

REGULATING PULSE WIDTH MODULATORS

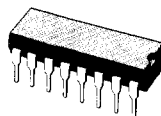
TO 35 V OPERATION
1 V REFERENCE TRIMMED TO $\pm 1\%$
00 Hz TO 500 KHz OSCILLATOR RANGE
SEPARATE OSCILLATOR SYNC TERMINAL
ADJUSTABLE DEADTIME CONTROL
INTERNAL SOFT-START
PULSE-BY-PULSE SHUTDOWN
INPUT UNDERVOLTAGE LOCKOUT WITH
HYSTERESIS
ATCHING PWM TO PREVENT MULTIPLE
PULSES
UAL SOURCE/SINK OUTPUT DRIVERS

shutdown, as well as soft-start recycle with longer shutdown commands. These functions are also controlled by an undervoltage lockout which keeps the outputs off and the soft-start capacitor discharged for sub-normal input voltages. This lockout circuitry includes approximately 500 mV of hysteresis for jitter-free operation. Another feature of these PWM circuits is a latch following the comparator. Once a PWM pulses has been terminated for any reason, the outputs will remain off for the duration of the period. The latch is reset with each clock pulse. The output stages are totem-pole designs capable of sourcing or sinking in excess of 200 mA. The SG1525A output stage features NOR logic, giving a LOW output for an OFF state. The SG1527A utilizes OR logic which results in a HIGH output level when OFF.

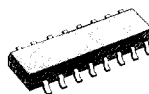
DESCRIPTION

SG1525A/1527A series of pulse width modulated integrated circuits are designed to offer improved performance and lowered external parts count. They are used in designing all types of switching power supplies. The on-chip $+5.1\text{ V}$ reference is trimmed to $\pm 1\%$ and the input common-mode range of the amplifier includes the reference voltage eliminating external resistors. A sync input to the oscillator allows multiple units to be slaved or a single unit synchronized to an external system clock. A resistor between the C_T and the discharge pins provides a wide range of dead time adjustment. These devices also feature built-in soft-circuitry with only an external timing capacitor required. A shutdown terminal controls both the soft-circuitry and the output stages, providing instantaneous turn off through the PWM latch with pulsed

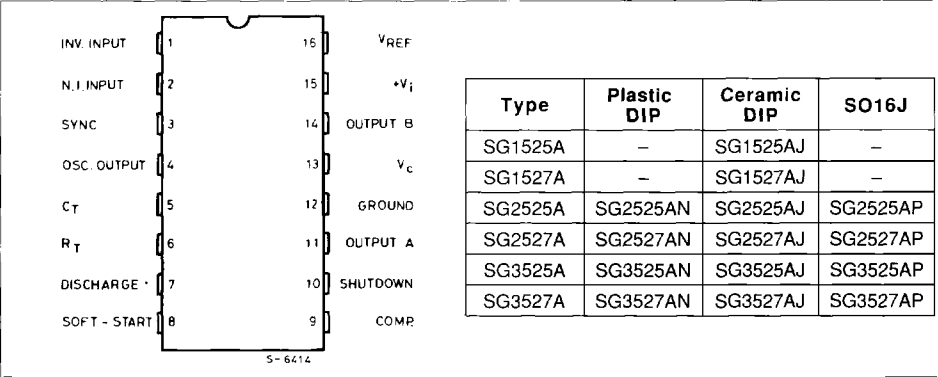
DIP-16
(Plastic -0.25 and Ceramic)



SO16J



CONNECTION DIAGRAM AND ORDERING NUMBERS (top view)



ABSOLUTE MAXIMUM RATINGS

Symbol	Parameter	Value	Unit
V _i	Supply Voltage	40	V
V _c	Collector Supply Voltage	40	V
I _{osc}	Oscillator Charging Current	5	mA
I _o	Output Current, Source or Sink	500	mA
I _R	Reference Output Current	50	mA
I _T	Current through C _T Terminal Logic Inputs Analog Inputs	5 - 0.3 to + 5.5 - 0.3 to V _i	mA V V
P _{tot}	Total Power Dissipation at T _{amb} = 70 °C	1000	mW
T _j	Junction Temperature Range	- 55 to 150	°C
T _{stg}	Storage Temperature Range	- 65 to 150	°C
T _{op}	Operating Ambient Temperature : SG1525A/27A SG2525A/27A SG3525A/27A	- 55 to 125 - 25 to 85 0 to 70	°C °C °C

THERMAL DATA (DIP-16)

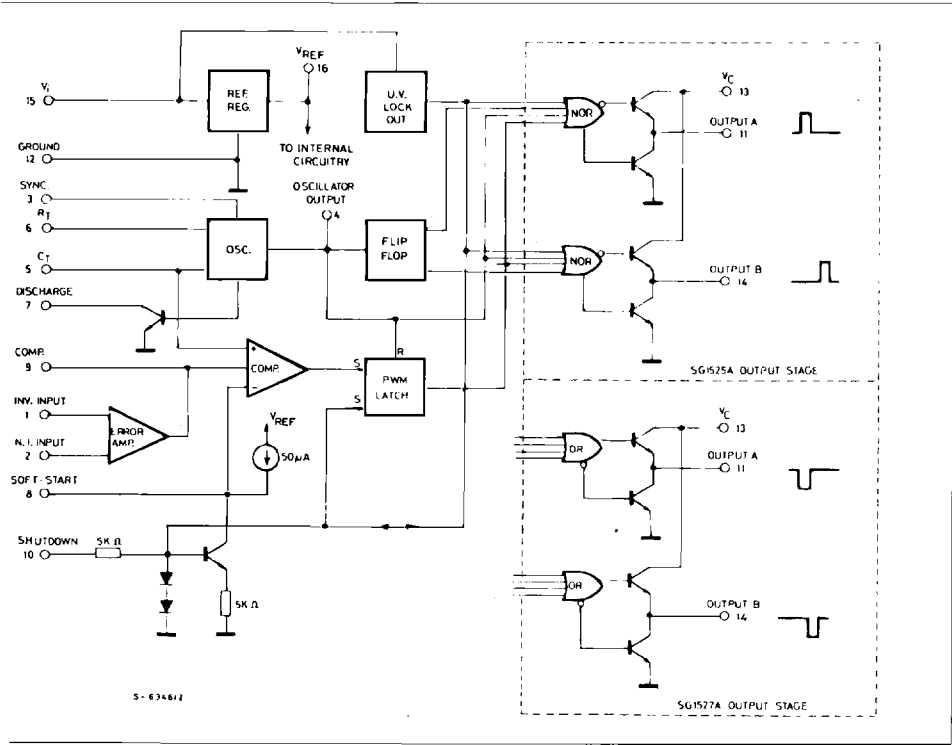
			Ceramic	Plastic
R _{th j-pins}	Thermal Resistance Junction-pins	Max	—	50 °C/W
R _{th j-amb}	Thermal Resistance Junction-ambient	Max	150 °C/W	80 °C/W

THERMAL DATA (SO16J)

R _{th j-alumina} *	Thermal Resistance Junction-alumina	Max	50	°C/W
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* Thermal resistance junction-alumina with the device soldered on the middle of an alumina supporting substrate measuring 15 × 20 mm ; 0 thickness with infinite heatsink.

OCK DIAGRAM



CRITICAL CHARACTERISTICS

(at $V_i = 20\text{ V}$, and over operating temperature, unless otherwise specified)

Symbol	Parameter	Test Conditions	SG1525A/2525A SG1527A/2527A			SG3525A SG3527A			Unit
			Min.	Typ.	Max.	Min.	Typ.	Max.	

REFERENCE SECTION

V_{REF}	Output Voltage	$T_j = 25^\circ\text{C}$	5.05	5.1	5.15	5	5.1	5.2	V
V_{REF}	Line Regulation	$V_i = 8\text{ to }35\text{ V}$		10	20		10	20	mV
V_{REF}	Load Regulation	$I_L = 0\text{ to }20\text{ mA}$		20	50		20	50	mV
$V_{REF}/\Delta T^\circ$	Temp. Stability	Over Operating Range		20	50		20	50	mV
*	Total Output Variation	Line, Load and Temperature	5		5.2	4.95		5.25	V
	Short Circuit Current	$V_{REF} = 0\text{ V}$, $T_j = 25^\circ\text{C}$		80	100		80	100	mA
*	Output Noise Voltage	$10\text{ Hz} \leq f \leq 10\text{ kHz}$, $T_j = 25^\circ\text{C}$		40	200		40	200	μVrms
V_{REF}^*	Long Term Stability	$T_j = 125^\circ\text{C}$, 1000 hrs		20	50		20	50	mV

ELECTRICAL CHARACTERISTICS (continued)

Symbol	Parameter	Test Conditions	SG1525A/2525A SG1527A/2527A			SG3525A SG3527A			Unit
			Min.	Typ.	Max.	Min.	Typ.	Max.	

OSCILLATOR SECTION**

*, *	Initial Accuracy	$T_j = 25\text{ }^{\circ}\text{C}$		± 2	± 6		± 2	± 6	%
*, *	Voltage Stability	$V_i = 8\text{ to }35\text{ V}$		± 0.3	± 1		± 1	± 2	%
$\Delta f/\Delta T^*$	Temperature Stability	Over Operating Range		± 3	± 6		± 3	± 6	%
f_{MIN}	Minimum Frequency	$R_T = 200\text{ K}\Omega$ $C_T = 0.1\text{ }\mu\text{F}$			120			120	Hz
f_{MAX}	Maximum Frequency	$R_T = 2\text{ K}\Omega$ $C_T = 470\text{ pF}$	400			400			KHz
	Current Mirror	$I_{RT} = 2\text{ mA}$	1.7	2	2.2	1.7	2	2.2	nA
*, *	Clock Amplitude		3	3.5		3	3.5		V
*, *	Clock Width	$T_j = 25\text{ }^{\circ}\text{C}$	0.3	0.5	1	0.3	0.5	1	μs
	Sync Threshold		1.2	2	2.8	1.2	2	2.8	V
	Sync Input Current	Sync Voltage = 3.5 V		1	2.5		1	2.5	nA

ERROR AMPLIFIER SECTION ($V_{CM} = 5.1\text{ V}$)

V_{OS}	Input Offset Voltage			0.5	5		2	10	mV
I_b	Input Bias Current			1	10		1	10	nA
I_{OS}	Input Offset Current				1			1	nA
	DC Open Loop Gain	$R_L \geq 10\text{ M}\Omega$	60	75		60	75		dB
*	Gain Bandwidth Product	$G_v = 0\text{ dB}$ $T_j = 25\text{ }^{\circ}\text{C}$	1	2		1	2		MHz
*, z	DC Transconduct.	$30\text{ K}\Omega \leq R_L \leq 1\text{ M}\Omega$ $T_j = 25\text{ }^{\circ}\text{C}$	1.1	1.5		1.1	1.5		mA/V
	Output Low Level			0.2	0.5		0.2	0.5	V
	Output High Level		3.8	5.6		3.8	5.6		V
CMR	Comm. Mode Reject.	$V_{CM} = 1.5\text{ to }5.2\text{ V}$	60	75		60	75		dB
PSR	Supply Voltage Rejection	$V_i = 8\text{ to }35\text{ V}$	50	60		50	60		dB

PWM COMPARATOR

	Minimum Duty-cycle				0			0	%
	Maximum Duty-cycle		45	49		45	49		%
*	Input Threshold	Zero Duty-cycle	0.7	0.9		0.7	0.9		V
		Maximum Duty-cycle		3.3	3.6		3.3	3.6	V
*	Input Bias Current			0.05	1		0.05	1	nA

CRITICAL CHARACTERISTICS (continued)

Symbol	Parameter	Test Conditions	SG1525A/2525A SG1527A/2527A			SG3525A SG3527A			Unit
			Min.	Typ.	Max.	Min.	Typ.	Max.	

TDOWN SECTION

	Soft Start Current	$V_{SD} = 0\text{ V}, \quad V_{SS} = 0\text{ V}$	25	50	80	25	50	80	μA
	Soft Start Low Level	$V_{SD} = 2.5\text{ V}$		0.4	0.7		0.4	0.7	V
	Shutdown Threshold	To outputs, $V_{SS} = 5.1\text{ V}$ $T_J = 25\text{ }^\circ\text{C}$	0.6	0.8	1	0.6	0.8	1	V
	Shutdown Input Current	$V_{SD} = 2.5\text{ V}$		0.4	1		0.4	1	mA
*	Shutdown Delay	$V_{SD} = 2.5\text{ V} \quad T_J = 25\text{ }^\circ\text{C}$		0.2	0.5		0.2	0.5	μs

PUT DRIVERS (each output) ($V_C = 20\text{ V}$)

	Output Low Level	$I_{\text{sink}} = 20\text{ mA}$		0.2	0.4		0.2	0.4	V
		$I_{\text{sink}} = 100\text{ mA}$		1	2		1	2	V
	Output High Level	$I_{\text{source}} = 20\text{ mA}$	18	19		18	19		V
		$I_{\text{source}} = 100\text{ mA}$	17	18		17	18		V
	Under-Voltage Lockout	V_{comp} and $V_{SS} = \text{High}$	6	7	8	6	7	8	V
I_C	Collector Leakage	$V_C = 35\text{ V}$			200			200	μA
t_r^*	Rise Time	$C_L = 1\text{ nF}, \quad T_J = 25\text{ }^\circ\text{C}$		100	600		100	600	ns
t_f^*	Fall Time	$C_L = 1\text{ nF}, \quad T_J = 25\text{ }^\circ\text{C}$		50	300		50	300	ns

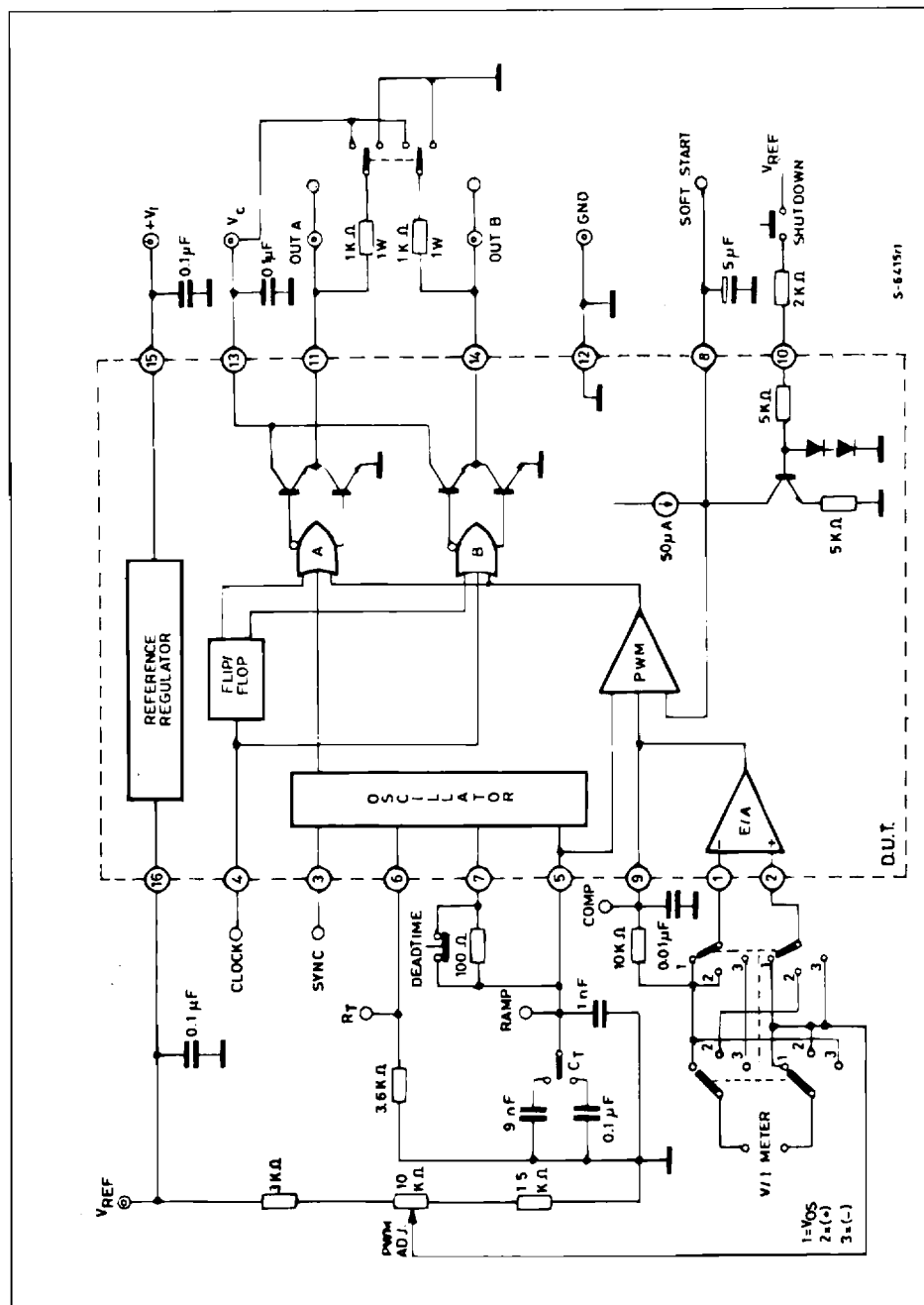
AL STANDBY CURRENT

I_s	Supply Current	$V_i = 35\text{ V}$		14	20		14	20	mA
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These parameters, although guaranteed over the recommended operating conditions, are not 100 % tested in production.
 * Measured at $f_{osc} = 40\text{ KHz}$ ($R_T = 3.6\text{ K}\Omega$, $C_T = 0.1\text{ }\mu\text{F}$, $R_O = 0\text{ }\Omega$). Approximate oscillator frequency is defined by :

$$f_{osc} = \frac{1}{C_T(0.7 R_T + 3 R_O)}$$

Transconductance (g_m) relates to DC open-loop voltage gain (G_v) according to the following equation : $G_v = g_m R_L$, where R_L is the resistance from pin 9 to ground. The minimum g_m specification is used to calculate minimum G_v , when the error amplifier output is loaded.



COMMENDED OPERATING CONDITIONS (*)

Parameter	Value
Input Voltage (V _I)	8 to 35 V
Collector Supply Voltage (V _C)	4.5 to 35 V
Sink/Source Load Current (steady state)	0 to 100 mA
Sink/Source Load Current (peak)	0 to 400 mA
Reference Load Current	0 to 20 mA
Oscillator Frequency Range	100 Hz to 400 KHz
Oscillator Timing Resistor	2 KΩ to 150 KΩ
Oscillator Timing Capacitor	0.001 μF to 0.1 μF
Load Time Resistor Range	0 to 500 Ω

Range over which the device is functional and parameter limits are guaranteed.

Figure 1 : Oscillator Charge Time vs. R_T and C_T.

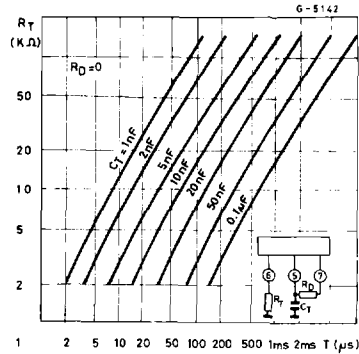


Figure 2 : Oscillator Discharge Time vs. R_D and C_T.

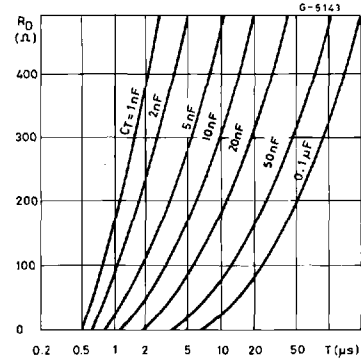


Figure 3 : SG1525A Output Saturation Characteristics.

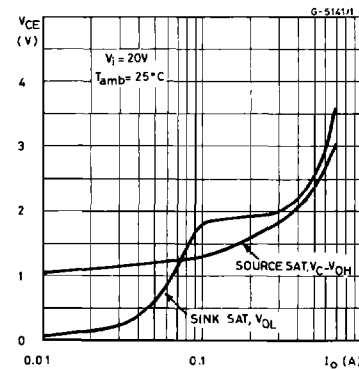


Figure 4 : Error Amplifier Voltage Gain and Phase vs. Frequency.

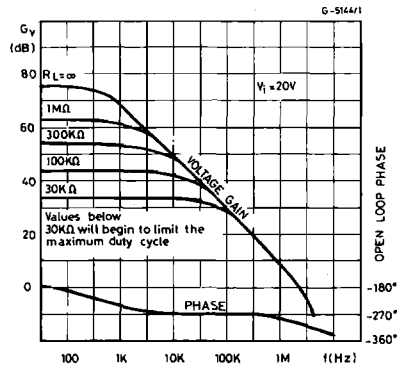
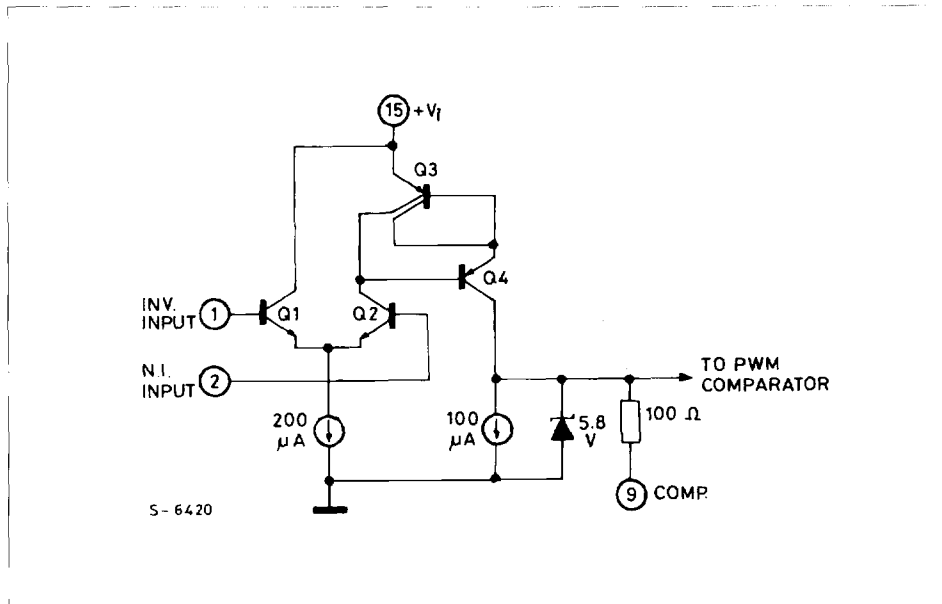


Figure 5 : SG1525A Error Amplifier.

PRINCIPLES OF OPERATION

SHUTDOWN OPTIONS (see Block Diagram)

Since both the compensation and soft-start terminals (Pins 9 and 8) have current source pull-ups, either can readily accept a pull-down signal which only has to sink a maximum of 100 μA to turn off the outputs. This is subject to the added requirement of discharging whatever external capacitance may be attached to these pins.

An alternate approach is the use of the shutdown circuitry of Pin 10 which has been improved to enhance the available shutdown options. Activating this circuit by applying a positive signal on Pin 10 performs two functions : the PWM latch is immedi-

tely set providing the fastest turn-off signal to outputs ; and a 150 μA current sink begins to charge the external soft-start capacitor. If the shutdown command is short, the PWM signal is terminated without significant discharge of the soft-start capacitor, thus, allowing, for example, a converter implementation of pulse-by-pulse current limit. Holding Pin 10 high for a longer duration, however, will ultimately discharge this external capacitor cycling slow turn-on upon release.

Pin 10 should not be left floating as noise pick could conceivably interrupt normal operation.

Figure 6 : SG1525A Oscillator Schematic.

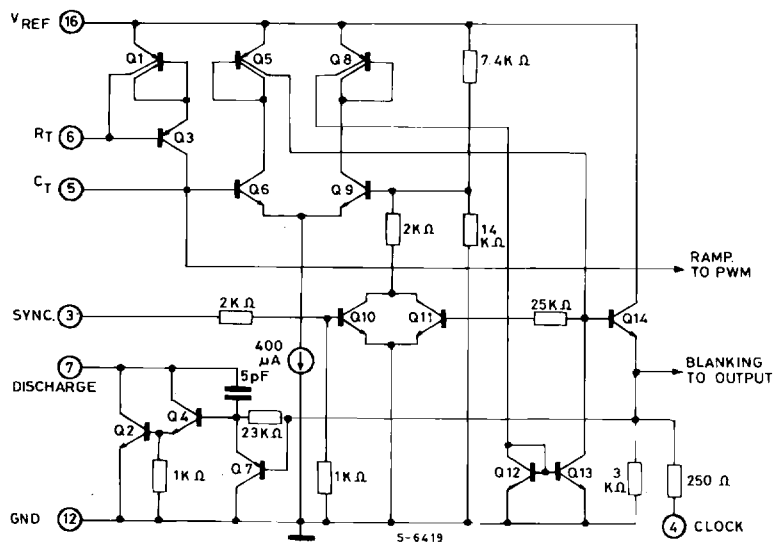


Figure 7 : SG1525A Output Circuit (1/2 circuit shown).

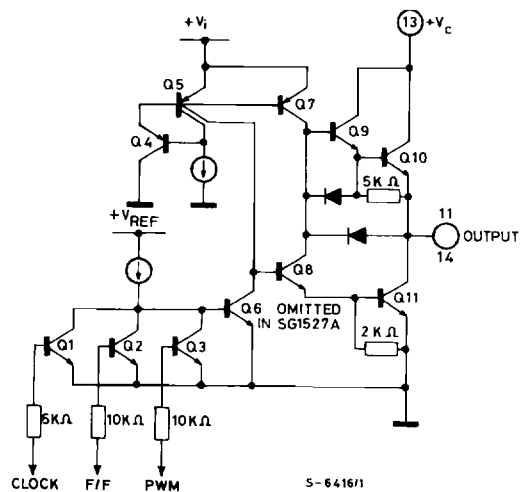
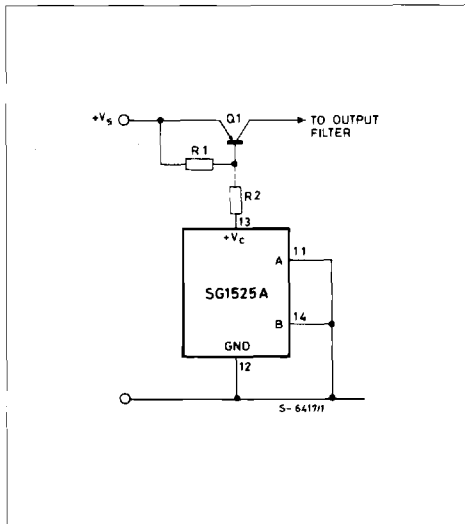
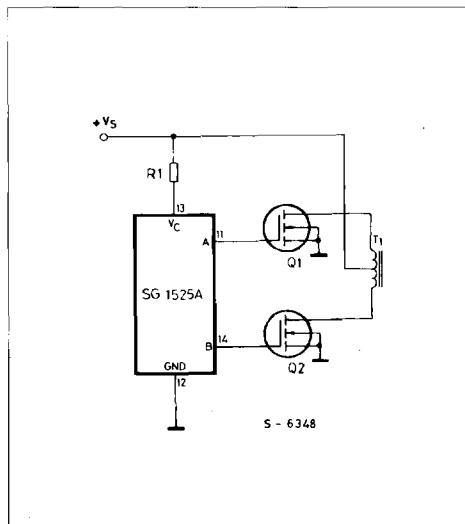


Figure 8.



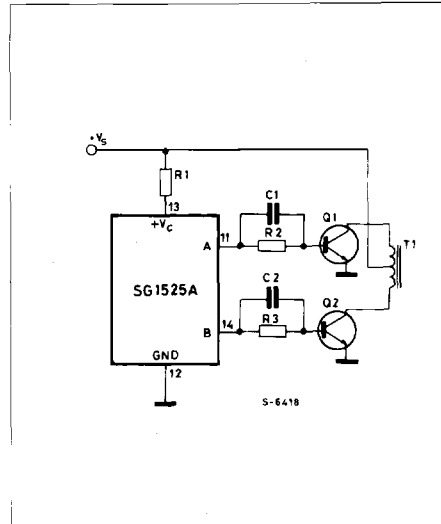
For single-ended supplies, the driver outputs are grounded. The V_c terminal is switched to ground by the totem-pole source transistors on alternate oscillator cycles.

Figure 10.



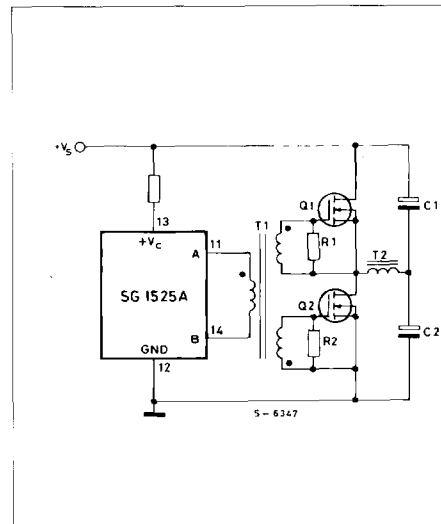
The low source impedance of the output drivers provides rapid charging of Power Mos input capacitance while minimizing external components.

Figure 9.



In conventional push-pull bipolar designs, low base drive is controlled by $R_1 - R_3$. Rapid turn times for the power devices are achieved by speed-up capacitors C_1 and C_2 .

Figure 11.



Low power transformers can be driven directly by the SG1525A. Automatic reset occurs during the time, when both ends of the primary winding are switched to ground.