$f_o = \frac{1}{2\pi R_1 C_1}$$

$$R_1 = R_2 = 2R_3$$

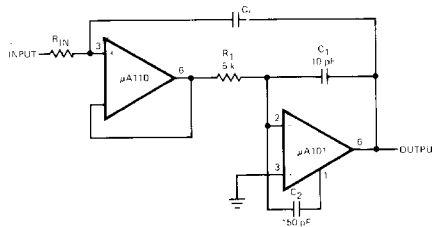
$$C_1 = C_2 = \frac{C_3}{2}$$

BANDPASS FILTER

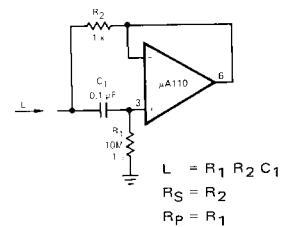


$$f_o = \frac{1}{2\pi\sqrt{R_1 R_2 C_1 C_2}}$$

FAST INTEGRATOR WITH LOW INPUT CURRENT



SIMULATED INDUCTOR

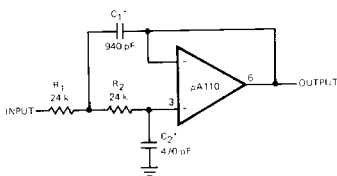


$$L = R_1 R_2 C_1$$

$$R_S = R_2$$

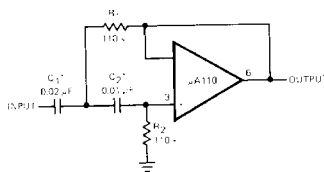
$$R_P = R_1$$

LOW PASS ACTIVE FILTER



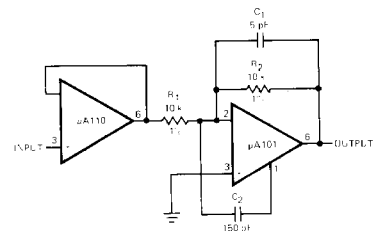
* Values are for 10 kHz cutoff. Use silvered mica capacitors for good temperature stability.

HIGH PASS ACTIVE FILTER

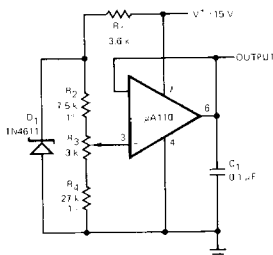


* Values are 100 Hz cutoff. Use metallized polycarbonate capacitors for good temperature stability

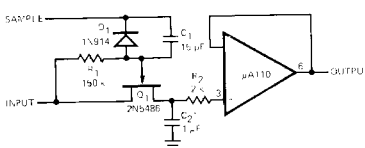
FAST INVERTING AMPLIFIER WITH HIGH INPUT IMPEDANCE



BUFFERED REFERENCE SOURCE

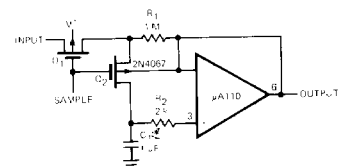


SAMPLE AND HOLD



* Use capacitor with polycarbonate teflon or polyethylene dielectric.

LOW DRIFT SAMPLE AND HOLD**



* Teflon, polyethylene or polycarbonate dielectric capacitor
** Worst case drift less than 3 mV/s